

TLV320AIC3268 Low Power Stereo Audio Codec With 105dB DAC, DirectPath Headphone and Class-D Speaker Amplifier and Integrated miniDSP

105dB SNR for Stereo Audio DAC to Differential • Low Power Portable Audio Devices Lineout • Consumer Audio Devices

[Product](http://www.ti.com/product/TLV320AIC3268?dcmp=dsproject&hqs=pf) Folder

- 95dB SNR for Stereo Line Input to Audio ADC Infotainment Systems
- 8-192kHz Playback and Record
- • 24 mW output power from DirectPathTM **3 Description**
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- -
	- 1.15 W (8Ω, 5V, 1% THDN)
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- Inputs environment.
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- - Serial Interfaces (ASI) Serial Interfaces (ASI) Serial Interfaces (ASI) Serial Interfaces (ASI) Serial Business
	-
-
-
- SPI and I²C Control Interfaces

4 Simplified Diagrams

1 Features 2 Applications

Tools & **[Software](http://www.ti.com/product/TLV320AIC3268?dcmp=dsproject&hqs=sw&#desKit)**

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Headphone Driver The TLV320AIC3268 (AIC3268) provides users with
121 mW output power from Differential Becouse an integrated method to implement an audio system an integrated method to implement an audio system • ¹²¹ mW output power from Differential Receiver with multiple digital and analog audio streams as Output Driver sources, and multiple output devices such as high- Class-D Speaker Driver with **powered** audio amplifiers. The integrated miniDSP – 1.45 W (8Ω, 5V, 10% THDN) allows users to differentiate their products by enabling customization of signal processing on the audio streams. The low power consumption of this device is Differential or Single-Ended Stereo Line Outputs in the last ideal for systems that are battery operated or that
Eight Single-Ended or 4 Fully-Differential Analog operate in a constrained power consumption operate in a constrained power consumption

Analog Microphone Inputs, and Up to 4 The AIC3268 is an integrated, low-power, low-voltage
Simultaneous Digital Microphone Channels stereo audio codec and also features four digital stereo audio codec and also features four digital Extensive Signal Processing Options **Extensive Signal Processing Options Fig. 2** Fig. 2 and *nicrophone* inputs, plus programmable outputs, predefined and parameterizable signal processing Fully-programmable Enhanced miniDSP
blocks and an integrated PLL. Extensive register-
hased control of power input and output channel • Three Independent Digital Audio Serial Interfaces based control of power, input and output channel configuration, gains, effects, terminal-multiplexing and – TDM and mono PCM support on all Audio clocks are included, allowing the device to be

– 8-channel Input and Output on ASI 1 The AIC3268 supports a range of applications from Programmable PLL, plus Low-Frequency Clocking mono 8kHz voice to stereo 192kHz music, making it ideal for wide variety of low power battery-operated Programmable 12-Bit SAR ADC and consumer audio applications.

Device Information

An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, **44** intellectual property matters and other important disclaimers. PRODUCTION DATA.

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Terminal Functions – 64 Terminal QFN (RGC) Package (continued)

7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

It's recommended to keep all AVDDx_18 supplies within ± 50 mV of each other.

(3) It's recommended to keep SVDD and SPK_V supplies within ±50 mV of each other.

7.2 Handling Ratings

7.3 Recommended Operating Conditions

(1) AVDDx_18 are within ±0.05 V of each other. SVDD and SPK_V are within ±0.05 V of each other.

- (2) For optimal performance with $CM=0.9V$, min $AVDDx_18 = 1.8V$.
(3) Minimum voltage for $HVDD_18$ and $RECVDD_33$ should be grea
- (3) Minimum voltage for HVDD_18 and RECVDD_33 should be greater than or equal to AVDDx_18.
- (4) At DVDD_18 values lower than 1.65V, the PLL and SAR ADC do not function.

Recommended Operating Conditions (continued)

(5) The PLL Input Frequency refers to clock frequency after PLL_CLKIN_DIV divider. Frequencies higher than 20MHz can be sent as an input to this PLL_CLKIN_DIV and reduced in frequency prior to input to the PLL.

7.4 Thermal Information

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](http://www.ti.com/lit/pdf/spra953). (2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as

specified in JESD51-7, in an environment described in JESD51-2a. (3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-

standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter, ψ_{JT} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining $θ_{JA}$, using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter, ψ_{JB} , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining θ_{JA} , using a procedure described in JESD51-2a (sections 6 and 7).

(7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

7.5 Electrical Characteristics, SAR ADC

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(1) When VBAT is not being sampled/converted. When VBAT is being sampled, effective input impedance to GND is 5.24kΩ.

(2) When utilizing External SAR reference, this external reference should be restricted V_{EXT} SAR REF^{≤AVDD_18} and AVDD2_18.

7.6 Electrical Characteristics, ADC

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

Electrical Characteristics, ADC (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(1) Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20-Hz to 20-kHz bandwidth using an audio analyzer.

(2) All performance measurements done with pre-analyzer 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

Electrical Characteristics, ADC (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(3) Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20-Hz to 20-kHz bandwidth using an audio analyzer.

(4) All performance measurements done with pre-analyzer 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

Electrical Characteristics, ADC (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

7.7 Electrical Characteristics, Bypass Outputs

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(1) All performance measurements done with 20-kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

Electrical Characteristics, Bypass Outputs (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(2) All performance measurements done with 20-kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

7.8 Electrical Characteristics, Microphone Interface

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(1) With Common Mode voltage of 0.9V, the MICBIAS_VDD voltage must be at minimum 3.05V to utilize Micbias Mode 4, and minimum of 3.2V to utilize Micbias Mode 5. With Common Mode voltage of 0.75V, the corresponding MICBIAS_VDD voltage must be minimum of 2.65V and 2.75V respectively.

Electrical Characteristics, Microphone Interface (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(2) PSRR is specified as 20*log₁₀(V_{sup}/V_{out}), where Vsup is the signal applied on the power supply and V_{out} is the measured analog output. For ADC, V_{out} is given as equivalent analog input signal which produces the same level of digital output signal.

7.9 Electrical Characteristics, Audio DAC Outputs

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

⁽¹⁾ Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20Hz to 20kHz bandwidth using an audio analyzer.

⁽²⁾ All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

Electrical Characteristics, Audio DAC Outputs (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
PSRR	Power Supply Rejection Ratio ⁽³⁾	217Hz, 100mVpp signal applied to AVDD_18, AVDD _x 18. HVDD_18,CPVDD_18,IOVDDx_33,DVDD_18		74		dB
		1kHz, 100mVpp signal applied to AVDD_18, AVDDx_18,HVDD_18,CPVDD_18,IOVDDx_33,DVD D_{18}		73		dB
$R_{\rm L}$	Output Resistive Load ⁽⁴⁾	Resistance to ground	9	10		$k\Omega$
AUDIO DAC - STEREO DIFFERENTIAL LINE OUTPUT						
	Device Setup	Load = $10k\Omega$ (differential), 56pF between RECP, RECMM and LOL, LOR Left DAC routed directly to RECP, RECM, Right DAC routed to LOL, LOR Output CM=0.9V DOSR = 128, MCLK=256* f_s Channel Gain = 0dB Processing Block = PRB P1, Power Tune = PTM P4 RECVDD_33, AVDD_18, AVDDx_18=1.8V				
	Full scale output voltage (0dBr)			$\mathbf{1}$		
		CM = 0.75V, AVDD_18, AVDDx_18, RECVDD_33=1.5V		0.75		V_{RMS}
SNR	Signal-to-noise ratio A-weighted ^{(1) (2)}	All zeros fed to DAC input	92	105		
		All zeros fed to DAC input, CM=0.75V, AVDD_18, AVDDx_18, RECVDD_33=1.5V		103		dB
DR	Dynamic range, A-weighted (5) (6)	-60dBFS input signal		105		dB
		-60dBFS input full-scale signal, CM=0.75V, AVDD_18, AVDDx_18, RECVDD_33=1.5V		102		
THD+N	Total Harmonic Distortion plus Noise	-3dBFS input signal		-94	-75	dB
		-3dBFS input signal, CM=0.75V, AVDD 18, AVDDx 18, RECVDD 33=1.5V		-92		
	DAC Gain Error	-3dBFS input signal		±0.03		dВ
		-3dBFS input signal, CM=0.75V, AVDD_18, AVDDx_18, RECVDD_33=1.5V		±0.04		
	Inter Channel Gain Mismatch	-3dBFS input signal		0.0		dB
	Output Offset	All zero's fed to DAC Input		±0.5		
		All zero's fed to DAC Input, CM=0.75V AVDD_18, AVDDx_18, RECVDD_33=1.5V		±0.4		mV
	DAC Mute Attenuation	Digital Volume Control is Muted		126		dB
PSRR	Power Supply Rejection Ratio ⁽⁷⁾	217kHz, 100mVpp signal applied to AVDD 18, AVDDx_18, RECVDD_33, HVDD_18, CPVDD_18, IOVDDx_33, DVDD_18		93		dB
		1kHz, 100mVpp signal applied to AVDD 18, AVDDx_18, RECVDD_33, HVDD_18, CPVDD_18, IOVDDx_33, DVDD_18		92		dB

⁽³⁾ PSRR is specified as 20*log₁₀($V_{\text{sup}}/V_{\text{out}}$), where Vsup is the signal applied on the power supply and V_{out} is the measured analog output. For ADC, V_{out} is given as equivalent analog input signal which produces the same level of digital output signal.

⁽⁴⁾ Minimum resistive load and Maximum capacitive load are specified by design.

⁽⁵⁾ Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20Hz to 20kHz bandwidth using an audio analyzer.

⁽⁶⁾ All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

⁽⁷⁾ PSRR is specified as 20*log₁₀(V_{sup}/V_{out}), where Vsup is the signal applied on the power supply and V_{out} is the measured analog output. For ADC, V_{out} is given as equivalent analog input signal which produces the same level of digital output signal.

Electrical Characteristics, Audio DAC Outputs (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(8) Minimum resistive load and Maximum capacitive load are specified by design.

⁽⁹⁾ Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20Hz to 20kHz bandwidth using an audio analyzer.

⁽¹⁰⁾ All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

⁽¹¹⁾ PSRR is specified as 20*log₁₀(V_{sup}/V_{out}), where Vsup is the signal applied on the power supply and V_{out} is the measured analog output. For ADC, V_{out} is given as equivalent analog input signal which produces the same level of digital output signal.

Electrical Characteristics, Audio DAC Outputs (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(12) Minimum resistive load and Maximum capacitive load are specified by design.

(13) Ratio of output level with 1kHz full-scale sine wave input, to the output level with the inputs short circuited, measured A-weighted over a 20Hz to 20kHz bandwidth using an audio analyzer.

(14) All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

7.10 Electrical Characteristics, Class-D Outputs

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(1) All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter may result in higher THD+N and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, may affect dynamic specification values.

Electrical Characteristics, Class-D Outputs (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V , MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

(2) PSRR is specified as 20*log₁₀(V_{sup}/V_{out}), where Vsup is the signal applied on the power supply and V_{out} is the measured analog output. For ADC, V_{out} is given as equivalent analog input signal which produces the same level of digital output signal.

(3) Over Temperature and Over Current Protection parameters are indicative values from design. Over Temperature trip point can be very heavily influenced by thermal properties of the actual PCB.

