

Double channel high-side driver with analog current sense for 24 V automotive applications

Datasheet – production data

Features

Max transient supply voltage	V_{CC}	58 V
Operating voltage range	V_{CC}	8 to 36V
Typ On-State resistance (per ch.)	R_{ON}	50 m Ω
Current limitation (typ)	I_{LIM}	34 A
Off state supply current	I_S	2 μ A

- General
 - Very low standby current
 - 3.0 V CMOS compatible input
 - Optimized electromagnetic emission
 - Very low electromagnetic susceptibility
 - Compliance with European directive 2002/95/EC
 - Fault reset standby pin (FR_Stby)
- Diagnostic Functions
 - Proportional load current sense
 - High current sense precision for wide range currents
 - Off-state open load detection
 - Output short to V_{CC} detection
 - Overload and short to ground latch off
 - Thermal shutdown latch-off
 - Very low current sense leakage
- Protections
 - Undervoltage shutdown
 - Overvoltage clamp
 - Load current limitation
 - Self limiting of fast thermal transients
 - Protection against loss of ground and loss of V_{CC}
 - Thermal shutdown
 - Electrostatic discharge protection



Application

- All types of resistive, inductive and capacitive loads

Description

The VND5T050AK-E is a monolithic device made using STMicroelectronics® VIPower® technology, intended for driving resistive or inductive loads with one side connected to ground. Active V_{CC} pin voltage clamp protects the device against low energy spikes.

This device integrates an analog current sense which delivers a current proportional to the load current.

Fault conditions such as overload, overtemperature or short to V_{CC} are reported via the current sense pin.

Output current limitation protects the device in overload condition. The device will latch off in case of overload or thermal shutdown.

The device is reset by a low level pass on the fault reset standby pin.

A permanent low level on the inputs and fault reset standby pin disable all outputs and set the device in standby mode.

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1 Block diagram and pin description

Figure 1. Block diagram

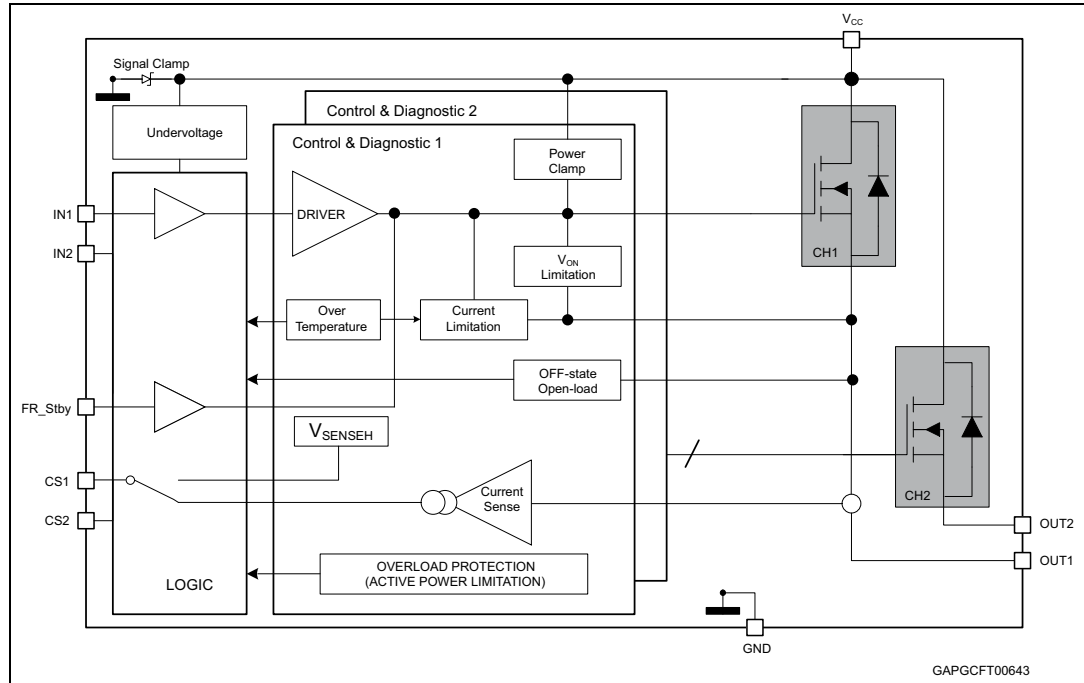


Table 1. Pin function

Name	Function
V _{CC}	Battery connection
OUT _{1,2}	Power output
GND	Ground connection
IN _{1,2}	Voltage controlled input pins with hysteresis, CMOS compatible. They control output switch state
CS _{1,2}	Analog current sense pins, they deliver a current proportional to the load current
FR_Stby	In case of latch-off for overtemperature/overcurrent conditions, a low pulse on the FR_Stby pin is needed to reset the channel. The device enters in standby mode if all inputs and the FR_Stby pin are low.

Figure 2. Configuration diagram PowerSSO-24 (top view)

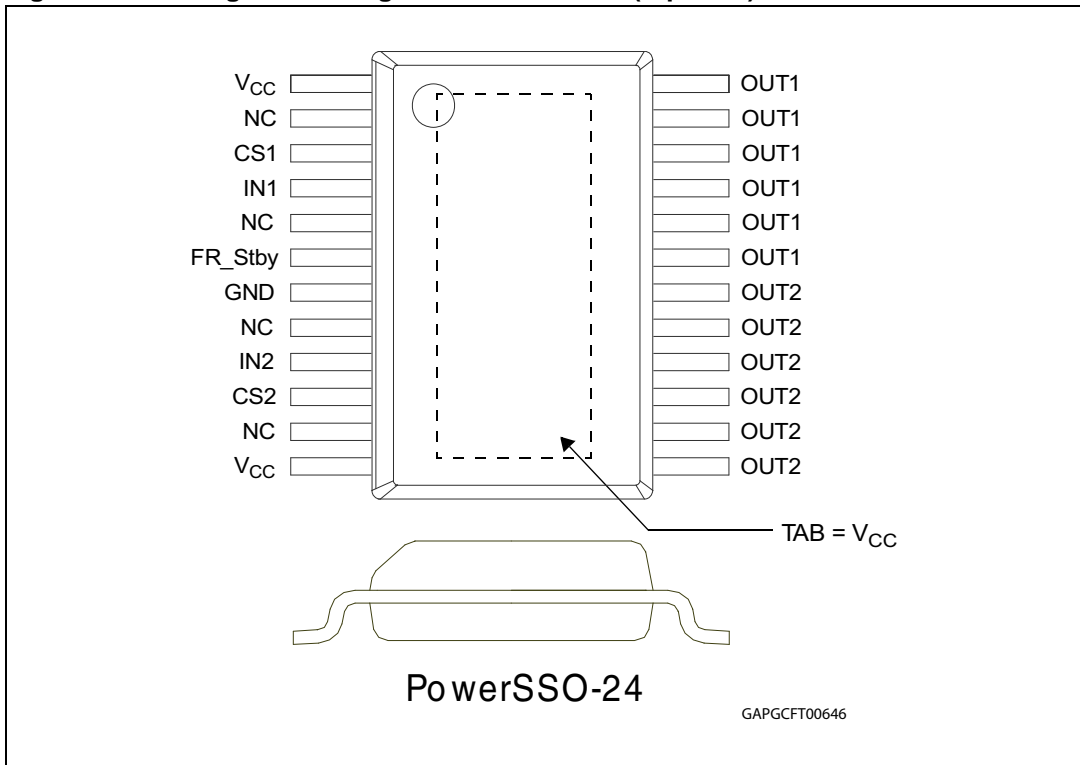


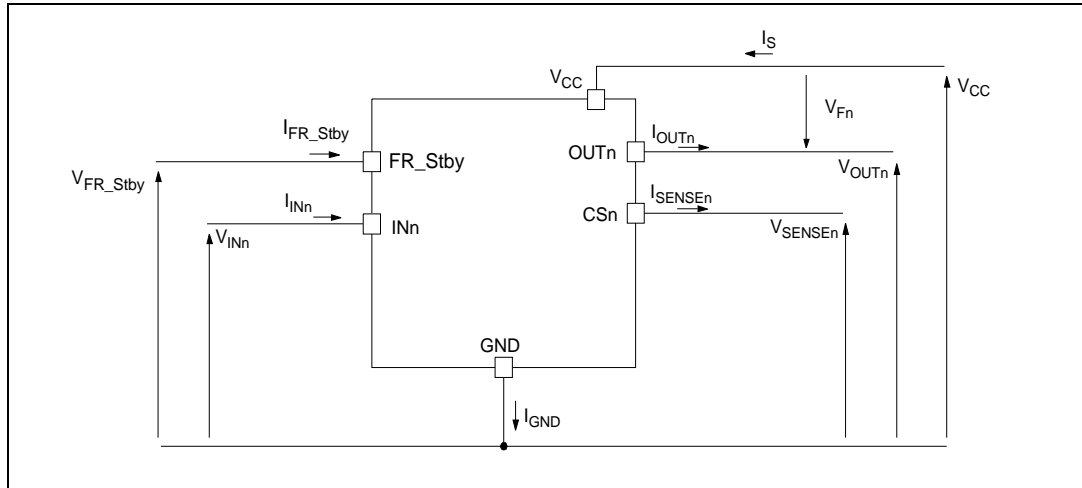
Table 2. Suggested connections for unused and not connected pins

Connection / pin	Current Sense	N.C.	Output	Input	FR_Stby
Floating	Not allowed	X ⁽¹⁾	X	X	X
To ground	Through 10 KΩ resistor	X	Not allowed	Through 10 KΩ resistor	Through 10 KΩ

1. X: do not care.

2 Electrical specifications

Figure 3. Current and voltage conventions



Note: $V_{Fn} = V_{OUTn} - V_{CC}$ during reverse battery condition.

2.1 Absolute maximum ratings

Stressing the device above the ratings listed in [Table 3](#) may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the Operating sections of this specification is not implied. Exposure to the conditions reported in this section for extended periods may affect device reliability.

