











SLOS489C - DECEMBER 2005-REVISED APRIL 2015

TS321

TS321 Low-Power Single Operational Amplifier

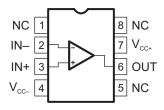
Features

- Wide Power-Supply Range
 - Single Supply from 3 V to 30 V
 - Dual Supply from ±1.5 V to ±15 V
- Large Output Voltage Swing from 0 V to 3.5 V (Minimum) ($V_{CC} = 5 \text{ V}$)
- Low Supply Current at 500 µA (Typical)
- Low Input Bias Current at 20 nA (Typical)
- Stable With High Capacitive Loads

Applications

- Desktop PCs
- HVAC: Heating, Ventilating, and Air Conditioning
- Portable Media Players
- Refrigerators
- Washing Machines: High-End and Low-End

Device Pinouts



3 Description

The TS321 is a bipolar operational amplifier for costsensitive applications in which space savings are important.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
TS321	SOIC (14)	4.90 mm × 3.90 mm		
15321	SOT-23 (5)	2.90 mm × 1.60 mm		

(1) For all available packages, see the orderable addendum at the end of the datasheet.

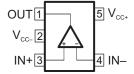




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5 Revision History

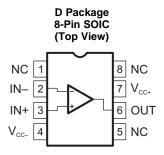
Changes from Revision B (December 2013) to Revision C

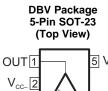
Page

- Added Pin Configuration and Functions section, ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section1



6 Pin Configuration and Functions







Pin Functions

	PIN		1/0	DESCRIPTION				
NAME SOIC NO		SOT-23 NO.	1/0	DESCRIPTION				
IN-	2	4	I	Negative input				
IN+	3	3	1	Positive input				
	1							
NC ⁽¹⁾	5	_	_	Do not connect				
	8							
OUT	6	1	0	Output				
V _{CC} -	4	2	_	Negative supply				
V _{CC+}	7	5	_	Positive supply				

(1) NC - No internal connection



7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
V	Cumplicivalitaria	Single Supply		32	V
V _{CC}	Supply voltage	Dual Supplies		±16	V
V_{ID}	Differential input voltage (2)			±32	V
VI	Input voltage range (3)		-0.3	32	V
I _{IK}	Input current			50	mA
t _{short}	Duration of output short circuit to ground			Unlimited	
TJ	Operating virtual junction temperature			150	°C
T _{stg}	Storage temperature	·	-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2500	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±1500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

			MIN	MAX	UNIT
V	Cumhu valtaga	Single supply	3	30	\/
V _{CC}	Supply voltage Dual suppl	Dual supply	±1.5	±15	V
T _A	Operating free-air temperature		-40	125	°C

7.4 Thermal Information

		TS	321	
	THERMAL METRIC ⁽¹⁾	D (SOIC)	DBV (SOT-23)	UNIT
		5 PIN	5 PIN	
$R_{\theta JA}$	Junction-to-ambient thermal resistance (2)(3)	97	206	°C/W

⁽¹⁾ For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

(3) The package thermal impedance is calculated in accordance with JESD 51-7.

⁽²⁾ Differential voltages are at IN+ with respect to IN-.

⁽³⁾ Input voltages are at IN with respect to V_{CC-}

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

⁽²⁾ Maximum power dissipation is a function of TJ(max), qJA, and TA. The maximum allowable power dissipation at any allowable ambient temperature is PD = (TJ(max) – TA)/qJA. Selecting the maximum of 150°C can affect reliability.