(4) Over Temperature and Over Current Protection parameters are indicative values from design. Over Temperature trip point can be very heavily influenced by thermal properties of the actual PCB.

7.11 Electrical Characteristics, Misc.

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

Electrical Characteristics, Misc. (continued)

 ${\sf T_A}$ = 25°C; <code>AVDD_18,</code> <code>AVDD ${\sf x}$ _18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; <code>RECVDD_33</code> = 3.3V; SVDD,</code> $\mathsf{SPK_V},\ \mathsf{MICBIAS_VDD} = 3.6 \lor; \ \mathsf{f}_\mathsf{S}\ (\mathsf{Audio}) = 48\mathsf{kHz};\ \mathsf{Audio\ Word\ Length} = 20\ \mathsf{bits};\ \mathsf{C}_\mathsf{ext} = 1\mathsf{\mu}\mathsf{F}\ \mathsf{on}\ \mathsf{VREF_SAR}\ \mathsf{and}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\$ VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

Electrical Characteristics, Misc. (continued)

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, $\mathsf{SPK_V},\ \mathsf{MICBIAS_VDD} = 3.6 \lor; \ \mathsf{f}_\mathsf{S}\ (\mathsf{Audio}) = 48\mathsf{kHz};\ \mathsf{Audio\ Word\ Length} = 20\ \mathsf{bits};\ \mathsf{C}_\mathsf{ext} = 1\mathsf{\mu}\mathsf{F}\ \mathsf{on}\ \mathsf{VREF_SAR}\ \mathsf{and}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\$ VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

7.12 Electrical Characteristics, Logic Levels, IOVDDx

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, $\mathsf{SPK_V},\ \mathsf{MICBIAS_VDD} = 3.6 \lor; \ \mathsf{f}_\mathsf{S}\ (\mathsf{Audio}) = 48\mathsf{kHz};\ \mathsf{Audio\ Word\ Length} = 20\ \mathsf{bits};\ \mathsf{C}_\mathsf{ext} = 1\mathsf{\mu}\mathsf{F}\ \mathsf{on}\ \mathsf{VREF_SAR}\ \mathsf{and}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\mathsf{ext} = 12\mathsf{R}\ \mathsf{in}\ \mathsf{C}_\$ VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

7.13 Interface Timing

Note: All timing specifications are measured at characterization but not tested at final test. The audio serial interface timing specifications are applied to Audio Serial Interface 1, Audio Serial Interface 2 and Audio Serial Interface 3.

7.13.1 Typical Timing Characteristics — Audio Data Serial Interface Timing (I²S)

WCLK represents WCLK1 terminal for Audio Serial Interface 1, WCLK2 terminal for Audio Serial Interface 2 and Word Clock for Audio Serial Interface 3. **BCLK** represents BCLK1 terminal for Audio Serial Interface 1, BCLK2 for Audio Serial Interface 2 and Bit Clock for Audio Serial Interface 3. **DOUT** represents DOUT1 terminal for Audio Serial Interface 1, DOUT2 for Audio Serial Interface 2 and Data Out for Audio Serial Interface 3. **DIN** represents DIN1 terminal for Audio Serial Interface 1, DIN2 for Audio Serial Interface 2 and Data In for Audio Serial Interface 3. Specifications are at 25° C with DVDD_18 = 1.8V.

Figure 2. I ²S,LJF,RJF Timing in Master Mode

Table 1. I ²S,LJF,RJF Timing in Master Mode (see [Figure](#page-21-1) 2)

Table 2. I ²S,LJF,RJF Timing in Slave Mode (see [Figure](#page-22-0) 3)

7.13.2 Typical DSP Timing Characteristics

Specifications are at 25° C with DVDD_18 = 1.8 V.

Figure 4. DSP,Mono PCM Timing in Master Mode

Figure 5. DSP/Mono PCM Timing in Slave Mode

Table 4. DSP,Mono PCM Timing in Slave Mode (see [Figure](#page-23-1) 5)

7.13.3 I ²C Interface Timing

Figure 6. I ²C Interface Timing Diagram

Table 5. I ²C Interface Timing (see [Figure](#page-24-0) 6)

7.13.4 SPI Interface Timing

SS = SCL terminal, **SCLK** = I2C_ADDR_SCLK terminal, **MISO** = GPO1 terminal, and **MOSI** = SDA terminal. Specifications are at 25° C with DVDD_18 = 1.8V.

Figure 7. SPI Interface Timing Diagram

Table 6. SPI Interface Timing

7.14 Typical Characteristics

 $T_A = 25^{\circ}$ C; AVDD_18, AVDDx_18, HVDD_18, CPVDD_18, DVDD_18, IOVDDx_33 = 1.8V; RECVDD_33 = 3.3V; SVDD, SPK_V, MICBIAS_VDD = 3.6V; f_S (Audio) = 48kHz; Audio Word Length = 20 bits; C_{ext} = 1µF on VREF_SAR and VREF_AUDIO terminals; PLL disabled, Device CM = 0.9V unless otherwise noted.

Typical Characteristics (continued) 7.14.1 SAR ADC Performance

Typical Characteristics (continued) 7.14.2 Audio ADC Performance

Typical Characteristics (continued) 7.14.3 MICBIAS Performance

7.14.4 Audio DAC to Line Out Performance

Typical Characteristics (continued)

Typical Characteristics (continued) 7.14.5 Audio DAC to Headphone Performance

Typical Characteristics (continued) 7.14.6 Audio DAC to Receiver Performance

Typical Characteristics (continued) 7.14.7 Class-D Driver Performance

8 Detailed Description

8.1 Overview

The TLV320AIC3268 (AIC3268) is a flexible, highly-integrated, low-power, low-voltage stereo audio codec. The AIC3268 features four digital microphone inputs, plus programmable outputs, PowerTune capabilities, enhanced fully-programmable miniDSP, predefined and parameterizable signal processing blocks, integrated PLL, and flexible digital audio interfaces. Extensive register-based control of power, input and output channel configuration, gains, effects, terminal-multiplexing and clocks are included, allowing the device to be precisely targeted to its application.

The record path of the TLV320AIC3268 covers operations from 8kHz mono to 192kHz stereo recording, and contains programmable input channel configurations which cover single-ended and differential setups, as well as floating or mixing input signals. It also provides a digitally-controlled stereo microphone preamplifier and integrated microphone bias. The record path can also be configured for up to two stereo (that is up to 4) simultaneous digital microphone Pulse Density Modulation (PDM) interfaces typically used at 64Fs or 128Fs.

The playback path offers signal processing blocks for filtering and effects; headphone, line, receiver, and Class-D speaker output; flexible mixing of DAC; and analog input signals as well as programmable volume controls. The playback path contains two high-power DirectPath™ headphone output drivers which eliminate the need for ac coupling capacitors. A built in charge pump generates the negative supply for the ground centered headphone drivers. These headphone output drivers can be configured in multiple ways, including stereo, and mono BTL. The playback path also features stereo lineout drivers, which can be configured either for single-ended operation or differential output operation for high performance systems. In addition, playback audio can be routed to an integrated Class-D speaker driver or a differential receiver amplifier.

The integrated PowerTune technology allows the device to be tuned to just the right power-performance tradeoff. Mobile applications frequently have multiple use cases requiring very low-power operation while being used in a mobile environment. When used in a docked environment power consumption typically is less of a concern while lowest possible noise is important. With PowerTune the TLV320AIC3268 can address both cases.

The required internal clock of the TLV320AIC3268 can be derived from multiple sources, including the MCLK terminal, the BCLK1 terminal, the BCLK2 terminal, several general purpose IO terminals or the output of the internal PLL, where the input to the PLL again can be derived from similar terminals. Although using the internal fractional PLL ensures the availability of a suitable clock signal, it is not recommended for the lowest power settings. The PLL is highly programmable and can accept available input clocks in the range of 512kHz to 50MHz. To enable even lower clock frequencies, an integrated low-frequency clock multiplier can also be used as an input to the PLL.

The TLV320AIC3268 has a 12-bit SAR ADC converter that supports system voltage measurements. These system voltage measurements can be sourced from three dedicated analog inputs (IN1L/AUX1, IN1R/AUX2 or VBAT terminals), or, alternatively, an on-chip temperature sensor that can be read by the SAR ADC.

The device also features three full Digital Audio Serial Interfaces, each supporting I2S, DSP/TDM, RJF, LJF, and mono PCM formats. This enables three simultaneous digital playback and record paths to three independent digital audio buses or chips. Each of the Digital Audio Serial Interfaces can be run using separate power voltages to enable easy integration with separate chips with different IO voltages.

The device is available in the 9mm x 9mm x 0,9mm 64-terminal QFN package.

8.2 Functional Block Diagram

8.3 Feature Description

8.3.1 Analog Audio I/O

Figure 37. Analog Routing Diagram

8.3.1.1 Analog Low Power Bypass

The TLV320AIC3268 offers two analog-bypass modes. In either of the modes, an analog input signal can be routed from an analog input terminal to an amplifier driving an analog output terminal. Neither the ADC nor the DAC resources are required for such operation; this supports low-power operation during analog-bypass mode. In analog low-power bypass mode, line-level signals can be routed directly from the analog inputs IN1L to the left lineout amplifier (LOL) and IN1R to LOR. Additionally, line-level signals can be routed directly from these analog inputs to the amplifier.

8.3.1.2 Headphone Outputs

The stereo headphone drivers on terminals HPL and HPR can drive loads with impedances down to 16Ω in single-ended DC-coupled headphone configurations. An integral charge pump generates the negative supply required to operate the headphone drivers in dc-coupled mode, where the common mode of the output signal is made equal to the ground of the headphone load using a ground-sense circuit. Operation of headphone drivers in dc-coupled (ground centered mode) eliminates the need for large dc-blocking capacitors.

Figure 38. TLV320AIC3268 Ground-Centered Headphone Output

8.3.1.2.1 Using the Headphone Amplifier

The headphone drivers are capable of driving a mixed combination of DAC signal, left and right ADC PGA signal, and LOL and LOR output signals by configuring B0_P1_R27-R29. The ADC PGA signals can be attenuated up to 36dB before routing to headphone drivers by configuring B0_P1_R18 and B0_P1_R19. The line-output signals can be attenuated up to 78dB before routing to headphone drivers by configuring B0_P1_R28 and B0_P1_R29. The level of the DAC signal can be controlled using the digital volume control of the DAC by configuring B0_P0_R64-R66. To control the output-voltage swing of headphone drivers, the headphone driver volume control provides a range of –6.0dB to +14.0dB in steps of 1dB. These can be configured by programming B0_P1_R27, B0_P1_R31, and B0_P1_R32. In addition, finer volume controls are also available when routing LOL or LOR to the headphone drivers by controlling B0_P1_R27-R28. These level controls are not meant to be used as dynamic volume control, but more to set output levels during initial device configuration. Register B0_P1_R9_D[6:5] allows the headphone output stage to be scaled to tradeoff power delivered vs quiescent power consumption. (1)

8.3.1.2.2 Ground-Centered Headphone Amplifier Configuration

Among the other advantages of the ground-centered connection is inherent freedom from turn-on transients that can cause audible pops, sometimes at uncomfortable volumes.

⁽¹⁾ If the device must be placed into 'mute' from the –6.0dB setting, set the device at a gain of –5.0dB first, then place the device into mute.