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{CC}	DC supply voltage	58	V
-V _{CC}	Reverse DC supply voltage	0.3	V
-I _{GND}	DC reverse ground pin current	200	mA
I _{OUT}	DC output current	Internally limited	A
-I _{OUT}	Reverse DC output current	30	A
I _{IN}	DC input current	-1 to 10	mA
I _{FR_Stby}	Fault reset standby DC input current	-1 to 1.5	mA
-I _{CSENSE}	DC reverse CS pin current	200	mA
V _{CSENSE}	Current sense maximum voltage	V _{CC} - 58 to +V _{CC}	V
E _{MAX}	Maximum switching energy L = 50 mH; V _{bat} = 32 V; T _{jstart} = 150°C; I _{OUT} = 2 A	210	mJ

Table 3. Absolute maximum ratings (continued)

Symbol	Parameter	Value	Unit
L_{smax}	Maximum strain inductance in short circuit condition $R_L = 300\text{ m}\Omega$, $V_{batt} = 32\text{ V}$, $T_{jstart} = 150^\circ\text{C}$, $I_{out} = I_{LMHmax}$	40	μH
V_{ESD}	Electrostatic discharge (Human Body Model: $R = 1.5\text{ K}\Omega$; $C = 100\text{ pF}$)		
	– $IN_{1,2}$	4000	V
	– $CS_{1,2}$	2000	V
	– FR_Stby	4000	V
	– $OUT_{1,2}$	5000	V
	– V_{CC}	5000	V
V_{ESD}	Charge device model (CDM-AEC-Q100-011)	750	V
T_j	Junction operating temperature	-40 to 150	$^\circ\text{C}$
T_{stg}	Storage temperature	-55 to 150	$^\circ\text{C}$

2.2 Thermal data

Table 4. Thermal data

Symbol	Parameter	Value	Unit
$R_{thj-case}$	Thermal resistance junction-case (max.) (with one channel ON)	2	$^\circ\text{C}/\text{W}$
$R_{thj-amb}$	Thermal resistance junction-ambient (max.)	See Figure 27	$^\circ\text{C}/\text{W}$

2.3 Electrical characteristics

$8\text{ V} < V_{CC} < 36\text{ V}$; $-40^{\circ}\text{C} < T_j < 150^{\circ}\text{C}$, unless otherwise specified.

Table 5. Power section

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}	Operating supply voltage		8	24	36	V
V_{USD}	Undervoltage shutdown			3.5	5	V
$V_{USDhyst}$	Undervoltage shutdown hysteresis			0.5		V
R_{ON}	On-state resistance ⁽¹⁾	$I_{OUT} = 2\text{ A}$; $T_j = 25^{\circ}\text{C}$		50		m Ω
		$I_{OUT} = 2\text{ A}$; $T_j = 150^{\circ}\text{C}$			100	
V_{clamp}	Clamp voltage	$I_S = 20\text{ mA}$	58	64	70	V
I_S	Supply current	Off-state; $V_{CC} = 24\text{ V}$; $T_j = 25^{\circ}\text{C}$; $V_{IN} = V_{OUT} = V_{SENSE} = 0\text{ V}$		2 ⁽²⁾	5 ⁽²⁾	μA
		On-state; $V_{CC} = 24\text{ V}$; $V_{IN} = 5\text{ V}$; $I_{OUT} = 0\text{ A}$		4.2	6	mA
$I_{L(off)}$	Off-state output current	$V_{IN} = V_{OUT} = 0\text{ V}$; $V_{CC} = 24\text{ V}$; $T_j = 25^{\circ}\text{C}$	0	0.01	3	μA
		$V_{IN} = V_{OUT} = 0\text{ V}$; $V_{CC} = 24\text{ V}$; $T_j = 125^{\circ}\text{C}$	0		5	
V_F	Output - V_{CC} diode voltage	$-I_{OUT} = 2\text{ A}$; $T_j = 150^{\circ}\text{C}$			0.7	V

1. For each channel

2. PowerMOS leakage included

Table 6. Switching ($V_{CC} = 24\text{ V}$; $T_j = 25^{\circ}\text{C}$)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	$R_L = 12\ \Omega$		30		μs
$t_{d(off)}$	Turn-off delay time	$R_L = 12\ \Omega$		40		μs
$dV_{OUT}/dt_{(on)}$	Turn-on voltage slope	$R_L = 12\ \Omega$		0.7 V/us		V/ μs
$dV_{OUT}/dt_{(off)}$	Turn-off voltage slope	$R_L = 12\ \Omega$		0.8 V/us		V/ μs
W_{ON}	Switching energy losses during t_{won}	$R_L = 12\ \Omega$		0.5		mJ
W_{OFF}	Switching energy losses during t_{woff}	$R_L = 12\ \Omega$		0.3		mJ

Table 7. Logic inputs

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{IL}	Input low level voltage				0.9	V
I_{IL}	Low level input current	$V_{IN} = 0.9\text{ V}$	1			μA
V_{IH}	Input high level voltage		2.1			V
I_{IH}	High level input current	$V_{IN} = 2.1\text{ V}$			10	μA
$V_{I(hyst)}$	Input hysteresis voltage		0.25			V
V_{ICL}	Input clamp voltage	$I_{IN} = 1\text{ mA}$	5.5		7	V
		$I_{IN} = -1\text{ mA}$		-0.7		V
$V_{FR_Stby_L}$	Fault reset standby low level voltage				0.9	V
$I_{FR_Stby_L}$	Low level fault reset standby current	$V_{FR_Stby} = 0.9\text{ V}$	1			μA
$V_{FR_Stby_H}$	Fault reset standby high level voltage		2.1			V
$I_{FR_Stby_H}$	High level fault reset standby current	$V_{FR_Stby} = 2.1\text{ V}$			10	μA
$V_{FR_Stby(hyst)}$	Fault reset standby hysteresis voltage		0.25			V
$V_{FR_Stby_CL}$	Fault reset standby clamp voltage	$I_{FR_Stby} = 15\text{ mA (}t < 10\text{ ms)}$	11		15	V
		$I_{FR_Stby} = -1\text{ mA}$		-0.7		V
t_{reset}	Overload latch-off reset time	See Figure 4	2		24	μs
t_{stby}	Standby delay	See Figure 5	120		1200	μs

Figure 4. T_{reset} definition

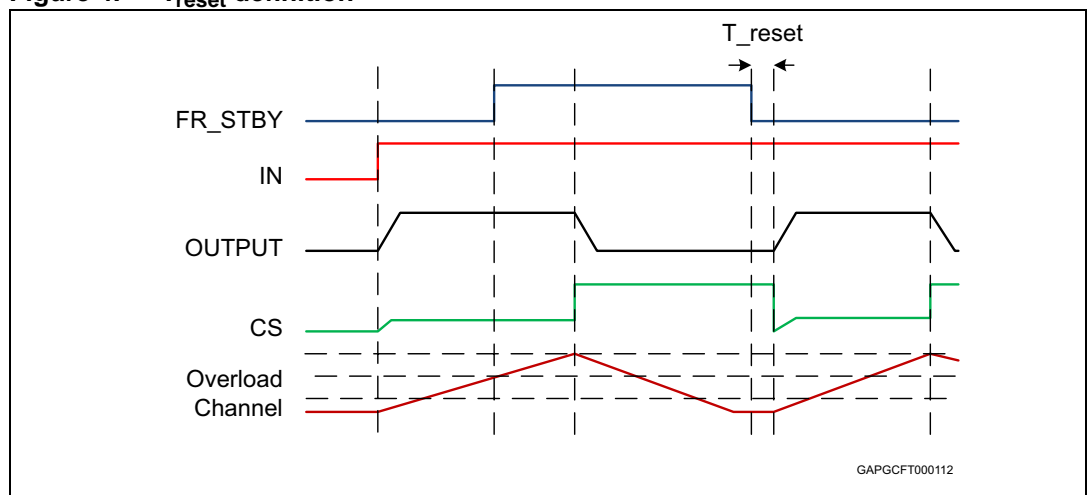


Figure 5. T_{stby} definition

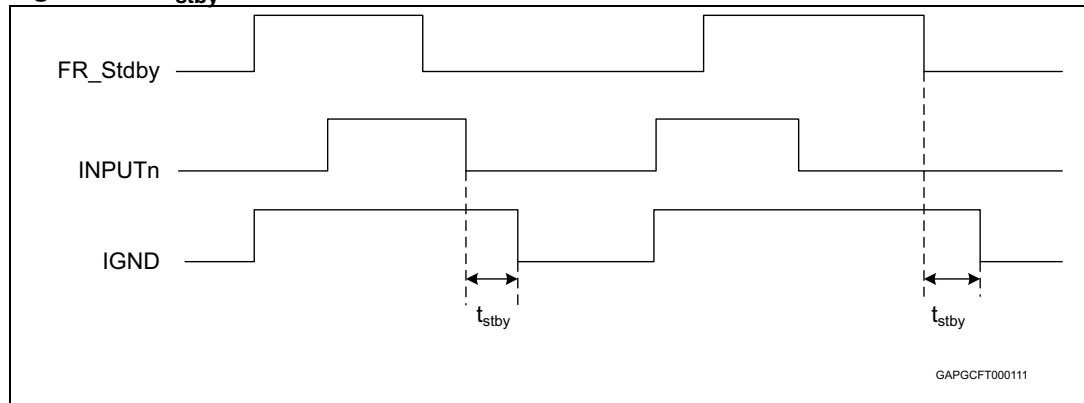


Table 8. Protections and diagnostics

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{limH}	DC short circuit current	$V_{CC} = 24\text{ V}$	24	34	46	A
		$5\text{ V} < V_{CC} < 36\text{ V}$			46	A
I_{limL}	Short circuit current during thermal cycling	$V_{CC} = 24\text{ V};$ $T_R < T_j < T_{TSD}$		8.5		A
T_{TSD}	Shutdown temperature		150	175	200	$^{\circ}\text{C}$
T_R	Reset temperature		$T_{RS} + 1$	$T_{RS} + 5$		$^{\circ}\text{C}$
T_{RS}	Thermal reset of status		135			$^{\circ}\text{C}$
T_{HYST}	Thermal hysteresis ($T_{TSD} - T_R$)			7		$^{\circ}\text{C}$
V_{DEMAG}	Turn-off output voltage clamp	$I_{OUT} = 2\text{ A}; V_{IN} = 0;$ $L = 6\text{ mH}$	$V_{CC} - 58$	$V_{CC} - 64$	$V_{CC} - 70$	V
V_{ON}	Output voltage drop limitation	$I_{OUT} = 100\text{ mA}$		25		mV