7.5 Electrical Characteristics

 $V_{CC+} = 5 \text{ V}, V_{CC-} = \text{GND}, V_O = 1.4 \text{ V} \text{ (unless otherwise noted)}$

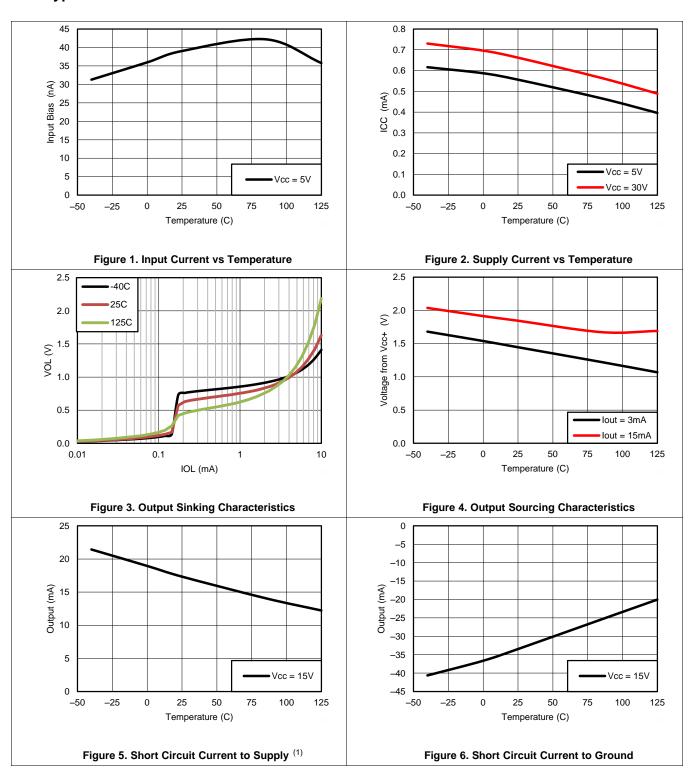
	PARAMETER	TEST CONDIT	IONS	T _A	MIN	TYP	MAX	UNIT
.,		R _S = 0, 5 V < V _{CC+} < 3) V.	25°C		0.5	4	.,
V_{IO}	Input offset voltage	$0 < V_{IC} < (V_{CC+} - 1.5 V_{CC+})$)	Full range			5	mV
				25°C		2	30	
I _{IO}	Input offset current			Full range			50	nA
	Input bias current ⁽¹⁾			25°C		20	150	^
I _{IB}	input bias current.			Full range			200	nA
٨	Large-signal differential voltage	$V_{CC} = 15 \text{ V}, R_{L} = 2 \text{ k}\Omega,$		25°C	50	100		V/mV
A _{VD}	amplification	$V_0 = 1.4 \text{ V to } 11.4 \text{ V}$		Full range	25			V/IIIV
V	Common-mode input voltage ⁽²⁾	V _{CC} = 30 V		25°C	0		V _{CC+} – 1.5	V
V _{ICR}	Common-mode input voltage	vCC = 20 v	Full range	0		V _{CC+} – 2	V	
			$R_L = 2 k\Omega$	25°C	26	27		
		V _{CC} = 30 V	IN 2 KS2	Full range	25.5			
V	High-level output voltage	vCC = 20 v	$R_L = 10 \text{ k}\Omega$	25°C	27	28		V
V _{OH}	r light-level output voltage		K[= 10 K22	Full range	26.5			
		V _{CC} = 5 V	$R_L = 2 k\Omega$	25°C	3.5			
		ACC = 2 A	IN 2 KS2	Full range	3			
V_{OL}	Low-level output voltage	$R_1 = 10 \text{ k}\Omega$		25°C		5	15	mV
VOL	Low-level output voltage	T(_ = 10 K22	Full range			20	111 V	
GBP	Gain bandwidth product	$V_{CC} = 30 \text{ V}, V_I = 10 \text{ m}$ f = 100 kHz, $C_L = 100 \text{ g}$	$'$, $R_L = 2 k\Omega$, oF	25°C		0.8		MHz
SR	Slew rate	$V_{CC} = 15 \text{ V}, V_{I} = 0.5 \text{ V}$ $R_{L} = 2 \text{ k}\Omega, C_{L} = 100 \text{ pF}$	to 3 V, , unity gain	25°C		0.4		V/µs
ϕ_{m}	Phase margin			25°C		60		0
CMRR	Common-mode rejection ratio	R _S ≤ 10 kΩ		25°C	65	85		dB
I _{SOURCE}	Output source current	$V_{CC} = 15 \text{ V}, V_{O} = 2 \text{ V}, V_{O} = 10 \text{ V}$	V _{ID} = 1 V	25°C	20	40		mA
lanus	Output sink current	V _{CC} = 15 V, V _{ID} = 1 V	V _O = 2 V	25°C	10	20		mA
I _{SINK}	Output sink current	vCC = 13 v, vID = 1 v	$V_0 = 0.2 \text{ V}$	25°C	12	50		μΑ
Io	Short-circuit to GND	V _{CC} = 15 V		25°C		40	60	mA
SVR	Supply-voltage rejection ratio	$V_{CC} = 5 \text{ V to } 30 \text{ V}$		25°C	65	110		dB
			$V_{CC} = 5 V$	25°C		500	800	
L	Total supply current	No load	$V_{CC} = 30 \text{ V}$	25 0		600	900	μA
I _{CC}	rotal supply culterit	I VO I O O O	V _{CC} = 5 V	Full range		600	900	μΛ
		V _{CC} = 30 V		i uli range			1000	
THD	Total harmonic distortion	$V_{CC} = 30 \text{ V}, V_{O} = 2 V_{pp}, A_{V} = 20 \text{ dB},$ $R_{L} = 2 \text{ k}\Omega, f = 1 \text{ kHz}, C_{L} = 100 \text{ pF}$		25°C		0.015%		
e _N	Equivalent input noise voltage	$V_{CC} = 30 \text{ V, f} = 1 \text{ kHz,}$	R _S = 100 Ω	25°C		50		nV/√Hz