8.3.1.2.2.1 Circuit Topology

The power supply hook up scheme for the ground centered configuration is shown in HVDD_18 terminal supplies the positive side of the headphone amplifier. CPVDD 18 terminal supplies the charge pump which in turn supplies the negative side of the headphone amplifier. Two capacitors are required for the charge pump circuit to work. These capacitors should be X7R rated.

Figure 39. Ground-Centered Headphone Connections

8.3.1.2.2.2 Charge Pump Setup and Operation

The built in charge pump draws charge from the CPVDD 18 supply, and by switching the external capacitor between CPFCP and CPFCM, generates the negative voltage on VNEG terminal. The charge-pump circuit uses the principles of switched-capacitor charge conservation to generate the VNEG supply in a very efficient fashion.

To turn on the charge pump circuit when headphone drivers are powered, program B0_P1_R35_D[1:0] to "'00". When the charge pump circuit is disabled, VNEG acts as a ground terminal, allowing unipolar configuration of the headphone amps. By default, the charge pump is disabled. The switching rate of the charge pump can be controlled by B0_P1_R33. Because the charge pump can demand significant inrush currents from the supply, it is important to have a capacitor connected in close proximity to the CPVDD_18 and CPVSS terminals of the device. At 500kHz clock rate this requires approximately a 10μF capacitor. The ESR and ESL of the capacitor must be low to allow fast switching currents.

The ground-centered mode of operation is enabled by configuring B0_P1_R31_D7 to "1". Note that the HPL and HPR gain settings are ganged in Ground-Cetered Mode of operation (B0_P1_R32_D7 = "1"). The HPL and HPR gain settings cannot be ganged if using the Stereo Unipolar Configuration.

8.3.1.2.2.3 Output Power Optimization

The device can be optimized for a specific output-power range. The charge pump and the headphone driver circuitry can be reduced in power so less overall power is consumed. The headphone driver power can be programmed in B0_P1_R9. The control of charge pump switching current is programmed in B0_P1_R34_D[4:2].

8.3.1.2.2.4 Offset Correction and Start-Up

The TLV320AIC3268 offers an offset-correction scheme that is based on calibration during power up. This scheme minimizes the differences in DC voltage between HPVSS_SENSE and HPL/HPR outputs.

The offset calibration happens after the headphones are powered up in ground-centered configuration. All other headphone configurations like signal routings, gain settings and mute removal must be configured before headphone powerup. Any change in these settings while the headphones are powered up may result in additional offsets and are best avoided.

The offset-calibration block has a few programmable parameters that the user must control. The user can either choose to calibrate the offset only for the selected input routing or all input configurations. The calibration data is stored in internal memory until the next hardware reset or until AVDDx power is removed.

Programming B0 P1 R34 D[1:0] as "10" causes the offset to be calibrated for the selected input mode. Programming B0_P1_R34_D[1:0] as "11" causes the offset to be calibrated for all possible configurations. All related blocks must be powered while doing offset correction.

Programming B0_P1_R34_D[1:0] as "00" (default) disables the offset correction block. While the offset is being calibrated, no signal should be applied to the headphone amplifier, that is the DAC should be kept muted and analog bypass routing should be kept at the highest attenuation.

8.3.1.3 Stereo Line Outputs

The TLV320AIC3268 features stereo line level drivers which can be configured in either fully differential configuration (RECP, RECM and LOL, LOR) or single-ended configuration (LOL and LOR).The stereo line level drivers can drive a wide range of line level resistive impedances in the range of 600Ω to 10kΩ. The output common mode of line level drivers can be configured to equal the analog input common-mode setting, either 0.75V or 0.9V. The line-level drivers can drive out a mixed combination of DAC signal and attenuated ADC PGA signal, and signal mixing is register-programmable.

8.3.1.3.1 Line Out Amplifier Configurations

The TLV320AIC3268 can support stereo differential lineout as shown in [Figure](#page-38-0) 40 for stereo DAC playback.

Figure 40. Stereo Differential Line-out

In this mode, the output common-mode setting of the receiver amplifier should be kept the same as the input common mode by programming B0_P1_R8_D[1:0]="00". Also, in this mode the receiver driver gain should be kept at 0dB by programming B0_P1_R40_D[5:0] and B0_P1_R41_D[5:0] as "000000". The RECVDD_33 supply should be connected to the AVDDx_18 power rails for symmetricity.

The TLV320AIC3268 can also support stereo single-ended line outputs as shown in [Figure](#page-39-0) 41 for stereo DAC playback.

Signal mixing can be configured by programming B0_P1_R22 and B0_P1_R23. To route the output of Left DAC and Right DAC for stereo single-ended output, as shown in [Figure](#page-39-0) 41, LDACM can be routed to LOL driver by setting $B0_P1_R22_D7 = '1'$, and RDACM can be routed to LOR driver by setting $B0_P1_R22_D6 = '1'$. Alternatively, stereo single-ended signals can also be routed through the mixer amplifiers by configuring B0_P1_R23_D[7:6]. For lowest-power operation, stereo single-ended signals can also be routed in direct terminal bypass with possible gains of 0dB, -6dB, or -12dB by configuring B0_P1_R23_D[4:3] and B0_P1_R23_D[1:0]. While each of these two bypass cases could be used in a stereo single-ended configuration, a mono differential input signal could also be used.

The output of the stereo line out drivers can also be routed to the stereo headphone drivers, with 0dB to -72dB gain controls in steps of 0.5dB on each headphone channel. This enables the DAC output or bypass signals to be simultaneously played back to the stereo headphone drivers as well as stereo line- level drivers. This routing and volume control is achieved in B0_P1_R28 and B0_P1_R29.

Figure 41. Stereo Single-Ended Line-out

Additionally, the two line-level drivers can be configured to act as a mono differential line level driver by routing the output of LOL to LOR (B0_P1_R22_D2 = '1'). This differential signal takes either LDACM, MAL, or IN1L-B as a single-ended mono signal and creates a differential mono output signal on LOL and LOR.

Figure 42. Single Channel Input to Differential Line-out

8.3.1.4 Differential Receiver Output

The differential receiver amplifier output spans the RECP and RECM terminals and can drive a 32Ω receiver driver. he receiver driver can drive up to a 1Vrms output signal.

The differential receiver driver is capable of driving a mixed combination of DAC signal through the Line Out amplifiers and the line-bypass from analog input IN1L and IN1R. Routing and volume level setting of the IN1L and IN1R input signals to the Positive and Negative driver is controlled by B0_P1_R38 and B0_P1_R39 respectively. These two registers enable fine tuning of the inputs to the receiver driver by allowing up to 78dB of attenuation. A single volume control can be utilized for both inputs by setting B0_P1_R39_D7. Routing and volume level setting of the LOL and LOR signals to the positive and negative inputs of the differential receiver driver is controlled by B0_P1_R36 and B0_P1_R37 respectively. These two registers enable fine tuning of the separate positive and negative differential signals by allowing up to 78dB of attenuation. A single volume control can be utilized for both inputs by setting B0_P1_R37_D7. Routing of LDACP and LDACM signals to the Positive and Negative driver is controlled by B0_P1_R42_D6 and B0_P1_R42_D5 respectively.

RECM RECP 32Ω LOL IN1L LOR IN1R LDA_C **LDACM**

Figure 43. Receiver Differential Output

The receiver driver can be powered on by writing 11 to B0_P1_R40_D[7:6]. The positive driver gain and muting can be controlled by writing to B0_P1_R40_D[5:0], and the negative driver gain can be controlled by writing to B0_P1_R41_D[5:0], with each amplifier providing -6dB to 29dB gains in steps of 1dB. A single volume control can be utilized for the differential receiver output drivers by setting B0_P1_R41_D7 to '1'.

The TLV320AIC3268 has an overcurrent/short-circuit protection feature for the receiver drivers that is always enabled to provide protection. If the output is shorted, this overcurrent condition either shuts down the output stage (if $B0_P1_R10_D0 = 1$) or starts to limit the amount of current (if $B0_P1_R10_D0 = 0$). The default condition for the receiver driver is current-limiting mode. In case of a short circuit, for automatic latching shutdown, the output is disabled and a status flag is provided as read-only bits B0_P0_R44_D7 for RECP and on B0_P0_R44_D6 for RECM.

The receiver driver also has an offset calibration for minimizing start-up transients. By default, this feature is enabled at every power-up by setting B0_P1_R42_D[4:3] to '01'. The status of the offset calibration can be read through the Receiver Offset Calibration Flag (B0_P1_R42_D7). Offset calibration should only be disabled if this driver is utilized as a second single-ended headphone configuration (that is, should not be utilized in differential configuration).

8.3.1.5 Class-D Speaker Outputs

The integrated Class-D speaker driver (SPKP, SPKN) is capable of driving an 8Ω differential load. The speaker driver can be powered directly from the power supply (2.7V to 5.5V) on the SVDD terminal, however the voltage (including spike voltage) must be limited below the Absolute Maximum Voltage of 6.0V.

The speaker driver capable of supplying 0.74 W at 10% THD+N with a 3.6-V power supply and 1.45 W at 10% THD+N with a 5.0V power supply. Separate left and right channels can be sent to the Class-D driver through the Lineout signal path, or from the mixer amplifiers in the ADC bypass. Additionally, the analog mixer before the Speaker amplifier can sum the left and right audio signals for monophonic playback.

The speaker driver is capable of driving a mixed combination of DAC signal through the Line Out amplifiers and the left and right ADC PGA signal. The ADC PGA signals can be routed to the speaker drivers by setting B0_P1_R45_D7 (Left Mixer amplifier to Speaker) and B0_P1_R45_D6 (Right Mixer amplifier to Speaker), and these signals can be attenuated up to 36dB before this routing to the speakers by configuring B0_P1_R18 and B0_P1_R19. Routing and volume level setting of the LOL and LOR signals to the speaker driver is controlled by B0_P1_R46 and B0_P1_R47 respectively. These two registers enable fine tuning of the separate stereo signals by allowing up to 78dB of attenuation. To play the stereo DAC signals through the Line Out amplifiers to the speaker, the DAC signals should be routed to the LOL/LOR drivers by setting B0_P1_R22_D[7:6]. The level of these DAC signal can also be controlled using the digital volume control of the DAC signal (B0_P0_R65 and B0 P0 R66).

Figure 44. Speaker Output

The class-D speaker driver can be powered on by writing to B0_P1_R45_D1. The driver gain can be controlled by writing to B0_P1_R48_D[6:4], and it can be muted by writing '000' to these bits.

The TLV320AIC3268 has a short-circuit protection feature for the speaker driver that is always enabled to provide protection. If the output is shorted, the output stage shuts down on the overcurrent condition. (Current limiting is not an available option for the higher-current speaker driver output stage.) In case of a short circuit, the output is disabled and a status flag is provided as a read-only bit on B0_P0_R44_D7. If shutdown occurs due to an overcurrent condition, then the device requires a reset to re-enable the output stage. Resetting can be done in two ways. First, the device master reset can be used, which requires either toggling the RESET terminal or using the software reset. If master reset is used, it resets all of the registers. Second, a dedicated speaker power-stage reset can be used that keeps all of the other device settings. The speaker power-stage reset is done by setting B0_P1_R45_D1. If the fault condition has been removed, then the device returns to normal operation. If the fault is still present, then another shutdown occurs. Repeated resetting (more than three times) is not recommended, as this could lead to overheating.