Table 9. Current sense ($8\text{ V} < V_{CC} < 36\text{ V}$)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K_{ol}		$I_{OUT} = 10\text{ mA}; V_{SENSE} = 0.5\text{ V};$ $T_j = -40^{\circ}\text{C} \dots 150^{\circ}\text{C}$ $T_j = 25^{\circ}\text{C} \dots 150^{\circ}\text{C}$	564 972	2800	5563 4895	
K_{led}		$I_{OUT} = 50\text{ mA}; V_{SENSE} = 0.5\text{ V};$ $T_j = -40^{\circ}\text{C} \dots 150^{\circ}\text{C}$ $T_j = 25^{\circ}\text{C} \dots 150^{\circ}\text{C}$	1193 1416	2650	4268 3958	
dK/K_{totled}	Current sense ratio drift	$I_{OUT} = 12\text{ mA to } 60\text{ mA};$ $V_{SENSE} = 0.5\text{ V}; I_{cal} = 35\text{ mA}$	-35		35	%
K_0	I_{OUT}/I_{SENSE}	$I_{OUT} = 100\text{ mA}; V_{SENSE} = 0.5\text{ V};$ $T_j = -40^{\circ}\text{C} \dots 150^{\circ}\text{C}$	1409	2540	3726	
$dK_0/K_0^{(1)}$	Current sense ratio drift	$I_{OUT} = 100\text{ mA}; V_{SENSE} = 0.5\text{ V};$ $T_j = -40^{\circ}\text{C to } 150^{\circ}\text{C}$	-20		20	%

Table 9. Current sense ($8\text{ V} < V_{CC} < 36\text{ V}$) (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
K_1	I_{OUT}/I_{SENSE}	$I_{OUT} = 0.7\text{ A}$; $V_{SENSE} = 2\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	1597	2190	2764	
$dK_1/K_1^{(1)}$	Current sense ratio drift	$I_{OUT} = 0.7\text{ A}$; $V_{SENSE} = 2\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	-15		15	%
K_2	I_{OUT}/I_{SENSE}	$I_{OUT} = 2\text{ A}$; $V_{SENSE} = 2\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	1850	2190	2550	
$dK_2/K_2^{(1)}$	Current sense ratio drift	$I_{OUT} = 2\text{ A}$; $V_{SENSE} = 2\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	-10		+10	%
K_3	I_{OUT}/I_{SENSE}	$I_{OUT} = 8\text{ A}$; $V_{SENSE} = 4\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	2050	2190	2280	
$dK_3/K_3^{(1)}$	Current sense ratio drift	$I_{OUT} = 8\text{ A}$; $V_{SENSE} = 4\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	-3		3	%
I_{SENSE0}	Analog sense leakage current	$I_{OUT} = 0\text{ A}$; $V_{SENSE} = 0\text{ V}$; $V_{IN} = 0\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	0		1	μA
		$I_{OUT} = 0\text{ A}$; $V_{SENSE} = 0\text{ V}$; $V_{IN} = 5\text{ V}$; $T_j = -40\text{ }^\circ\text{C} \dots 150\text{ }^\circ\text{C}$	0		2	μA
V_{SENSE}	Max analog sense output voltage	$I_{OUT} = 8\text{ A}$; $R_{SENSE} = 3.9\text{ K}\Omega$	5			V
V_{SENSEH}	Analog sense output voltage in fault condition ⁽²⁾	$V_{CC} = 24\text{ V}$; $R_{SENSE} = 3.9\text{ K}\Omega$	7.5	8.5	9.5	V
I_{SENSEH}	Analog sense output current in fault condition ⁽²⁾	$V_{CC} = 24\text{ V}$; $V_{SENSE} = 5\text{ V}$	4.9	7	12	mA
$t_{DSENSE2H}$	Delay response time from rising edge of INPUT pins	$V_{SENSE} < 4\text{ V}$; $0.15\text{ A} < I_{OUT} < 8\text{ A}$ $I_{SENSE} = 90\%$ of $I_{SENSE\text{ max}}$ (see Figure 6)		150	300	μs
$\Delta t_{DSENSE2H}$	Delay response time between rising edge of output current and rising edge of current sense	$V_{SENSE} < 4\text{ V}$; $I_{SENSE} = 90\%$ of $I_{SENSE\text{ MAX}}$; $I_{OUT} = 90\%$ of $I_{OUT\text{ MAX}}$ $I_{OUT\text{ MAX}} = 2\text{ A}$ (see Figure 10)			250	μs
$t_{DSENSE2L}$	Delay response time from falling edge of INPUT pins	$V_{SENSE} < 4\text{ V}$, $0.15\text{ A} < I_{OUT} < 8\text{ A}$ $I_{SENSE} = 10\%$ of $I_{SENSE\text{ max}}$ (see Figure 6)		5	20	μs

1. Parameter guaranteed by design; it is not tested.

2. Fault condition includes: power limitation, overtemperature and open load in off-state condition.

Table 10. Open-load detection ($V_{FR_STBY} = 5\text{ V}$)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{OL}	Open-load off-state voltage detection threshold	$V_{IN} = 0\text{ V}; 8\text{ V} < V_{CC} < 36\text{ V}$	2	—	4	V
t_{DSTKON}	Output short circuit to V_{CC} detection delay at turn off	See Figure 7	180	—	1800	μs
$I_{L(off2)}$	Off-state output current at $V_{OUT} = 4\text{ V}$	$V_{IN} = 0\text{ V}; V_{SENSE} = 0\text{ V}; V_{OUT}$ rising from 0 V to 4 V	-120	—	0	μA
t_{d_vol}	Delay response from output rising edge to V_{SENSE} rising edge in openload	$V_{OUT} = 4\text{ V}; V_{IN} = 0\text{ V}; V_{SENSE} = 90\% \text{ of } V_{SENSEH}$ $R_{sense} = 3.9\text{ K}$		—	20	μs
t_{DFRSTK_ON}	Output short circuit to V_{CC} detection delay at FRSTBY activation	See Figure 10 ; $\text{Input}_{1,2} = \text{low}$		—	50	μs

Figure 6. Current sense delay characteristics

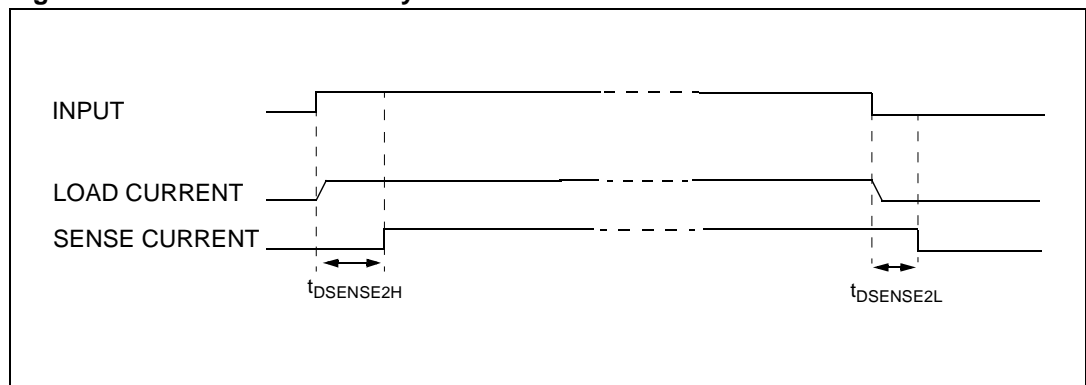
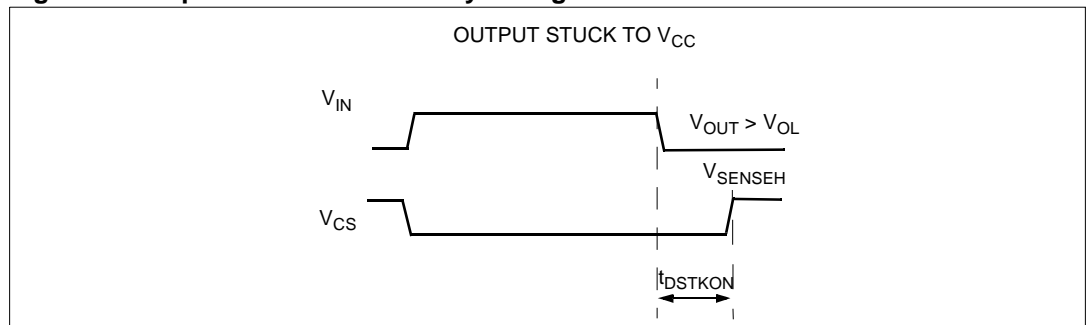


Figure 7. Open-load off-state delay timing



1. $V_{fr_stby} = \text{high}$.