⁽¹⁾ The direction of the input current is out of the device. This current essentially is constant, independent of the state of the output, so no loading change exists on the input lines.

⁽²⁾ The input common-mode voltage of either input signal should not be allowed to go negative by more than 0.3 V. The upper end of the common-mode voltage range is V_{CC+} – 1.5 V, but either or both inputs can go to 32 V without damage.

TEXAS INSTRUMENTS

7.6 Typical Characteristics



(1) Short circuits from outputs to VCC can cause excessive heating and eventual destruction.

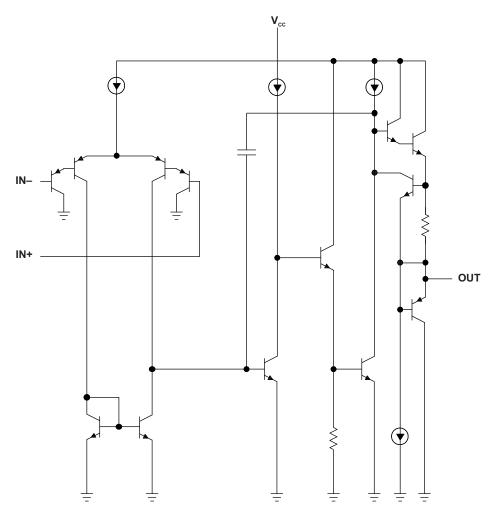


8 Detailed Description

8.1 Overview

The TS321 is a single-channel operational amplifier. It can handle a single supply between 3 V and 30 V or a dual-supply between ±1.5 V and ±15 V. Available in the small SOT-23 package, the TS321 is great for saving space in any application.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Operating Voltage

The TS321 can be powered from a single supply between 3 V and 30 V or a dual-supply between ± 1.5 V and ± 15 V.

8.3.2 Gain Bandwidth Product

Gain bandwidth product is found by multiplying an amplifier's measured bandwidth by the gain at which that bandwidth was measured. The TS321 has a gain bandwidth of 0.8 MHz.