The TLV320AIC3268 has an overtemperature thermal-protection (OTP) feature for the speaker driver which is always enabled to provide protection. If the device is overheated, then the output stops switching. When the device cools down, the device resumes switching. An overtemperature status flag is provided as a read-only bit on B0_P0_R45_D7, and this status flag can be routed to INT1 interrupt (B0_P0_R48_D1 = '1') or INT2 interrupt (B0_P0_R49_D1 = '1'). The OTP feature is for self-protection of the device. If die temperature can be controlled at the system/board level, then overtemperature does not occur.

To minimize battery current leakage, the SVDD voltage levels should not be less than the AVDDx_18 voltage levels.

8.3.2 ADC / Digital Microphone Interface

The TLV320AIC3268 includes a stereo audio ADC, which uses a delta-sigma modulator with a programmable oversampling ratio, followed by a digital decimation filter and a programmable miniDSP. The ADC supports sampling rates from 8kHz to 192kHz. In order to provide optimal system power management, the stereo recording path can be powered up one channel at a time, to support the case where only mono record capability is required.

The ADC path of the TLV320AIC3268 features a large set of options for signal conditioning as well as signal routing:

- 2 ADCs
- 8 analog inputs which can be mixed or multiplexed in single-ended or differential configuration
- 2 programmable gain amplifiers (PGA) with a range of 0 to +47.5dB
- 2 mixer amplifiers for analog bypass
- 2 low power analog bypass channels
- Fine gain adjust of digital channels with 0.1 dB step size
- Digital volume control with a range of -12 to +20dB
- Mute function

- Automatic gain control (AGC)
- In addition to the standard set of ADC features the TLV320AIC3268 also offers the following special functions:
- Built in microphone biases

NSTRUMENTS

- Four-channel digital microphone interface
	- Allows 4 total microphones
	- Up to 4 digital microphones
	- Up to 2 analog microphones
- Channel-to-channel phase adjustment
- Fast charge of ac-coupling capacitors
- Anti thump
- Adaptive coefficient update mode

8.3.2.1 ADC Signal Routing

As shown in [Figure](#page-35-0) 37, the TLV320AIC3268 includes eight analog inputs which can be configured as either 4 stereo single-ended pairs or 4 fully-differential pairs. These terminals connect through series resistors and switches to the virtual ground terminals of two fully-differential amplifiers (one per ADC/PGA channel). By turning on only one set of switches per amplifier at a time, the inputs can be effectively multiplexed to each ADC PGA channel. By turning on multiple sets of switches per amplifier at a time, audio sources can be mixed. The TLV320AIC3268 supports the ability to mix up to five single-ended analog inputs or up to three fully-differential analog inputs into each ADC PGA channel.

In most applications, high input impedance is desired for analog inputs. However when used in conjunction with high gain as in the case of microphone inputs, the higher input impedance results in higher noise or lower dynamic range. The TLV320AIC3268 allows the user the flexibility of choosing the input impedance from 10kΩ, 20kΩ and 40kΩ. When multiple inputs are mixed together, by choosing different input impedances, level adjustment can be achieved. For example, if one input is selected with 10k Ω input impedance and the second input is selected with 20kΩ input impedance, then the second input is attenuated by half as compared to the first input. Note that this input level control is not intended to be a volume control, but instead used occasionally for level setting. Also, note that this input-level configurability is available on IN1L, IN1R, IN2L, IN2R, IN3L, and IN3R; for IN4L and IN4R, this input impedance is fixed at 20kΩ.

Mixing of multiple inputs can easily lead to PGA outputs that exceed the range of the internal amplifiers, resulting in saturation and clipterminalg of the mixed output signal. Whenever mixing is being implemented, the system designer is advised to take adequate precautions to avoid such a saturation from occurring. In general, the mixed signal should not exceed 0dB.

Typically, voice or audio signal inputs are capacitively coupled to the device. This allows the device to independently set the common mode of the input signals to values chosen by register control of B0_P1_R8_D2 to either 0.9V or 0.75V. The correct value maximizes the dynamic range across the entire analog supply range. Failure to capacitively connect the input to the device can cause high offset due to mismatch in source commonmode and device common-mode setting. In extreme cases it could also saturate the analog channel, causing distortion.

8.3.2.1.1 High Impedance Input Mode

The TLV320AIC3268 supports a special high impedance input mode on terminals IN1L/AUX1 and IN1R/AUX2 to enable interfacing with sensors having high output impedance. By programming B0_P1_R17_D[5] and B0_P1_R17_D[4] to '1' IN1L/AUX1 and IN1R/AUX2 can be used in high impedance mode respectively. While using this mode, IN1L/AUX1 and IN1R/AUX2 should not be routed to Left and Right ADC PGA's or B0_P1_R52_D[7:6], B0_P1_R52_D[1:0], B0_P1_R55_D[7:6] and B0_P1_R57_D[5:4] should be programmed as "00". While using this mode the signal should be externally biased around a common mode which is close to the device common mode programmed via B0_P1_R8_D[2]. When using high impedance mode, routing of MAL and MAR amplifiers to Speaker Amplifier and Lineout Drivers is not supported.

8.3.2.2 ADC Gain Setting

When the gain of the ADC Channel is kept at 0dB and the common mode set to 0.75V, a single-ended input of $0.375V_{RMS}$ results in a full-scale digital signal at the output of ADC channel. Similarly, when the gain is kept at 0dB, and common mode is set to 0.9V, a single-ended input of $0.5V_{RMS}$ results in a full-scale digital signal at the output of the ADC channel. However various block functions control the gain through the channel. The gain applied by the PGA is described in [Table](#page-43-0) 7. Additionally, the digital volume control adjusts the gain through the channel as described in Digital [Volume](#page-43-1) Control. A finer level of gain adjustment is possible and described in [Fine](#page-44-0) Digital Gain [Adjustment](#page-44-0). The decimation filters A, B and C along with the delta-sigma modulator contribute to a DC gain of 1.0 through the channel.

8.3.2.2.1 Analog Programmable Gain Amplifier (PGA)

The TLV320AIC3268 features a built-in low-noise PGA for boosting low-level signals, such as direct microphone inputs, to full-scale to achieve high SNR. This PGA can provide a gain in the range of 0dB to 47.5dB for singleended inputs or 6dB to 53.5dB for fully-differential inputs (gain calculated w.r.t. input impedance setting of 10kΩ, 20kΩ input impedance will result in 6dB lower and 40kΩ will result in 12dB lower gain). This gain can be user controlled by writing to B0_P1_R59 and B0_P1_R60. In the AGC mode this gain can also be automatically controlled by the built-in hardware AGC.

Book 0, Page 1, Register 59, D[6:0] $(BO_P1_R59_D[6:0])$ Book 0, Page 1, Register 60, D[6:0] B0_P1_R60_D[6:0]	EFFECTIVE GAIN APPLIED BY PGA						
	SINGLE-ENDED			DIFFERENTIAL			
	$R_{IN} = 10k\Omega$	$R_{IN} = 20k\Omega$	$R_{\text{IN}} = 40k\Omega$	$R_{IN} = 10k\Omega$	$R_{IN} = 20k\Omega$	$R_{IN} = 40k\Omega$	
000 0000	0 dB	-6 dB	-12 dB	6.0 dB	0 dB	-6.0 dB	
000 0001	0.5 dB	-5.5 dB	-11.5 dB	6.5 dB	0.5 dB	-5.5 dB	
000 0010	1.0 dB	-5.0 dB	-11.0 dB	7.0 dB	7.5 dB	-5.0 dB	
\cdots	\cdots	\cdots	\cdots	\cdots	\cdots	\cdots	
101 1110	47.0 dB	41.0 dB	35.0 dB	53.0 dB	47.0 dB	41.0 dB	
101 1111	47.5 dB	41.5 dB	35.5 dB	53.5 dB	47.5 dB	41.5 dB	

Table 7. Analog PGA vs Input Configuration

The gain changes are implemented with an internal soft-stepterminalg algorithm that only changes the actual volume level by one 0.5-dB step every one or two ADC output samples, depending on the register value (see registers B0_P0_R81_D[1:0]). This soft-stepterminalg ensures that volume control changes occur smoothly with no audible artifacts. On reset, the PGA gain defaults to a mute condition, and at power down, the PGA soft-steps the volume to mute before shutting down. A read-only flag B0_P0_R36_D7 and B0_P0_R36_D3 is set whenever the gain applied by the PGA equals the desired value set by the register. The soft-stepterminalg control can also be disabled by programming B0_P0_R81_D[1:0].

8.3.2.2.2 Digital Volume Control

The TLV320AIC3268 also has a digital volume-control block with a range from -12dB to +20dB in steps of 0.5dB. It is set by programming B0_P0_R83 and B0_P0_R84 respectively for left and right channels.

+19.5 010 0111 +20.0 010 1000

Table 8. Digital Volume Control for ADC

During volume control changes, the soft-stepterminalg feature is used to avoid audible artifacts. The softstepterminalg rate can be set to either 1 or 2 gain steps per sample. Soft-stepterminalg can also be entirely disabled. This soft-stepterminalg is configured via B0_P0_R81_D[1:0], and is common to soft-stepterminalg control for the analog PGA. During power-down of an ADC channel, this volume control soft-steps down to -12.0dB before powering down. Due to the soft-stepterminalg control, soon after changing the volume control setting or powering down the ADC channel, the actual applied gain may be different from the one programmed through the control register. The TLV320AIC3268 gives feedback to the user, through read-only flags B0_P0_R36_D7 for Left Channel and B0_P0_R36_D3 for the right channel.

8.3.2.2.3 Fine Digital Gain Adjustment

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Additionally, the gains in each of the channels is finely adjustable in steps of 0.1dB. This is useful when trying to match the gain between channels. By programming B0_P0_R82, the gain can be adjusted from 0dB to -0.4dB in steps of 0.1dB. This feature, in combination with the regular digital volume control allows the gains through the left and right channels be matched in the range of -0.5dB to +0.5dB with a resolution of 0.1dB.

8.3.2.2.4 AGC

The TLV320AIC3268 includes Automatic Gain Control (AGC) for ADC recording. AGC can be used to maintain a nominally-constant output level when recording speech. As opposed to manually setting the PGA gain, in the AGC mode, the circuitry automatically adjusts the PGA gain as the input signal becomes overly loud or very weak, such as when a person speaking into a microphone moves closer or farther from the microphone. The AGC algorithm has several programmable parameters, including target gain, attack and decay time constants, noise threshold, and max PGA applicable, that allow the algorithm to be fine tuned for any particular application. The algorithm uses the absolute average of the signal (which is the average of the absolute value of the signal) as a measure of the nominal amplitude of the output signal. Since the gain can be changed at the sample interval time, the AGC algorithm operates at the ADC sample rate.