Figure 8. Output stuck to V_{CC} detection delay time at FR_{STBY} activation

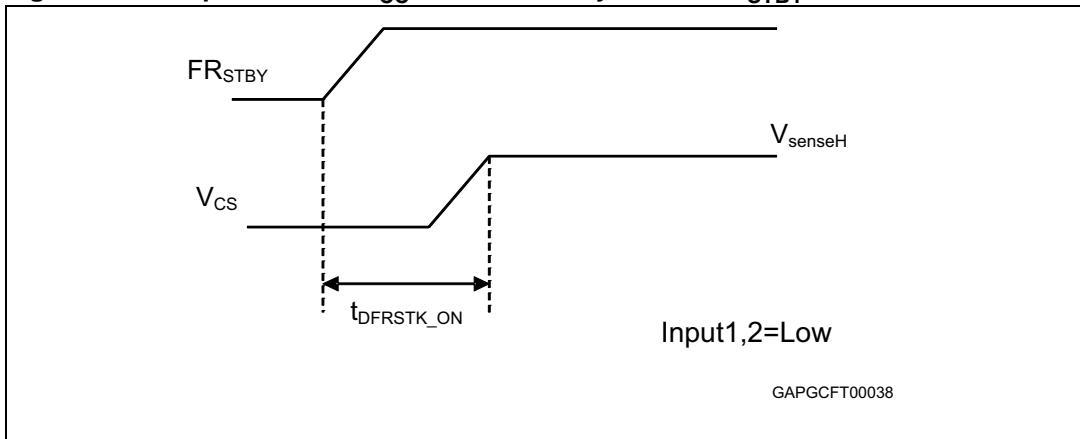


Figure 9. Switching characteristics

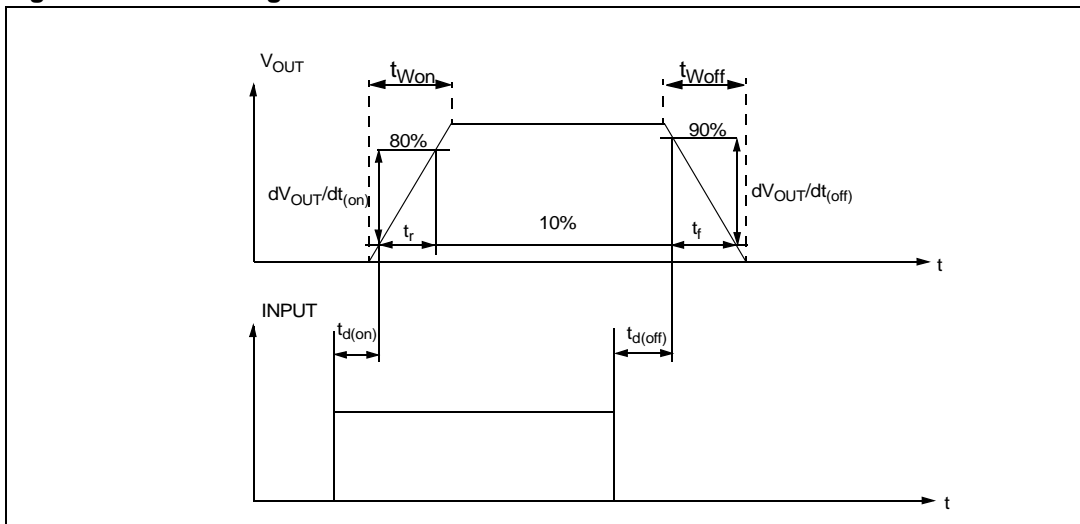


Figure 10. Delay response time between rising edge of output current and rising edge of current sense

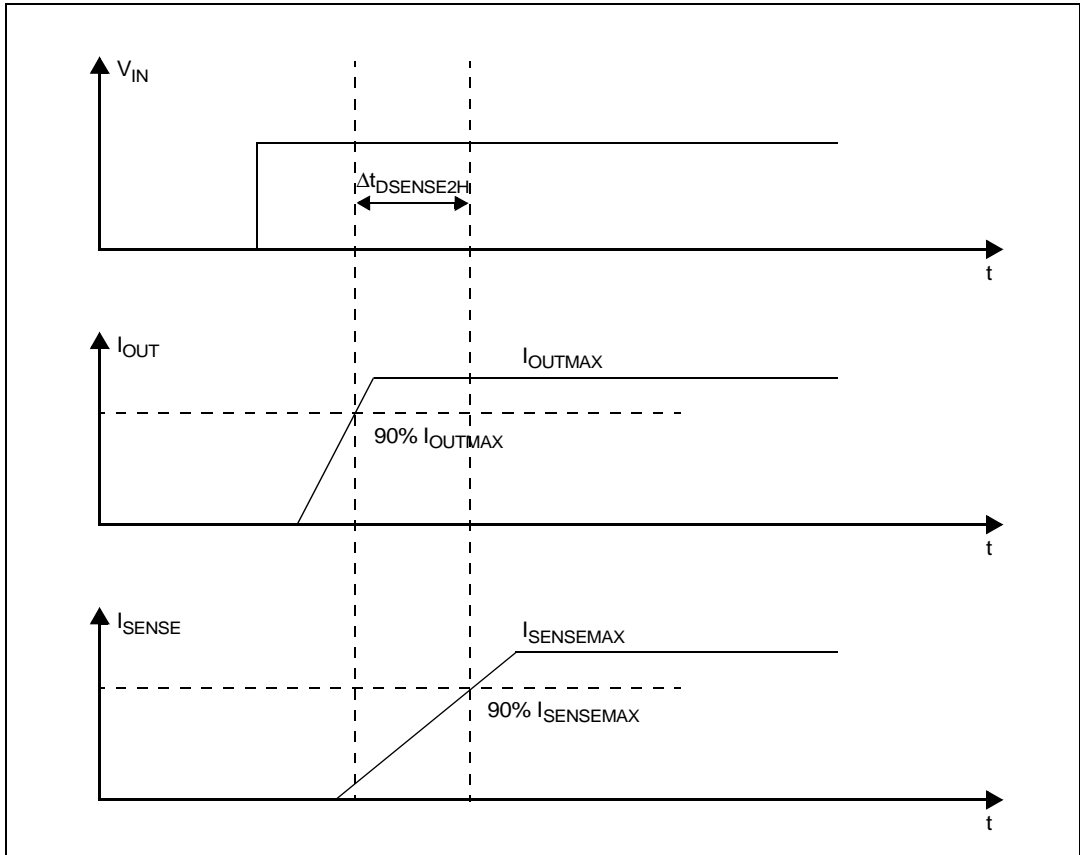


Figure 11. Output voltage drop limitation

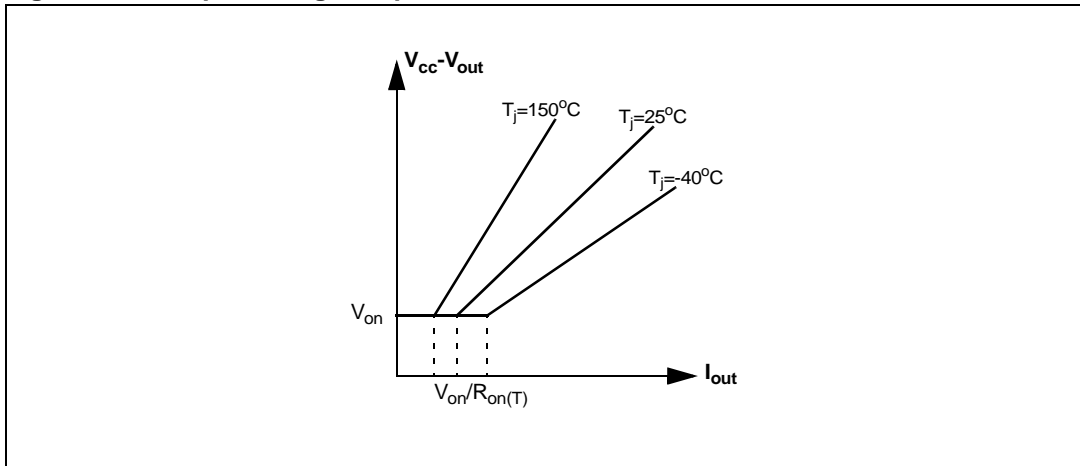


Figure 12. Device behavior in overload condition

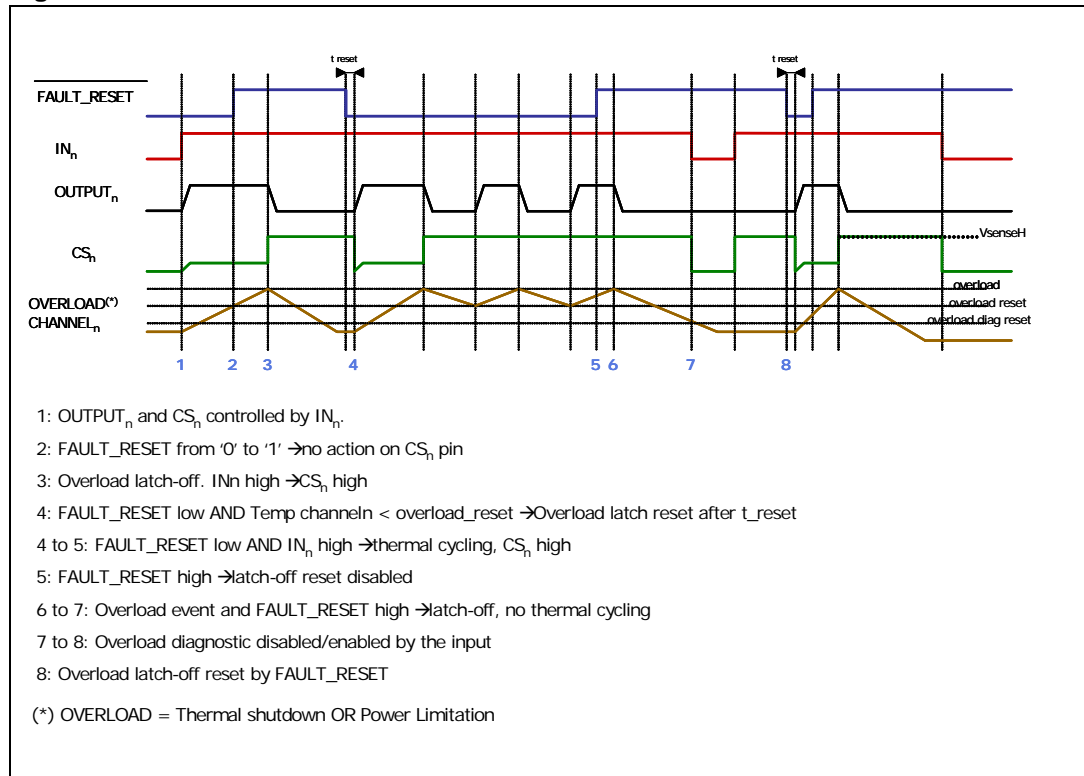


Table 11. Truth table

Conditions	Fault reset standby	Input	Output	Sense
Standby	L	L	L	0
Normal operation	X	L	L	0
	X	H	H	Nominal
Overload	X	L	L	0
	X	H	H	> Nominal
Overtemperature / short to ground	X	L	L	0
	L	H	Cycling	V_{SENSEH}
	H	H	Latched	V_{SENSEH}
Undervoltage	X	X	L	0
Short to V_{BAT}	L	L	H	0
	H	L	H	V_{SENSEH}
	X	H	H	< Nominal
Open load off-state (with pull-up)	L	L	H	0
	H	L	H	V_{SENSEH}
	X	H	H	0
Negative output voltage clamp	X	L	Negative	0