8.3.3 Slew Rate

The slew rate is the rate at which an operational amplifier can change its output when there is a change on the input. The TS321 has a 0.4-V/ μ s slew rate.

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Feature Description (continued)

8.3.4 Input Common Mode Range

The valid common mode range is from device ground pin to VCC - 1.5 V (VCC - 2 V across temperature). Inputs may exceed VCC up to the maximum VCC without device damage. At least one input must be in the valid input common mode range for output to be correct phase. If both inputs exceed valid range then output phase is undefined. If either input is less than -0.3 V then input current should be limited to 1 mA and output phase is undefined.

8.3.5 Stability With High Capacitive Loads

Operational amplifiers have reduced phase margin when there is a direct capacitance on the output. The stability is affected most when the amplifier is set to unity gain. Small signal response to a step input of 100 mV reveals the loop stability with a range of capacitors. See application note SLVA381 to correlate response waveform to phase margin. The responses at 1 nF or less indicate acceptable phase margin. The responses at 1 uF and above indicate good phase margin.

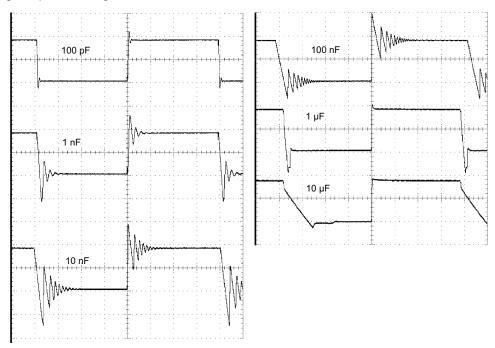


Figure 7. Small Signal Response

8.4 Device Functional Modes

The TS321 is powered on when the supply is connected. This device can be operated as a single supply operational amplifier or dual supply amplifier depending on the application.



Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The TS321 operational amplifier is useful in a wide range of signal conditioning applications. Inputs can be powered before VCC for flexibility in multiple supply circuits.

9.2 Typical Application

A typical application for an operational amplifier in an inverting amplifier. This amplifier takes a positive voltage on the input, and makes it a negative voltage of the same magnitude. In the same manner, it also makes negative voltages positive.

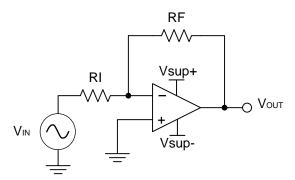


Figure 8. Typcial Application Schematic

9.2.1 Design Requirements

The supply voltage must be chosen such that it is larger than the input voltage range and output range. For instance, this application will scale a signal of ±0.5 V to ±1.8 V. Setting the supply at ±12 V is sufficient to accommodate this application.

9.2.2 Detailed Design Procedure

Determine the gain required by the inverting amplifier:

$$A_{v} = \frac{VOUT}{VIN}$$

$$A_{v} = \frac{1.8}{-0.5} = -3.6$$
(2)

Once the desired gain is determined, choose a value for RI or RF. Choosing a value in the kilohm range is desirable because the amplifier circuit will use currents in the milliamp range. This ensures the part will not draw too much current. This example will choose 10 k Ω for RI which means 36 k Ω will be used for RF. This was determined by Equation 3.

$$A_v = -\frac{RF}{RI} \tag{3}$$

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Typical Application (continued)

9.2.3 Application Curve

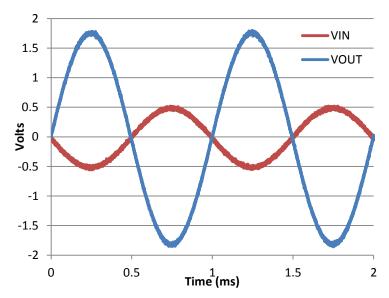


Figure 9. Input and output voltages of the inverting amplifier

10 Power Supply Recommendations

The TS321 is specified to operate between 3 V and 30 V or a dual supply between ±1.5 V and ±15 V.