- 1. **Target Level** represents the nominal output level at which the AGC attempts to hold the ADC output signal level. The TLV320AIC3268 allows programming of eight different target levels, which can be programmed from –5.5 dB to –24 dB relative to a full-scale signal. Since the TLV320AIC3268 reacts to the signal absolute average and not to peak levels, it is recommended that the target level be set with enough margin to avoid clipterminalg at the occurrence of loud sounds.
- 2. **Attack Time** sets how quickly the AGC circuitry reduces the PGA gain when the output signal level exceeds the target level due to increase in input signal level. Wide range of attack time programmability is supported in terms of number of samples (that is, number of ADC sample frequency clock cycles).
- 3. **Decay Time** sets how quickly the PGA gain is increased when the output signal level falls below the target level due to reduction in input signal level. Wide range of decay time programmability is supported in terms of number of samples (that is, number of ADC sample frequency clock cycles).
- 4. **Gain Hysteresis** is the hysteresis applied to the required gain calculated by the AGC function while changing its mode of operation from attack to decay or vice-versa. For example, while attacking the input signal, if the current applied gain by the AGC is *x*dB, and suddenly because of input level going down, the new calculated required gain is *y*dB, then this gain is applied provided *y* is greater than *x* by the value set in Gain Hysteresis. This feature avoids the condition when the AGC function can fluctuate between a very narrow band of gains leading to audible artifacts. The Gain Hysteresis can be adjusted or disabled by the user.
- 5. **Noise threshold** sets the level below which if the input signal level falls, the AGC considers it as silence, and thus brings down the gain to 0 dB in steps of 0.5 dB every FS and sets the noise threshold flag. The gain stays at 0 dB unless the input speech signal average rises above the noise threshold setting. This ensures that noise is not 'gained up' in the absence of speech. Noise threshold level in the AGC algorithm is programmable from -30dB to -90 dB of full-scale. When AGC Noise Threshold is set to –70dB, –80db, or –90dB, the microphone input Max PGA applicable setting must be greater than or equal to 11.5dB, 21.5dB, or 31.5dB respectively. This operation includes hysteresis and debounce to avoid the AGC gain from cycling between high gain and 0 dB when signals are near the noise threshold level. When utilizing the AGC noise threshold, it is recommended to configure the 1st order IIR filter as a high-pass filter to achieve best performance. The noise (or silence) detection feature can be entirely disabled by the user.
- 6. **Max PGA applicable** allows the designer to restrict the maximum gain applied by the AGC. This can be used for limiting PGA gain in situations where environmental noise is greater than the programmed noise threshold. Microphone input Max PGA can be programmed from 0 dB to 58 dB in steps of 0.5 dB.
- 7. **Hysteresis**, as the name suggests, sets a window around the Noise Threshold which must be exceeded to either detect that the recorded signal is indeed noise or signal. If initially the energy of the recorded signal is greater than the Noise Threshold, then the AGC recognizes it as noise only when the energy of the recorded signal falls below the Noise Threshold by a value given by Hysteresis. Similarly, after the recorded signal is recognized as noise, for the AGC to recognize it as a signal, its energy must exceed the Noise Threshold by a value given by the Hysteresis setting. In order to prevent the AGC from jumterminalg between noise and signal states, (which can happen when the energy of recorded signal is very close to the Noise threshold) a non-zero hysteresis value should be chosen. The Hysteresis feature can also be disabled.
- 8. **Debounce Time (Noise and Signal)** sets the hysteresis in time domain for noise detection. The AGC continuously calculates the energy of the recorded signal. If the calculated energy is less than the set Noise Threshold, then the AGC does not increase the input gain to achieve the Target Level. However, to handle audible artifacts which can occur when the energy of the input signal is very close to the Noise Threshold, the AGC checks if the energy of the recorded signal is less than the Noise Threshold for a time greater than the Noise Debounce Time. Similarly the AGC starts increasing the input-signal gain to reach the Target Level when the calculated energy of the input signal is greater than the Noise Threshold. Again, to avoid audible artifacts when the input-signal energy is very close to Noise Threshold, the energy of the input signal needs to continuously exceed the Noise Threshold value for the Signal Debounce Time. If the debounce times are kept very small, then audible artifacts can result by rapid enabling and disabling the AGC function. At the same time, if the Debounce time is kept too large, then the AGC may take time to respond to changes in levels of input signals with respect to Noise Threshold. Both noise and signal debounce time can be disabled.
- 9. The **AGC Noise Threshold Flag** is a read-only flag indicating that the input signal has levels lower than the Noise Threshold, and thus is detected as noise (or silence). In such a condition the AGC applies a gain of 0 dB.
- 10. **Gain Applied by AGC** is a ready-only register setting which gives a real-time feedback to the system on the gain applied by the AGC to the recorded signal. This, along with the Target Setting, can be used to determine the input signal level. In a steady state situation Target Level (dB) = Gain Applied by AGC (dB) + Input Signal Level (dB) When the AGC noise threshold flag is set, then the status of gain applied by AGC should be ignored.
- 11. The **AGC Saturation Flag** is a read-only flag indicating that the ADC output signal has not reached its Target Level. However, the AGC is unable to increase the gain further because the required gain is higher than the Maximum Allowed PGA gain. Such a situation can happen when the input signal has very low energy and the Noise Threshold is also set very low. When the AGC noise threshold flag is set, the status of AGC saturation flag should be ignored.
- 12. The **ADC Saturation Flag** is a read-only flag indicating an overflow condition in the ADC channel. On overflow, the signal is clipped and distortion results. This typically happens when the AGC Target Level is kept very high and the energy in the input signal increases faster than the Attack Time.
- 13. An **AGC low-pass filter** is used to help find the average level of the input signal. This average level is compared to the programmed detection levels in the AGC to provide the correct functionality. This low pass filter is in the form of a first-order IIR filter. Three 8-bit registers are used to form the 24-bit digital coefficient as shown on the register map. In this way, a total of 9 registers are programmed to form the 3 IIR coefficients. The transfer function of the filter implemented for signal level detection is given by

$$
H(z) = \frac{N_0 + N_1 z^{-1}}{2^{23} - D_1 z^{-1}}
$$

Where:

Coefficient N0 can be programmed by writing into B40_P1_R12, B40_P1_R13, and B40_P1_R14. Coefficient N1 can be programmed by writing into B40_P1_R16, B40_P1_R17, and B40_P1_R18. Coefficient D1 can be programmed by writing into B40_P1_R20, B40_P1_R21, and B40_P1_R22.

N0, N1 and D1 are 24-bit 2's complement numbers and their default values implement a low-pass filter with cut-off at 0.002735*ADC_FS .

See [Table](#page-46-0) 9 for various AGC programming options. AGC can be used only if analog microphone input is routed to the ADC channel.

(1)

Table 9. AGC Parameter Settings

Figure 45. AGC Characteristics

The TLV320AIC3268 ADC channel includes a built-in digital decimation filter to process the oversampled data from the to generate digital data at Nyquist sampling rate with high dynamic range. The decimation filter can be chosen from three different types, depending on the required frequency response, group delay and sampling rate.

8.3.2.3 ADC Processing Blocks

The TLV320AIC3268 offers a range of processing blocks which implement various signal processing capabilities along with decimation filtering. These processing blocks give users the choice of how much and what type of signal processing they may use and which decimation filter is applied.

[Table](#page-48-0) 10 gives an overview of the available processing blocks of the ADC channel and their properties. The Resource Class Column (RC) gives an approximate indication of power consumption.

The signal processing blocks available are:

- **First-order IIR**
- AGC

Additional signal processing functions are acheived by programming the miniDSP. For specific details on available processing functions, please contact Texas Instruments.

The processing blocks are tuned for mono or stereo use cases and can achieve high anti-alias filtering. The available first order IIR, has fully user programmable coefficients.

Table 10. ADC Processing Blocks

(1) Default

8.3.2.4 ADC Processing Blocks – Details

8.3.2.4.1 1 st order IIR, AGC, Filter A

Figure 46. Signal Chain for PRB_R1 and PRB_R4

8.3.2.5 User Programmable Filters

The built-in processing block in TLV320AIC3268 has a user programmable first order IIR filter. This filter can be used for dc-blocking purposes. The user programmable coefficients allow the user to control the cut-off frequency of the high pass filter.

The coefficients of these filters are each 24-bits wide, in two's-complement and occupy 3 consecutive 8-bit registers in the register space. For default values please see [Table](#page-235-0) 56.

8.3.2.5.1 1 st Order IIR Section

The transfer function for the first order IIR Filter is given by

$$
H(z) = \frac{N_0 + N_1 z^{-1}}{2^{23} - D_1 z^{-1}}
$$

(2)

The frequency response for the 1st order IIR Section with default coefficients is flat at a gain of 0dB. Details on ADC coefficient default values are given in [Table](#page-235-0) 56.

Filter	Filter Coefficient	ADC Coefficient Left Channel	ADC Coefficient Right Channel		
1 st Order $\overline{\text{IIR}}$	N0	C4 (B40 P1 R24-R26)	C36 (B40 P2 R32-R34)		
	N1	C5 (B40 P1 R28-R30)	C37 (B40_P2_R36-R38)		
	D1	C6 (B40 P1 R32-R34)	C39 (B40 P2 R40-R42)		

Table 11. ADC 1st order IIR Filter Coefficients

8.3.2.6 Decimation Filter

The TLV320AIC3268 offers 3 different types of decimation filters. The integrated digital decimation filter removes high-frequency content and down samples the audio data from an initial sampling rate of AOSR*f_S to the final output sampling rate of f_S . The decimation filtering is achieved using a higher-order CIC filter followed by linearphase FIR filters. The decimation filter cannot be chosen by itself, it is implicitly set through the chosen processing block.

The following subsections describe the properties of the available filters A, B and C.

8.3.2.6.1 Decimation Filter A

This filter is intended for use at sampling rates up to 48kHz. When configuring this filter, the oversampling ratio of the ADC can either be 128 or 64. For highest performance the oversampling ratio must be set to 128. Using AOSR of 64 for sampling rates up to 48kHz, gives lower power consumption, but degrades the signal to noise ratio (SNR).

Filter A can also be used for 96kHz at an AOSR of 64.

ADC Channel Response for Decimation Filter A

(Red line corresponds to –73 dB)

8.3.2.6.2 Decimation Filter B

Filter B is intended to support sampling rates up to 96kHz at a oversampling ratio of 64.

ADC Channel Response for Decimation Filter B (Red line corresponds to –44 dB)

Figure 48. ADC Decimation Filter B, Frequency Response

8.3.2.6.3 Decimation Filter C

Filter type C along with AOSR of 32 is specially designed for 192ksps operation for the ADC. The pass band which extends up to $0.11[*]f_S$ (corresponds to 21kHz), is suited for audio applications.

ADC Channel Response for Decimation Filter C (Red line corresponds to –60 dB)

Figure 49. ADC Decimation Filter C, Frequency Response

8.3.2.7 ADC Data Interface

The decimation filter and signal processing block in the ADC channel passes 32-bit data words to the audio serial interface once every cycle of Fs,ADC. During each cycle of Fs,ADC, a pair of data words (for left and right channel) are passed. The audio serial interface rounds the data to the required word length of the interface before converting to serial data as per the different modes for audio serial interface.

8.3.2.8 ADC Special Functions

8.3.2.8.1 Power Tune Modes

As a part of PowerTune strategy, the analog performance of ADC can be adjusted. As a consequence the power consumption on AVDDx_18 supplies can be traded off with the speed of operation (ADC_MOD_CLK) and performance.

The TLV320AIC3268 supports 4 different power tune modes for ADC, PTM_R1, PTM_R2, PTM_R3 and PTM, R4, which can be set by programming B0, P1, R61, D[7:6]. The PTM, R4 is the default mode and gives the best performance for ADC with AOSR=128 and ADC_FS up to 48ksps. At lower speeds of ADC_MOD_CLK (AOSR*ADC_FS) lower PTM modes can be used for the benefit of lower power consumption, for example for AOSR=64 and ADC_FS=8ksps, PTM_R1 can be used. Using lower PTM modes for higher frequencies of ADC_MOD_CLK can result in reduction of peak amplitude of analog inputs where the distortion performance sets in. In general , PTM_R1 is recommended till ADC_MOD_CLK up to 0.768MHz, PTM_R2 for ADC_MOD_CLK up to 1.536MHz and PTM_R3 for ADC_MOD_CLK up to 3.072MHz. In applications where power consumption is not very critical, PTM_R4 is recommended to be used for best performance.