Table 12. Electrical transient requirements (part 1)

ISO 7637-2: 2004(E) Test pulse	Test levels ⁽¹⁾		Number of pulses or test times	Burst cycle/pulse repetition time		Delays and impedance
	III	IV				
1	- 450 V	- 600 V	5000 pulses	0.5 s	5 s	1 ms, 50 Ω
2a	+ 37 V	+ 50 V	5000 pulses	0.2 s	5 s	50 μs, 2 Ω
3a	- 150 V	- 200 V	1h	90 ms	100 ms	0.1 μs, 50 Ω
3b	+ 150 V	+ 200 V	1h	90 ms	100 ms	0.1 μs, 50 Ω
4	- 12 V	- 16 V	1 pulse			100 ms, 0.01 Ω
5b ⁽¹⁾	+ 123 V	+ 174 V	1 pulse			350 ms, 1 Ω

1. Valid in case of external load dump clamp: 58 V maximum referred to ground.

Table 13. Electrical transient requirements (part 2)⁽¹⁾

ISO 7637-2: 2004(E) Test pulse	Test level results	
	III	IV
1	C	C
2a	C	C
3a	C	C
3b ⁽²⁾	E	E
3b ⁽³⁾	C	C
4	C	C
5b ⁽⁴⁾	C	C

1. In order to guarantee the ISO transient classes a minimum 10 KΩ protection resistors are needed on logic pins.
2. Without capacitor between V_{CC} and GND.
3. With 10 nF between V_{CC} and GND.
4. External load dump clamp, 58 V maximum, referred to ground.

Table 14. Electrical transient requirements (part 3)

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device are not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