CAUTION

Supply voltages larger than 32 V for a single supply, or outside the range of ± 16 V for a dual supply can permanently damage the device (see the *Absolute Maximum Ratings*).

Place 0.1-µF bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high impedance power supplies. For more detailed information on bypass capacitor placement, refer to the *Layout*.

11 Layout

11.1 Layout Guidelines

For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole, as well as the
 operational amplifier. Bypass capacitors are used to reduce the coupled noise by providing low impedance
 power sources local to the analog circuitry.
 - Connect low-ESR, 0.1-µF ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from V+ to ground is applicable for single supply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective
 methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes.
 A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital
 and analog grounds, paying attention to the flow of the ground current. For more detailed information, refer to
 application note SLOA089.
- To reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If it is not possible to keep them separate, it is much better to cross the sensitive trace perpendicular as



Layout Guidelines (continued)

opposed to in parallel with the noisy trace.

- Place the external components as close to the device as possible. Keeping RF and RG close to the inverting input minimizes parasitic capacitance, as shown in *Layout Example*.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.

11.2 Layout Example

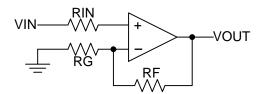


Figure 10. Operational Amplifier Schematic for Noninverting Configuration

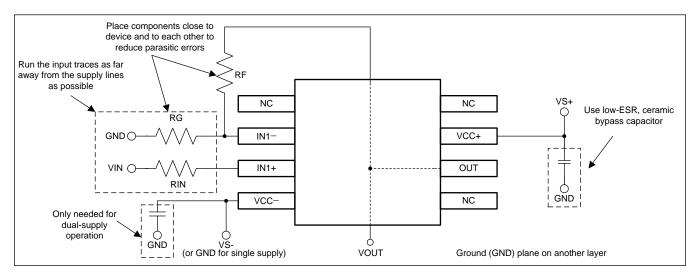


Figure 11. Operational Amplifier Board Layout for Noninverting Configuration

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12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For more information, see the following:

- Simplifying Stability Checks, SLVA381
- Circuit Board Layout Techniques, SLOA089

12.2 Trademarks

All trademarks are the property of their respective owners.

12.3 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.4 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.





28-May-2015

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing		Package Qty		Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
TS321ID	ACTIVE	SOIC	Diawing	8	75	Green (RoHS & no Sb/Br)	(6) CU NIPDAU	(3) Level-1-260C-UNLIM	-40 to 125	(4/5) SR321I	Samples
TS321IDBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU CU SN	Level-1-260C-UNLIM	-40 to 125	(9C1G ~ 9C1S)	Samples
TS321IDBVRE4	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	9C1G	Samples
TS321IDBVRG4	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	9C1G	Samples
TS321IDBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU CU SN	Level-1-260C-UNLIM	-40 to 125	(9C1G ~ 9C1S)	Samples
TS321IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	SR321I	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



PACKAGE OPTION ADDENDUM

28-May-2015

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF TS321:

Automotive: TS321-Q1

NOTE: Qualified Version Definitions:

Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





Α0	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TS321IDBVR	SOT-23	DBV	5	3000	180.0	8.4	3.23	3.17	1.37	4.0	8.0	Q3
TS321IDBVR	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TS321IDBVRG4	SOT-23	DBV	5	3000	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TS321IDBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
TS321IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

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Device	Device Package Type		Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TS321IDBVR	SOT-23	DBV	5	3000	202.0	201.0	28.0
TS321IDBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
TS321IDBVRG4	SOT-23	DBV	5	3000	180.0	180.0	18.0
TS321IDBVT	SOT-23	DBV	5	250	180.0	180.0	18.0
TS321IDR	SOIC	D	8	2500	340.5	338.1	20.6

DBV (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
- D. Falls within JEDEC MO-178 Variation AA.



DBV (R-PDSO-G5)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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