8.3.2.8.2 Microphone Bias

The TLV320AIC3268 has two built-in low noise Microphone Bias terminals for electret-condenser microphones: MICBIAS and MICBIAS_EXT. Typically, MICBIAS is utilized for onboard microphones, while MICBIAS_EXT provides a microphone bias for inserted headsets. Each bias amplifier can support up to 8mA of load current to support multiple microphones. Each bias amplifier has been designed to provide a combination of high PSRR, low noise and programmable bias voltages to allow the user to fine tune the biasing to specific microphone combinations. To support a wide range of bias voltages, the bias amplifier can work off either a low analog supply or the higher AVDD3_33 analog supply. To support a wide range of bias voltages, the MICBIAS and MICBIAS_EXT voltage are generated through an onchip low-dropout regulator. Thus, programmed voltages should be 300mV below MICBIAS_VDD.

Table 15. MICBIAS Voltage Control

Table 16. MICBIAS_EXT Voltage Control

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Table 16. MICBIAS_EXT Voltage Control (continued)

8.3.2.8.3 Digital Microphone Function

In addition to supporting analog microphones, the TLV320AIC3268 also interfaces to digital microphones.

The TLV320AIC3268 outputs internal clock ADC_MOD_CLK on several digital IO terminals. This clock can act as DigMic Clock. The generation of ADC_MOD_CLK is described in [Figure](#page-65-0) 58. The digital microphone data can be accepted on several terminals on both the rising edge as well as the falling edge of the DigMic Clock. [Table](#page-54-0) 17 describes the various ways in which digital microphone interface can be implemented using the several terminal options available in TLV320AIC3268. The TLV320AIC3268 supports two stereo channels of digital microphone. The stereo Digital Mic 1 channel can be muxed with the ADC modulator for analog inputs. The

additional stereo Digital Mic 2 channel is seperately supported. Both these channels use the same DigMic Clock. The Digital Mic 1 channel can be processed either by the built in processing available in PRB_Rx modes or be custom processed in miniDSP. The Digital Mic 2 channel can be processed by using the miniDSP. Each data line configured for digital microphone data can support either mono data or stereo data by using both edges of DigMic Clock. Some of the common digital microphone features as listed in [Table](#page-54-0) 17.

Table 17. Digital Microphone Features

Since the digital microphone signals do not pass through the ADC PGA block, the hardware AGC should not be enabled while using digital microphone inputs for left and right ADC channels.

Figure 51. Timing Diagram for Digital Microphone Interface

The digital-microphone mode can be selectively enabled for only-left, only-right, or stereo channels. When the digital microphone mode is enabled, the analog section of the ADC can be powered down and bypassed for power efficiency. The AOSR value for the ADC channel must be configured to select the desired decimation ratio to be achieved based on the external digital microphone properties.

[Figure](#page-55-0) 52 is a typical circuit diagram showing one possibility for connecting digital microphones. All terminal assignment options for digital microphones are described in Rows E and J of the terminal muxing tables in and (located in [Multifunction](#page-111-0) Terminals). Depending on the performance of the digital microphone (for example PSRR) and the noise level on the IOVDDx_33 power supply, some additional filtering may be needed for Vmic near the digital microphone for best performance.

8.3.2.8.4 Channel-to-Channel Phase Adjustment

The TLV320AIC3268 has a built-in feature to fine-adjust the phase between the stereo ADC record signals. The phase compensation is particularly helpful to adjust delays when using dual microphones for noise cancellation and other processing. This delay can be controlled in fine amounts in the following fashion.

Delay(7:0) = B0_P0_R85_D[7:0]

Where

$$
RIGHT _ ADC _ PHASE _ COMP(t) = RIGHT _ ADC _ OUT(t - t_{pr})
$$
\n(3)

where

$$
t_{pr} = \frac{\left(\text{Delay}(4:0) + \text{Delay}(6:5)^* \text{AOSR}^* k_f\right)}{\text{AOSR}^* \text{ADC} - \text{FS}}
$$
\n(4)

Where k_f is a function of the decimation filter:

and

LEFT $_ ADC _\text{PHASE} _\text{COMP} (t) = \text{LEFT} _\text{ADC} _\text{OUT} (t - t_{\text{pl}})$

Where

$$
t_{pl} = \frac{\text{Delay}(7)}{\text{ADC}_\text{FS}}\tag{6}
$$

This feature is available for stereo analog inputs or Digital Mic 1 channel.

(5)

8.3.2.8.5 Fast Charging AC Capacitors

The value of the coupling capacitor must be so chosen that the high-pass filter formed by the coupling capacitor and the input impedance do not affect the signal content. At power-up, before proper recording can begin, this coupling capacitor must be charged up to the common-mode voltage. To enable quick charging, the TLV320AIC3268 has modes to speed up the charging of the coupling capacitor. These are controlled by the values in B0_P1_R122_D[1:0].

8.3.2.8.6 Anti Thump

For normal voice or audio recording, the analog input terminals of the TLV320AIC3268, must be AC-coupled to isolate the DC-common mode voltage of the driving circuit from the common-mode voltage of the TLV320AIC3268.

When the analog inputs are not selected for any routing, the input terminals are 3-stated and the voltage on the terminals is undefined. When the unselected inputs are selected for any routing, the input terminals must charge from the undefined voltage to the input common-mode voltage. This charging signal can cause audible artifacts. In order to avoid such artifacts the TLV320AIC3268 also incorporates anti-thump circuitry to allow connection of unused inputs to the common-mode level. This feature is disabled by default, and can be enabled by writing the appropriate values into B0_P1_R58_D[7:0]. The use of this feature in combination with the PTM_R1 setting in B0_P0_R61 when the ADC channel is powered down causes the additional current consumption of 700μA from AVDDx_18 and 125μA from DVDDx_18 in the sleep mode.

8.3.2.8.7 Adaptive Filtering

After the ADC is running, the filter coefficients are locked and cannot be accessed for read or write. However the TLV320AIC3268 offers an adaptive filter mode as well. Setting B40_P0_R1_D2=1 turns on double buffering of the coefficients. In this mode filter coefficients can be updated through the host and activated without stopterminalg and restarting the ADC, enabling advanced adaptive filtering applications.

To support double buffering, all coefficients are stored in two buffers (Buffer A and B). When the ADC is running and adaptive filtering mode is turned on, setting the control bit B40_P0_R1_D0=1 switches the coefficient buffers at the next start of a sampling period. The bit reverts to 0 after the switch occurs. At the same time, the flag B40_P0_R1_D1 toggles.

The flag in B40_P0_R1_D1 indicates which of the two buffers is actually in use.

For B40 P0 R1 D1=0: Buffer A is in use by the ADC engine. For B40_P0_R1_D1=1: Buffer B is in use.

While the device is running, coefficient updates are always made to the buffer not in use by the ADC, regardless to which buffer the coefficients have been written.

8.3.3 DAC

The TLV320AIC3268 includes a stereo audio DAC supporting data rates from 8kHz to 192kHz. Each channel of the stereo audio DAC consists of a signal-processing engine with fixed processing blocks, a digital interpolation filter, multi-bit digital delta-sigma modulator, and an analog reconstruction filter. The DAC is designed to provide enhanced performance at low sampling rates through increased oversampling and image filtering, thereby keeping quantization noise generated within the delta-sigma modulator and signal images strongly suppressed within the audio band to beyond 20kHz. To handle multiple input rates and optimize power dissipation and performance, the TLV320AIC3268 allows the system designer to program the oversampling rates over a wide range from 1 to 1024. The system designer can choose higher oversampling ratios for lower input data rates and lower oversampling ratios for higher input data rates.

The TLV320AIC3268 DAC channel includes a built-in digital interpolation filter to generate oversampled data for the sigma-delta modulator. The interpolation filter can be chosen from three different types depending on required frequency response, group delay and sampling rate.

The DAC path of the TLV320AIC3268 features many options for signal conditioning and signal routing:

- 2 headphone amplifiers
	- Usable in single-ended stereo or differential mono mode
	- $-$ Analog volume setting with a range of -6 to $+14$ dB
- Line-out amplifiers
	- Usable in streo single-ended or stereo differential modes
- Class-D speaker amplifier
	- Usable with left, right, or monophonic mix modes
	- Analog volume control with a settings of +6, +12, +18, +24, and +30 dB
- 1 Receiver amplifier
	- Usable in mono differential mode
	- $-$ Analog volume setting with a range of -6 to $+29$ dB
- Digital volume control with a range of -63.5 to +24dB
- Mute function

In addition to the standard set of DAC features the TLV320AIC3268 also offers the following special features:

- Digital auto mute
- Adaptive coefficient update mode

In addition to the above signal processing functions, extensive algorithms are available by programming the miniDSP. For specific details on available algorithms, please contact Texas Instruments.

8.3.3.1 DAC Processing Blocks

The TLV320AIC3268 implements signal processing capabilities and interpolation filtering via processing blocks. These fixed processing blocks give users the choice of how much and what type of signal processing they may use and which interpolation filter is applied.

[Table](#page-58-0) 18 gives an overview over all available processing blocks of the DAC channel and their properties. The Resource Class Column (RC) gives an approximate indication of power consumption.

The signal processing blocks available are:

- First-order IIR
- Scalable number of biquad filters

The processing blocks can be chosen based on mono or stereo playback requirement and need for additional filtering for frequency shaping. The available first-order IIR and biquad filters have fully user-programmable coefficients.

Table 18. Overview – DAC Predefined Processing Blocks

(1) Default

8.3.3.2 DAC Processing Blocks – Details

8.3.3.2.1 2 Biquads, Interpolation Filter A

Figure 53. Signal Chain for PRB_P1 (Stereo)

8.3.3.2.2 6 Biquads, 1st order IIR, Interpolation Filter A or B

Figure 54. Signal Chain for PRB_P3 (Stereo) and PRB_P6 (Left)

8.3.3.3 User Programmable Filters

The TLV320AIC3268 allows either 3 or 6 user programmable BiQuad filters in DAC channel. Some processing blocks also support a first order IIR filter for dc-blocking.

The coefficients of the available filters are arranged as sequentially-indexed coefficients in two banks. If adaptive coefficient update mode is chosen, the coefficient banks can be switched on-the-fly. For more details on adaptive filtering please see [Adaptive](#page-63-0) Filtering.

The coefficients of these filters are each 24-bits wide, in two's-complement and occupy 3 consecutive 8-bit registers in the register space. For default values please see DAC [Defaults](#page-241-0).