2.4 Electrical characteristics curves

Figure 13. Off-state output current

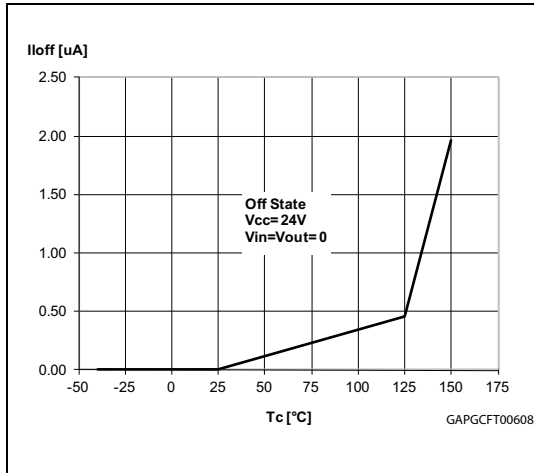


Figure 14. High level input current

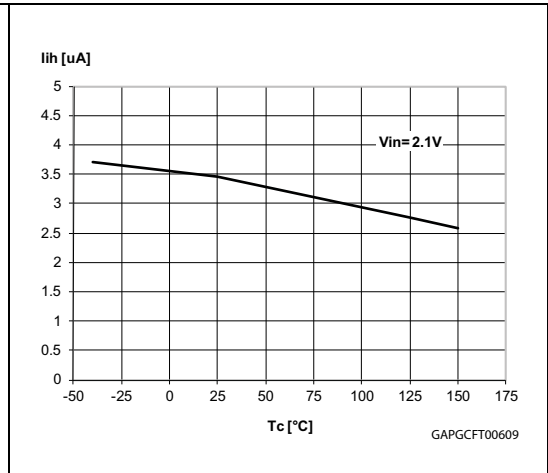


Figure 15. Input clamp voltage

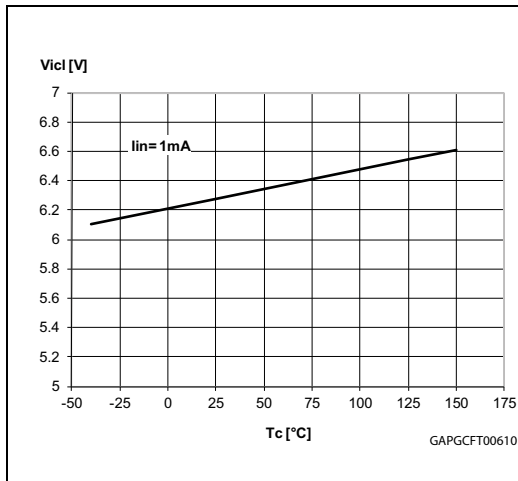


Figure 16. Input low level voltage

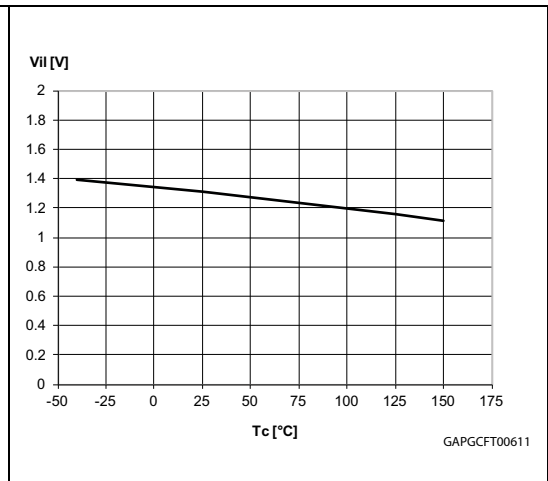


Figure 17. Input high level voltage

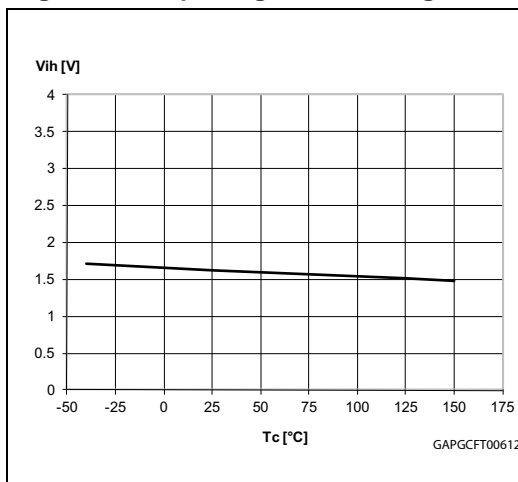


Figure 18. Input hysteresis voltage

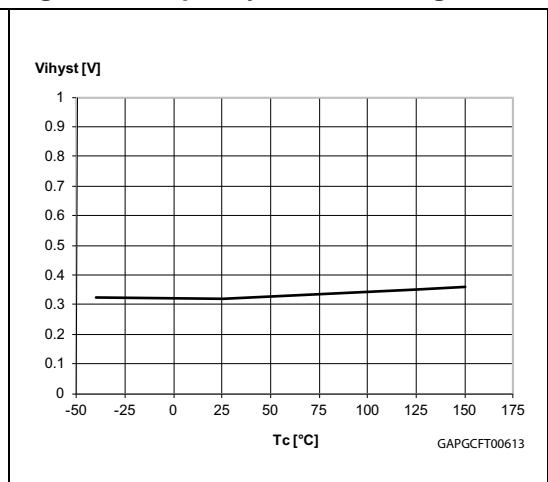


Figure 19. On-state resistance vs T_{case}

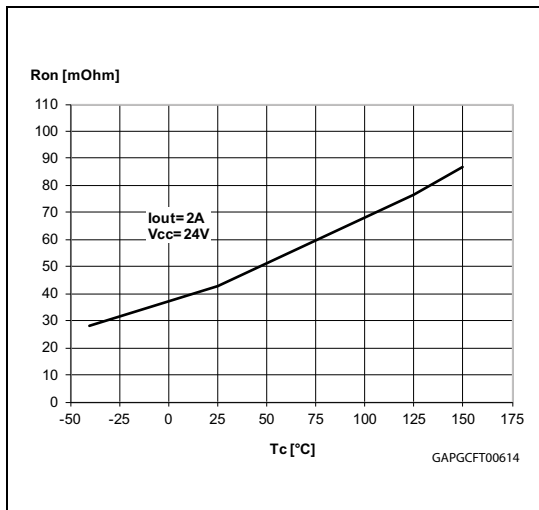


Figure 20. On-state resistance vs V_{CC}

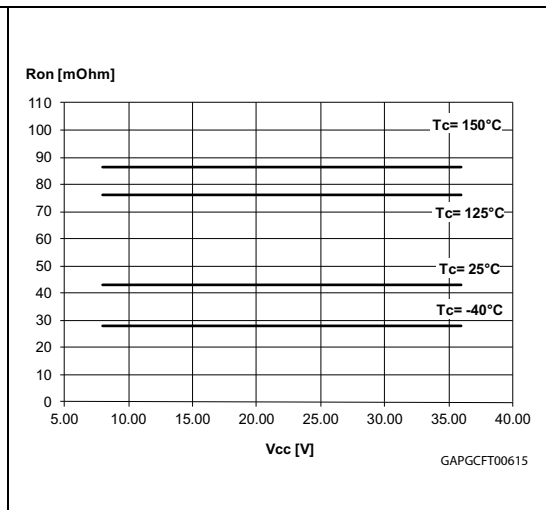


Figure 21. I_{LIMH} vs T_{case}

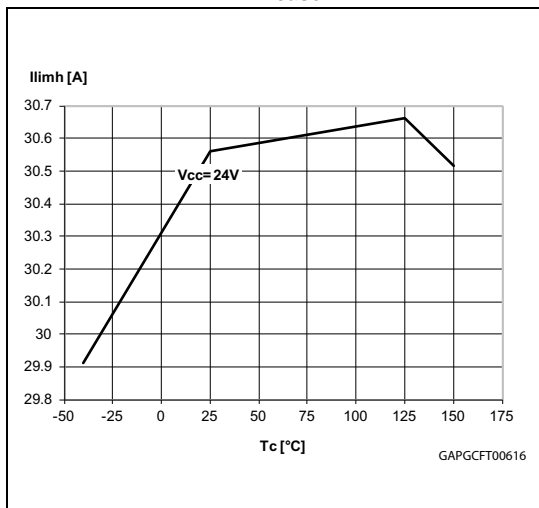


Figure 22. Turn-on voltage slope

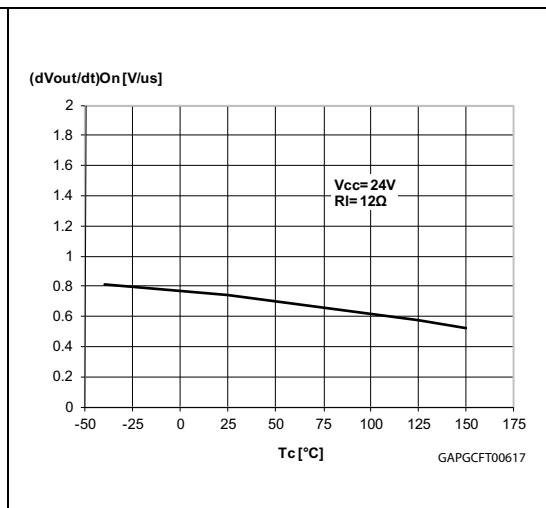
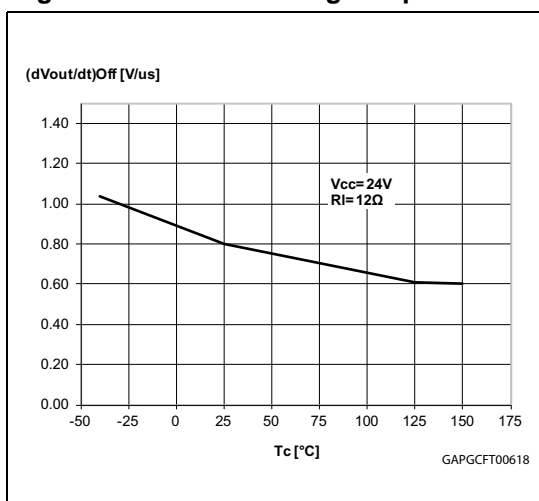
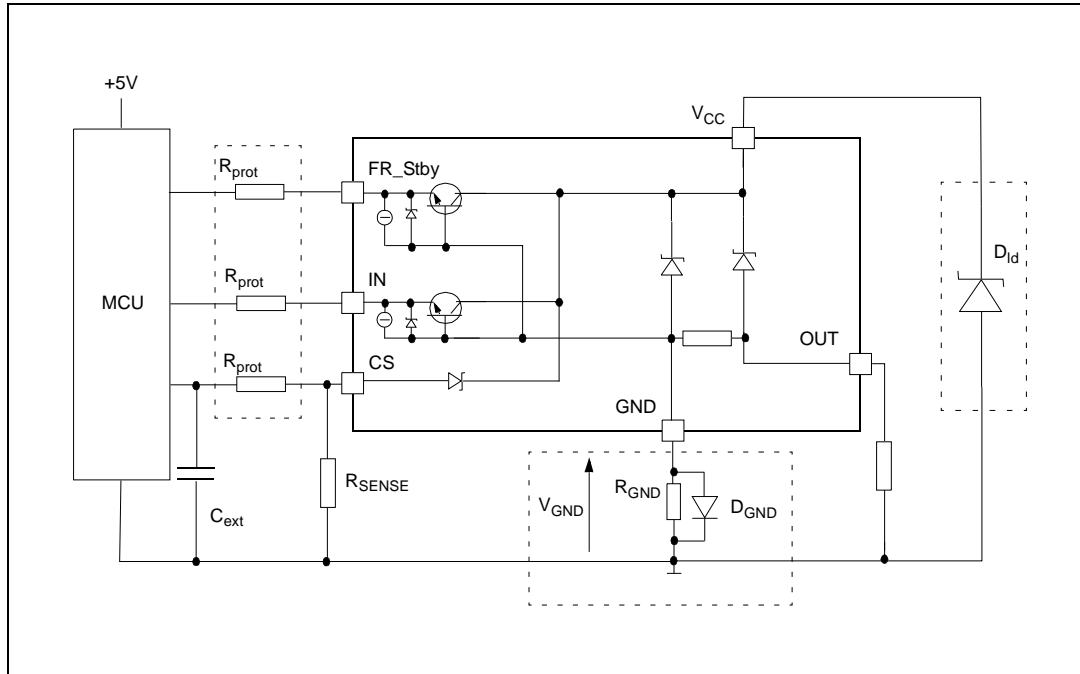


Figure 23. Turn-off voltage slope



3 Application information

Figure 24. Application schematic



3.1 GND protection network against reverse battery

3.1.1 Solution 1: resistor in the ground line (R_{GND} only)

This solution can be used with any type of load.

The following equations are an indication on how to size the R_{GND} resistor.

1. $R_{GND} \leq 600 \text{ mV} / (I_{S(on)max})$.
2. $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where $-I_{GND}$ is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device datasheet.

Power dissipation in R_{GND} (when $V_{CC} < 0$: during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where $I_{S(on)max}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not shared by the device ground then the R_{GND} will produce a shift ($I_{S(on)max} * R_{GND}$) in the input thresholds and the status output values. This shift will vary depending on how many devices are ON in the case of several high side drivers sharing the same R_{GND} .

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor then ST suggests Solution 2 is used (see below).

3.1.2 Solution 2: diode (D_{GND}) in the ground line

A resistor ($R_{GND} = 4.7 \text{ k}\Omega$) should be inserted in parallel to D_{GND} if the device drives an inductive load.

This small signal diode can be safely shared amongst several different HSDs. Also in this case, the presence of the ground network will produce a shift ($\approx 600 \text{ mV}$) in the input threshold and in the status output values, if the microprocessor ground is not common to the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

D_{ld} is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds to V_{CC} max DC rating. The same applies if the device is subject to transients on the V_{CC} line that are greater than the ones shown in the ISO T/R 7637/2 table.

3.3 MCU I/Os protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins will be pulled negative. ST suggests that a resistor (R_{prot}) be inserted in line to prevent the microcontroller I/O pins from latching-up.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os.

$$-V_{CCpeak}/I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

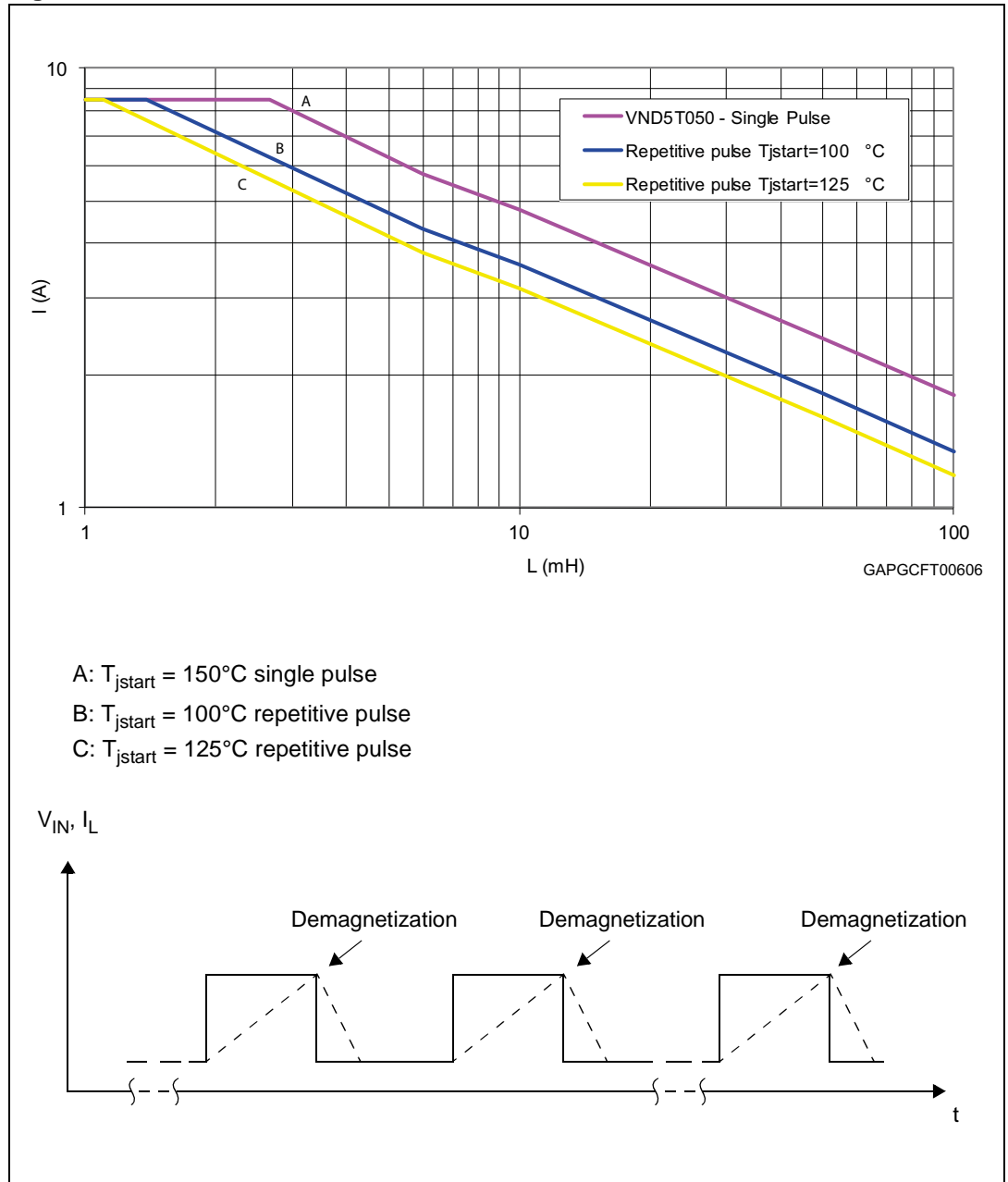
For $V_{CCpeak} = -600 \text{ V}$ and $I_{latchup} \geq 20 \text{ mA}$; $V_{OH\mu C} \geq 4.5 \text{ V}$

$30 \text{ k}\Omega \leq R_{prot} \leq 180 \text{ k}\Omega$.

Recommended R_{prot} value is $60 \text{ k}\Omega$.

3.4 Maximum demagnetization energy ($V_{CC} = 24\text{ V}$)

Figure 25. Maximum turn-off current versus inductance

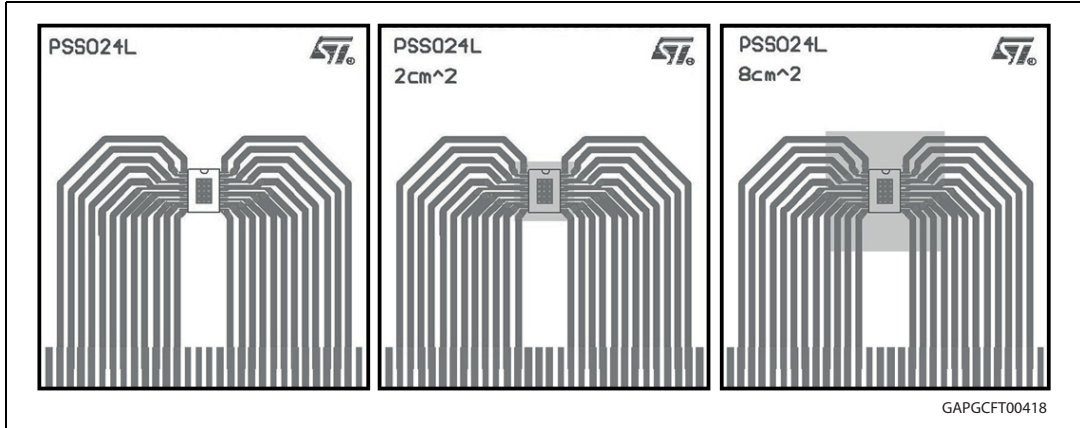


1. Values are generated with $R_L = 0\ \Omega$.
In case of repetitive pulses, T_{jstart} (at the beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves A and B.

4 Package and PCB thermal data

4.1 PowerSSO-24 thermal data

Figure 26. PowerSSO-24 PC board



1. Layout condition of R_{th} and Z_{th} measurements (board finish thickness 1.6 mm +/- 10%; board double layer; board dimension 77x86; board Material FR4; Cu thickness 0.070mm (front and back side); thermal vias separation 1.2 mm; thermal via diameter 0.3 mm +/- 0.08 mm; Cu thickness on vias 0.025 mm; footprint dimension 4.1 mm x 6.5 mm).

Figure 27. $R_{thj-amb}$ vs PCB copper area in open box free air condition (one channel ON)

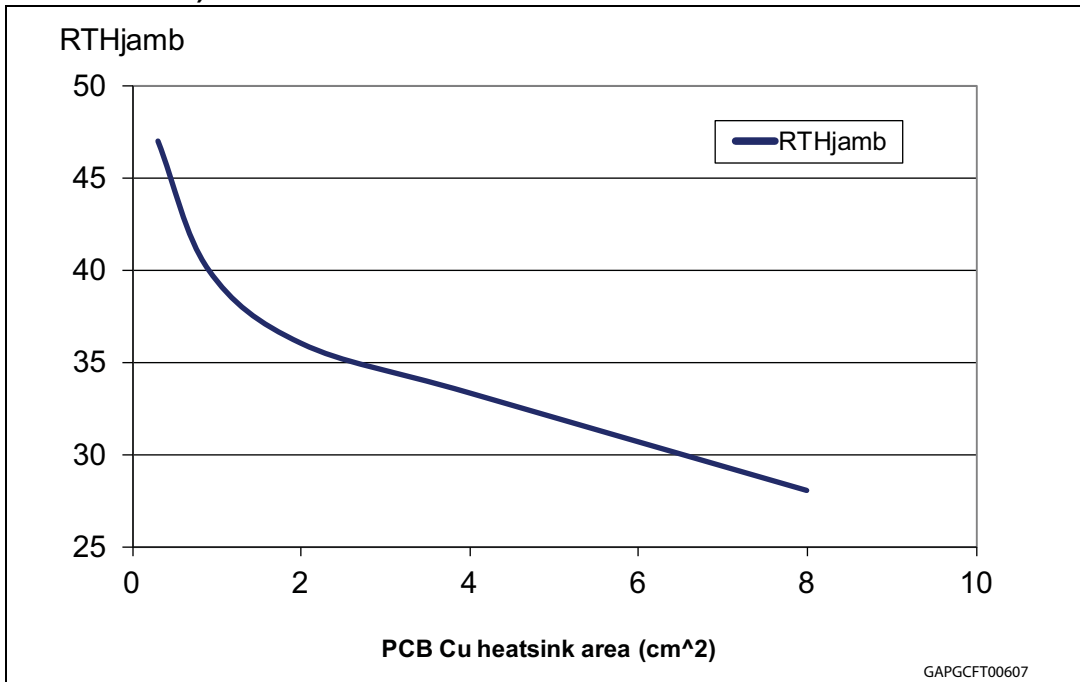


Figure 28. PowerSSO-24 thermal impedance junction ambient single pulse (one channel ON)

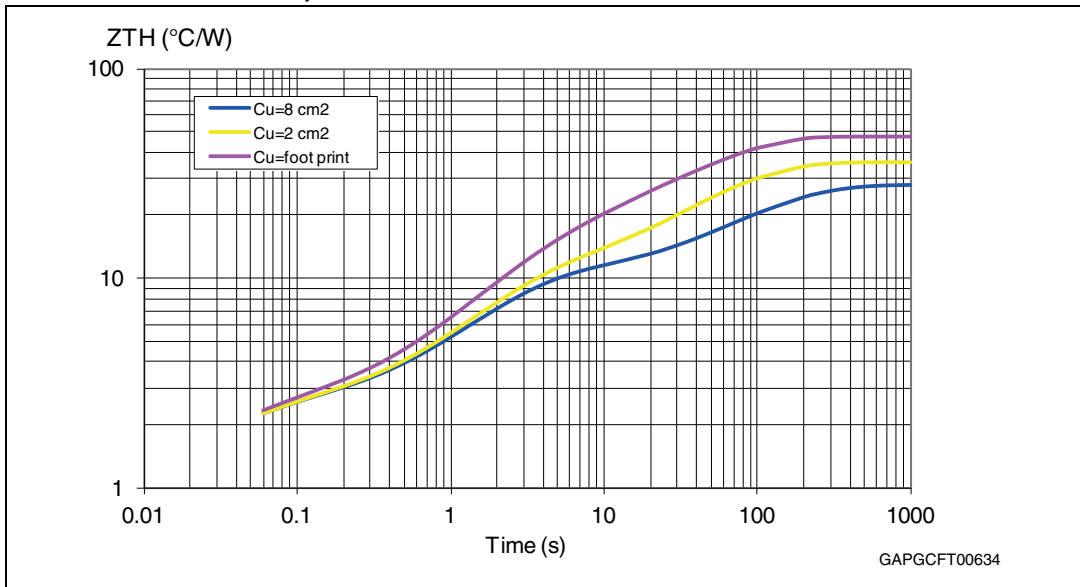
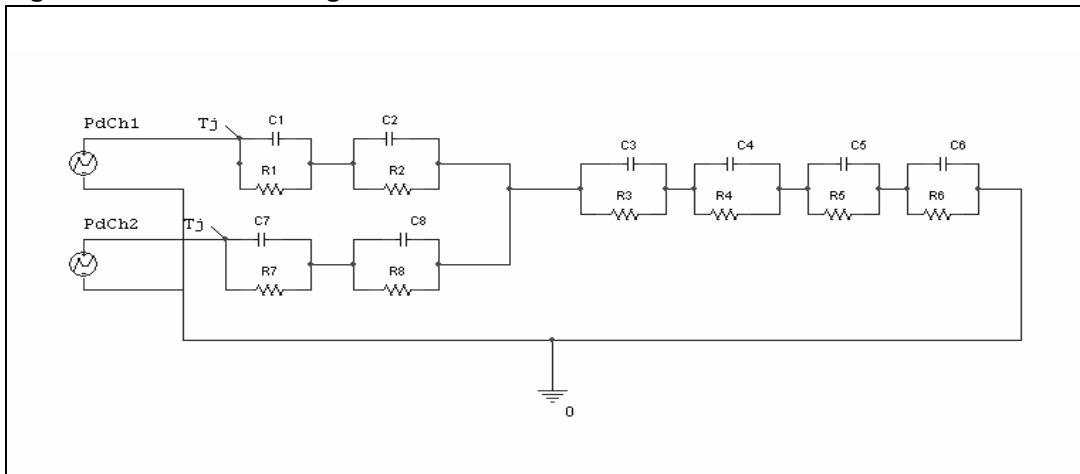


Figure 29. Thermal fitting model of a double channel HSD in PowerSSO-24



1. The fitting model is a simplified thermal tool and is valid for transient evolutions where the embedded protections (power limitation or thermal cycling during thermal shutdown) are not triggered

Equation 1: Pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where $\delta = t_p/T$

Table 15. Thermal parameters

Area/island (cm ²)	Footprint	2	8
R1 = R7 (°C/W)	0.6		
R2 = R8 (°C/W)	0.75		
R3 (°C/W)	1		
R4 (°C/W)	7.7		
R5 (°C/W)	9	9	8
R6 (°C/W)	28	17	10
C1 = C7 (W.s/°C)	0.005		
C2 = C8 (W.s/°C)	0.01		
C3 (W.s/°C)	0.05		
C4 (W.s/°C)	0.3		
C5 (W.s/°C)	1	4	9
C6 (W.s/°C)	2.2	5	17

5 Package and packing information

5.1 ECOPACK®

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com.