8.3.3.3.1 1st-Order IIR Section

The IIR is of first-order and its transfer function is given by

$$
H(z) = \frac{N_0 + N_1 z^{-1}}{2^{23} - D_1 z^{-1}}
$$

(7)

The frequency response for the 1st order IIR Section with default coefficients is flat. Details on DAC coefficient default values are given in DAC [Defaults.](#page-241-0)

Table 19. DAC IIR Filter Coefficients

(8)

8.3.3.3.2 Biquad Section

The transfer function of each of the Biquad Filters is given by

$$
H(z) = \frac{N_0 + 2 \cdot N_1 z^{-1} + N_2 z^{-2}}{2^{23} - 2 \cdot D_1 z^{-1} - D_2 z^{-2}}
$$

The frequency response for each biquad section with default coefficients is flat at a gain of 0dB. Details on DAC coefficient default values are given in DAC [Defaults.](#page-241-0)

Table 20. DAC Biquad Filter Coefficients

8.3.3.4 Interpolation Filters

8.3.3.4.1 Interpolation Filter A

Filter A is designed for an Fs up to 48ksps with a flat passband of 0kHz–20kHz.

Table 21. DAC Interpolation Filter A, Specification

G016 **Figure 55. DAC Interpolation Filter A, Frequency Response**

8.3.3.4.2 Interpolation Filter B

Filter B is specifically designed for an Fs of above 96ksps. Thus, the flat pass-band region easily covers the required audio band of 0-20kHz.

Table 22. DAC Interpolation Filter B, Specification

DAC Channel Response for Interpolation Filter B (Red line corresponds to –58 dB)

8.3.3.4.3 Interpolation Filter C

Filter C is specifically designed for the 192ksps mode. The pass band extends up to 0.40*Fs (corresponds to 80kHz), more than sufficient for audio applications.

Figure 57. DAC Interpolation Filter C, Frequency Response

8.3.3.5 DAC Gain Setting

8.3.3.5.1 PowerTune Modes

As part of the PowerTune strategy, the analog properties of the DAC are adjusted. As a consequence, the fullscale signal swing achieved at the headphone outputs must be adjusted.

Please see [Table](#page-62-0) 24 for the proper gain compensation values across the different combinations.

Table 24. DAC Gain versus PowerTune Modes

8.3.3.5.2 Digital Volume Control

The TLV320AIC3268 signal processing blocks incorporate a digital volume control block that can control the volume of the playback signal from +24dB to –63.5dB in steps of 0.5dB. These can be controlled by writing to B0 P0 R65 and B0 P0 R66. The volume control of left and right channels by default can be controlled independently, but by programming B0_P0_R64_D[1:0], they can be made interdependent. The volume changes are soft-stepped in steps of 0.5dB to avoid audible artifacts during gain change. The rate of soft-stepping can be controlled by programming B0_P0_R63_D[1:0] to either one step per frame (DAC_FS) or one step per 2 frames. The soft-stepping feature can also be entirely disabled. During soft-stepping the value of the actual applied gain would differ from the programmed gain in register. The TLV320AIC3268 gives a feedback to the user in form of register readable flag to indicate that soft-stepping is currently in progress. The flags for left and right channels can be read back by reading B0_P0_R38_D4 and B0_P0_R38_D0 respectively. A value of 0 in these flags indicates a soft-stepping operation in progress, and a value of 1 indicates that soft-stepping has completed. A soft-stepping operation comes into effect during a) power-up, when the volume control soft-steps from –63.5dB to programmed gain value b) volume change by user when DAC is powered up and c) power-down, when the volume control block soft-steps to –63.5dB before powering down the channel.

8.3.3.6 DAC Special Functions

8.3.3.6.1 Digital Auto Mute

The TLV320AIC3268 also incorporates a special feature, in which the DAC channel is auto-muted when a continuous stream of DC-input is detected. When using PRB_Px modes, the Data Input 1 (L1,R1) ports of miniDSP_D are monitoted for DC-inputs. Signals routed to Data Input 1 port is controlled by configuring B0_P4_R118_D[5:4].By default, this feature is disabled. It can be enabled by writing a non-000 value into B0_P0_R64_D[6:4]. The non-zero value controls the duration of continuous stream of DC-input before which the auto-mute feature takes effect. This feature is especially helpful for eliminating high-frequency-noise power being delivered into the load even during silent periods of speech or music.

8.3.3.6.2 Adaptive Filtering

When the DAC is running, the user-programmable filter coefficients are locked and cannot be accessed for either read or write.

However, the TLV320AIC3268 offers an adaptive filter mode as well, and the DAC contains two separate adaptive filter coefficient banks (Primary Adaptive Bank in Book 80, and Secondary Adaptive Bank in Book 82). Setting B80_P0_R1_D2=1 for the Primary Adaptive Bank will turn on double buffering of the coefficients. Similarly, setting B82_P0_R1_D2=1 will turn on double buffering of the coefficients in the Secondary Adaptive Bank. In this mode, filter coefficients can be updated through the host, and activated without stopping and restarting the DAC. This enables advanced adaptive filtering applications.

In the double-buffering scheme, all coefficients are stored in two buffers (Buffers A and B). When the DAC is running and adaptive filtering mode is turned on, setting the control bit B80 P0 R1 D0=1 (B82 P0 R1 D0=1 if using Secondary Bank) switches the coefficient buffers at the next start of a sampling period. This bit is set back to 0 after the switch occurs. At the same time, the flag B80 P0 R1 D1 (B82 P0 R1 D1 if using Secondary Bank) toggles.

The flag in B80 P0 R1 D1 indicates which of the two buffers in the Primary Bank is actually in use.

B80 P0 R1 D1=0: Buffer A is in use by the DAC engine, Bit D1=1: Buffer B is in use.

While the device is running, coefficient updates are always made to the buffer not in use by the DAC, regardless to which buffer the coefficients have been written.

The user programmable coefficients C1 to C70 are defined on B80_P1-P3 for Buffer A and B80_P9-P11 for Buffer B. For the Secondary Bank, the coefficients are located on similar pages on Book 82.

8.3.4 Clock Generation and PLL

To minimize power consumption, the system ideally provides a master clock that is a suitable integer multiple of the desired sampling frequencies. In such cases, internal dividers can be programmed to set up the required internal clock signals at very low power consumption. For cases where such master clocks are not available, the built-in PLL can be used to generate a clock signal that serves as an internal master clock. In fact, this master clock can also be routed to an output terminal and may be used elsewhere in the system. The clock system is flexible enough that it even allows the internal clocks to be derived directly from an external clock source, while the PLL is used to generate some other clock that is only used outside the TLV320AIC3268.

The ADC_CLKIN and DAC_CLKIN can then be routed through highly-flexible clock dividers to generate the various clocks required for ADC, DAC and the miniDSP sections.

The TLV320AIC3268 supports a wide range of options for generating clocks for the ADC and DAC sections as well as the interface and other control blocks as shown in [Figure](#page-65-0) 58. The clocks for the ADC and the DAC require a source reference clock. In the TLV320AIC3268 the ADC and DAC clock-trees can have different root clocks. These clocks can be provided on a variety of device terminals such as MCLK, BCLK1, BCLK2, GPIO1, GPIO2 and GPIO3, and the onchip high-frequency reference clock (HF_REF_CLK) and high-frequency oscillator clock (HF_OSC_CLK) can also be provided as sources. The source reference clock for the ADC can be chosen by programming the ADC_CLKIN value on B0_P0_R4_D[3:0]. The source reference clock for the DAC can be chosed by programming the DAC_CLKIN value on B0_P0_R4_D[7:4]. The ADC_CLKIN and DAC_CLKIN can then be routed through highly flexible clock dividers shown in [Figure](#page-65-0) 58 to generate the various clocks required for the ADC, DAC, and miniDSP sections. In the event that the desired audio miniDSP clocks cannot be generated from the reference clocks coming from the device terminals listed above, the TLV320AIC3268 also provides the option of using the on-chip PLL which supports a wide range of fractional multiplication values to generate the required clocks. Starting from ADC_CLKIN and DAC_CLKIN, the TLV320AIC3268 provides several programmable clock dividers to help achieve a variety of sampling rates for the ADC and DAC, as well as clocks for the miniDSP sections.

The DAC and ADC clocks are obtained as follows:

$$
DAC_{s} = \frac{DAC_{c}CLKIN}{NDAC \times MDAC \times DOSR}
$$
\n
$$
DAC_{c} = \frac{DAC_{c}CLKIN}{NDAC \times MDAC}
$$
\n
$$
ADC_{s} = \frac{ADC_{c}CLKIN}{NADC \times MDAC}
$$
\n
$$
(10)
$$
\n
$$
ADC_{s} = \frac{ADC_{c}CLKIN}{NADC \times MADC \times AOSR}
$$
\n
$$
(11)
$$

The DAC Modulator is clocked by DAC_MOD_CLK. For proper power-up of the DAC Channel, these clocks must be enabled by configuring the NDAC and MDAC clock dividers (B0_P0_R11_D7=1 and B0_P0_R12_D7=1). When the DAC channel is powered down, the device internally initiates a power-down sequence for proper shutdown. During this shut-down sequence, the NDAC and MDAC dividers must not be powered down, or else a proper low power shut-down may not take place. The user can read the power-status flag in B0_P0_R37_D7 for the Left DAC and B0_P0_R37_D3 for the Right DAC. When both flags indicate power-down, the MDAC divider may be powered down, followed by the NDAC divider.

The ADC modulator is clocked by ADC_MOD_CLK. For proper power-up of the ADC Channel, these clocks are enabled by the NADC and MADC clock dividers (B0_P0_R18_D7=1 and B0_P0_R19_D7=1). When the ADC channel is powered down, the device internally initiates a power-down sequence for proper shut-down. During this shut-down sequence, the NADC and MADC dividers must not be powered down, or else a proper low power shut-down may not take place. The user can read the power-status flag in B0_P0_R36_D6 for the Left ADC and B0 P0 R36 D2 for the Right ADC. When both flags indicate power-down, the MADC divider may be powered down, followed by NADC divider.

Table 25. DAC CLKIN and ADC CLKIN Clock Dividers

Table 26. DAC and ADC Clock Selectors

DAC_CLKIN BO_PO_R4_D[7:4] MCLK1, MCLK2, BCLK1, GPIO1, PLL_CLK, BCLK2, GPI1, HR_REF_CLK, BOLACLAREF_CLK, GPIO2, GPI2

ADC_CLKIN B0_P0_R4_D[3:0] MCLK1, MCLK2, BCLK1, GPIO1, PLL_CLK, BCLK2, GPI1, HR_REF_CLK, ADC_CLK, GPIO2, GPI2

By default ADC_CLK = DAC_CLK and ADC_MOD_CLK = DAC_MOD_CLK.

The registers used for DAC and ADC clock selection are listed in [Table](#page-66-0) 26.

ADC_CLK B0_P0_R18_D7 NDAC output (DAC_CLK), NADC output ADC_MOD_CLK B0_P0_R19_D7 MDAC output (DAC_MOD_CLK), MADC output

Selector Bits Inputs

or

 $\mathsf{ADC}_\mathsf{MDD}_\mathsf{CLK} = \frac{\mathsf{DAC}_\mathsf{CLKIN}}{\mathsf{MDAG}_\mathsf{CLKID}}$

 Ω

DAC_CLKIN ADC_MOD_CLK = NDAC MDAC ´

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 (14)

(13)

(12)

STRUMENTS

When ADC_CLK is derived from the NDAC divider output, the NDAC must be kept powered up till the powerdown status flags for ADC do not indicate that the ADC is still in the process of powering down. When the input to the AOSR clock divider is derived from DAC_MOD_CLK, then MDAC must be powered up when ADC_FS is needed (that is when WCLK is generated by TLV320AIC3268 or AGC is enabled) and can be powered down only after the ADC power-down flags indicate power-down status.