ECOPACK® is an ST trademark.

5.2 PowerSSO-24 mechanical data

Figure 30. PowerSSO-24 package dimensions

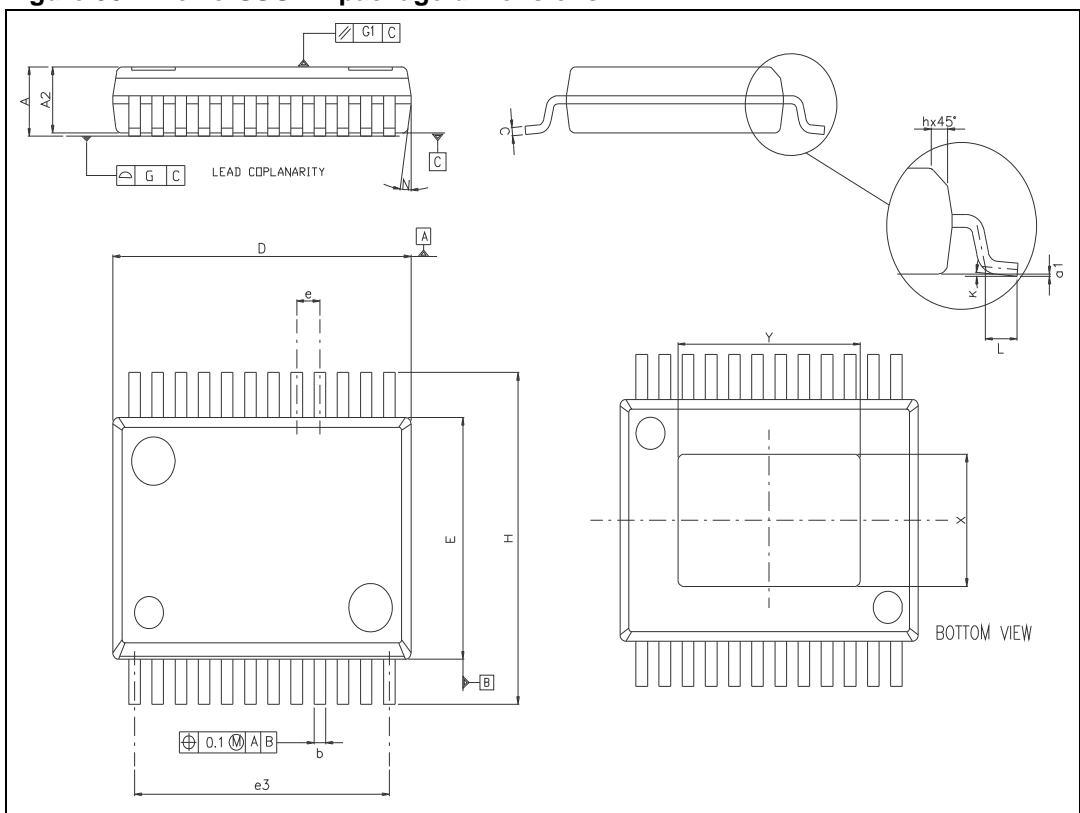


Table 16. PowerSSO-24 mechanical data

Symbol	Millimeters		
	Min.	Typ.	Max.
A	2.15		2.47
A2	2.15		2.40
a1	0		0.075
b	0.33		0.51
c	0.23		0.32
D	10.10		10.50
E	7.4		7.6
e		0.8	
e3		8.8	
G			0.1
G1			0.06
H	10.1		10.5
h			0.4
k		5°	
L	0.55		0.85
N			10°
X	4.1		4.7
Y	6.5		7.1

5.3 PowerSSO-24 packing information

Figure 31. PowerSSO-24 tube shipment (no suffix)

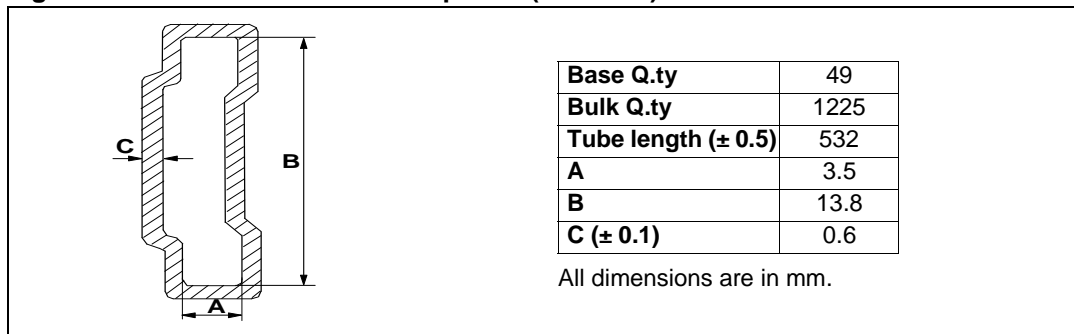
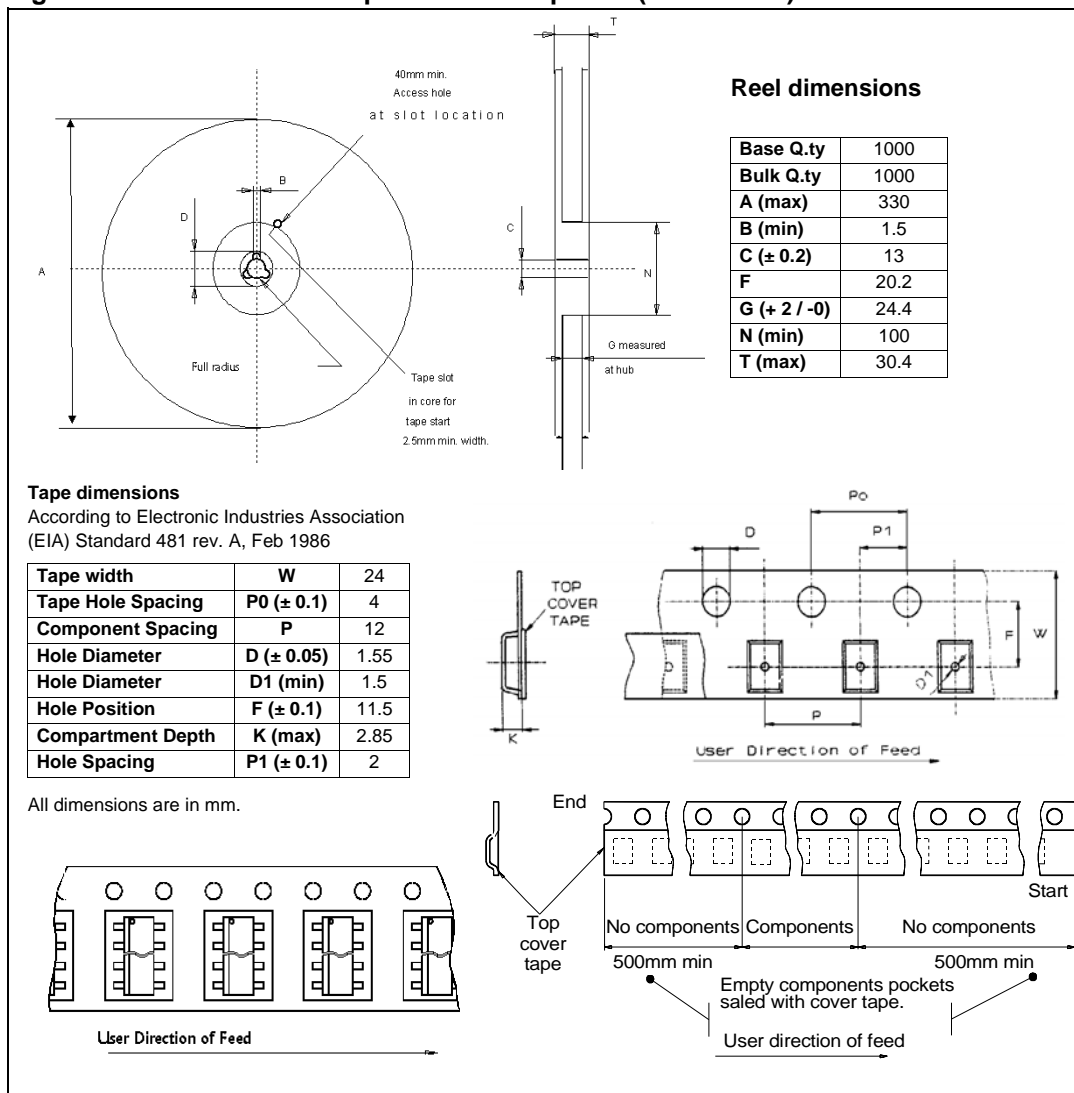


Figure 32. PowerSSO-24 tape and reel shipment (suffix "TR")



6 Order codes

Table 17. Device summary

Package	Order codes	
	Tube	Tape and reel
PowerSSO-24	VND5T050AK-E	VND5T050AKTR-E

7 Revision history

Table 18. Document revision history

Date	Revision	Changes
07-Feb-2012	1	Initial release.
30-Mar-2012	2	Updated <i>Table 2: Suggested connections for unused and not connected pins</i> <i>Table 9: Current sense (8 V < V_{CC} < 36 V):</i> – dK ₀ /K ₀ : updated test condition from I _{OUT} = 100 A to I _{OUT} = 100 mA <i>Table 13: Electrical transient requirements (part 2):</i> – added note
18-Sep-2013	3	Updated disclaimer.

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