In general, all the root clock dividers should be powered down only after the child clock dividers have been powered down for proper operation.

The TLV320AIC3268 also has options for routing some of the internal clocks to the output terminals of the device to be used as general purpose clocks in the system.

For example, the TLV320AIC3268 can be configured to drive the bit clock signals ASI1_BCLK_OUT, ASI2_BCLK_OUT, and ASI3_BCLK_OUT on the three serial interfaces as shown in [Figure](#page-67-0) 59.

Figure 59. Bit Clock Output Options for ASI1, ASI2, and ASI3

When TLV320AIC3268 is configured to drive ASI1 BCLK OUT, the clock signal can be selected via B0_P4_R14_D[6:4] to come from ASI1_BDIV_OUT, ASI2_BDIV_OUT, ASI3_BDIV_OUT, ASI2_BCLK Input, or ASI3_BCLK Input.

When TLV320AIC3268 is configured to drive ASI2 BCLK OUT, the clock signal can be selected via B0_P4_R30_D[6:4] to come from ASI1_BDIV_OUT, ASI2_BDIV_OUT, ASI3_BDIV_OUT, ASI1_BCLK Input, or ASI3_BCLK Input.

When TLV320AIC3268 is configured to drive ASI3_BCLK_OUT, the clock signal can be selected via B0_P4_R46_D[6:4] to come from ASI1_BDIV_OUT, ASI2_BDIV_OUT, ASI3_BDIV_OUT, ASI1_BCLK, or ASI2_BCLK.

ASI1_BDIV_OUT is a divided value of ASI1_BDIV_CLKIN, where the division value can be programmed in B0_P4_R12_D[6:0] from 1 to 128, and this bit clock divider can be powered on by setting B0_P4_R12_D7. The ASI1 BDIV CLKIN can itself be configured to be one of DAC CLK, DAC MOD CLK, ADC CLK or ADC_MOD_CLK by configuring the ASI1_BDIV_CLKIN mux in B0_P4_R11_D[1:0].

ASI2_BDIV_OUT is a divided value of ASI2_BDIV_CLKIN, where the division value can be programmed in B0_P4_R28_D[6:0] from 1 to 128, and this bit clock divider can be powered on by setting B0_P4_R28_D7. The ASI2 BDIV CLKIN can itself be configured to be one of DAC CLK, DAC MOD CLK, ADC CLK or ADC_MOD_CLK by configuring the ASI2_BDIV_CLKIN mux in B0_P4_R27_D[1:0].

ASI3 BDIV OUT is a divided value of ASI3 BDIV CLKIN, where the division value can be programmed in B0_P4_R44_D[6:0] from 1 to 128, and this bit clock divider can be powered on by setting B0_P4_R44_D7. The ASI3_BDIV_CLKIN can itself be configured to be one of DAC_CLK, DAC_MOD_CLK, ADC_CLK or ADC_MOD_CLK by configuring the ASI3_BDIV_CLKIN mux in B0_P4_R43_D[1:0].

The TLV320AIC3268 can also be configured to provide the world clocks for ASI1, ASI2, and ASI3 as shown in [Figure](#page-68-0) 60.

Figure 60. Word Clock Options for ASI1, ASI2, and ASI3

ASI1_WCLK_OUT can be selected to come from DAC_FS, ADC_FS, ASI1_WDIV_OUT, ASI2_WDIV_OUT, ASI3_WDIV_OUT, as well as ASI2_WCLK Input, and ASI3_WCLK Input using B0_P4_R14_D[2:0]. ASI1 WDIV OUT is driven as a divided value of ASI1 BCLK, where the division can be programmed in B0_P4_R13_D[6:0] from 32 to 128, and this word clock divider can be powered on by setting B0_P4_R13_D7.

ASI2_WCLK_OUT can be selected to come from DAC_FS, ADC_FS, ASI1_WDIV_OUT, ASI2_WDIV_OUT, ASI3_WDIV_OUT, as well as ASI1_WCLK Input, and ASI3_WCLK Input using B0_P4_R30_D[2:0]. ASI2_WDIV_OUT is driven as a divided value of ASI2_BCLK, where the division can be programmed in B0_P4_R29_D[6:0] from 32 to 128, and this word clock divider can be powered on by setting B0_P4_R29_D7.

ASI3_WCLK_OUT can be selected to come from DAC_FS, ADC_FS, ASI1_WDIV_OUT, ASI2_WDIV_OUT, ASI3_WDIV_OUT, as well as ASI1_WCLK Input, and ASI2_WCLK Input using B0_P4_R46_D[2:0]. ASI3_WDIV_OUT is driven as a divided value of ASI3_BCLK, where the division can be programmed in B0_P4_R45_D[6:0] from 32 to 128, and this word clock divider can be powered on by setting B0_P4_R45_D7.

The bit clock and work clock dividers are summarized in [Table](#page-68-1) 27. The bit clock and word clock selectors are summarized in [Table](#page-69-0) 28.

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Table 28. ASI1, ASI2, and ASI3 Bit and Word Clock Selection

Additionally a general purpose clock CLKOUT can be driven out on DOUT1, WCLK2, BCLK2, GPIO1, GPIO2, or GPO1 according to the settings in [Table](#page-69-1) 29.

Table 29. CLKOUT Selection

This clock can be a divided down version of CDIV_CLKIN. The value of this clock divider can be programmed from 1 to 128 by writing to B0_P0_R22_D[6:0], and this CDIV clock divider can be powered on by setting B0_P4_R22_D7. The CDIV_CLKIN can itself be programmed as one of the clocks among the list shown in [Figure](#page-70-0) 61. This can be controlled by programming the mux in B0_P0_R21_D[3:0].

Figure 61. General Purpose Clock Output Options

8.3.4.1 PLL

The TLV320AIC3268 has an on-chip PLL to generate the clock frequency for the audio ADC, DAC, and Digital Signal Processing blocks. The programmability of the PLL allows operation from a wide variety of clocks that may be available in the system. The PLL Clocking and muxing is shown in [Figure](#page-71-0) 62.

(15)

The PLL input supports clocks varying from 512kHz to 20MHz and is register programmable to enable generation of required sampling rates with fine resolution. The PLL can be turned on by writing to B0_P0_R6_D7. When the PLL is enabled, the PLL output clock PLL_CLK is given by the following equation PLL_CLK = (PLL_CLKIN x R x J.D)/(PxPLL_CLKIN_DIV)

$$
PLL_CLK = \frac{PLL_CLKIN \times R \times J.D}{PxPLL_CLKIN_DIV}
$$

 $R = 1, 2, ... 16$.

 $J = 1, 2, 3, 4, \ldots$ 63, and $D = 0, 1, 2, 3, 4, \ldots$ 9999

$$
P = 1, 2, 3, 4, \ldots 8
$$

PLL CLKIN DIV = 1, 2, ... 128.

R, J, D, P, and PLL_CLKIN_DIV are register programmable.

The PLL can be programmed via B0_P0_R6-R10. The PLL can be turned on via B0_P0_R6_D7. The variable P can be programmed via B0 P0 R6 D[6:4]. The default register value for P is 1. The variable R can be programmed via B0_P0_R6_D[3:0]. The default register value for R is 1. The variable J can be programmed via B0_P0_R7_D[5:0]. The default register value for J is 4. The variable D is 12-bits, programmed into two registers. The MSB portion can be programmed via B0_P0_R8_D[5:0], and the LSB portion is programmed via B0_P0_R9_D[7:0]. The default register value for D is 0. The PLL_CLKIN_DIV value can be programmed via B0_P0_R10_D[6:0]. The default register value for PLL_CLKIN_DIV is 1.

When the PLL is enabled the following conditions must be satisfied

When the PLL is enabled and $D = 0$, the following conditions must be satisfied for PLL CLKIN:

$$
512 \text{ kHz} \leq \frac{PLL_CLKIN}{P \times PL_CLKIN_DIV} \leq 20 \text{ MHz}
$$
\n(16)

When the PLL is enabled and $D \neq 0$, the following conditions must be satisfied for PLL CLKIN:

$$
10 \text{ MHz} \leq \frac{\text{PLL_CLKIN}}{\text{P} \times \text{PLL_CLKIN_DIV}} \leq 20 \text{ MHz}
$$
\n(17)

In the TLV320AIC3268 the PLL_CLK supports a wide range of output clock values, based on register settings and power-supply conditions.

AVdd	PLL Mode B0 P0 R5 D6	Min PLL CLK frequency (MHz)	Max PLL CLK frequency (MHz)		
$≥1.5V$		80	103		
		95	110		
$≥1.65V$		80	118		
		92	123		
$≥1.80V$		80	132		
		92	137		

Table 31. PLL_CLK Frequency Range

The PLL can be powered up independently from the ADC and DAC blocks, and can also be used as a general purpose PLL by selecting its output as an input to the General Purpose Output Clock mux (enabling routing to a variety of digital output terminals). After powering up the PLL, PLL_CLK is available typically after 10ms. The PLL output frequency is controlled by J.D and R dividers

The D-divider value is 14-bits wide and is controlled by 2 registers. For proper update of the D-divider value, B0_P0_R8 must be programmed first followed immediately by B0_P0_R9. Unless the write to B0_P0_R9 is completed, the new value of D will not take effect.

The clocks for codec and various signal processing blocks, ADC_CLKIN and DAC_CLKIN can be generated from MCLK, BCLK1, GPIO1, BCLK2, GPIO3, HF_REF_CLK, HF_OSC_CLK, GPIO2, GPIO6, BCLK3 or PLL_CLK (B0_P0_R4_D[7:0]).

If the ADC_CLKIN and/or the DAC_CLKIN are derived from the PLL, then the PLL must be powered up first and powered down last.

[Table](#page-73-0) 32 lists several example cases of typical MCLK rates and how to program the PLL to achieve a sample rate Fs of either 44.1kHz or 48kHz.

$Fs = 44.1kHz$											
MCLK (MHz)	PLL CLKIN DIV	PLLP	PLLR	PLLJ	PLLD	MADC	NADC	AOSR	MDAC	NDAC	DOSR
2.8224	1	1	3	10	0	3	5	128	3	5	128
5.6448	1	$\mathbf{1}$	3	5	Ω	3	5	128	3	5	128
12	$\mathbf{1}$	$\mathbf{1}$	$\mathbf{1}$	$\overline{7}$	560	3	5	128	3	5	128
13	1	$\mathbf{1}$	2	4	2336	13	3	64	4	6	104
16	1	1	$\mathbf{1}$	5	2920	3	5	128	3	5	128
19.2	1	1	1	4	4100	3	5	128	3	5	128
48	1	$\overline{4}$	1	$\overline{7}$	560	3	5	128	3	5	128
$Fs = 48kHz$											
2.048	1	$\mathbf 1$	3	14	$\mathbf 0$	2	$\overline{7}$	128	$\overline{7}$	$\overline{2}$	128
3.072	1	$\mathbf 1$	4	$\overline{7}$	0	2	$\overline{7}$	128	$\overline{7}$	2	128
4.096	1	1	3	$\overline{7}$	$\mathbf 0$	2	$\overline{7}$	128	$\overline{7}$	2	128
6.144	1	1	$\overline{2}$	$\overline{7}$	$\mathbf 0$	2	$\overline{7}$	128	$\overline{7}$	2	128
8.192	1	$\mathbf{1}$	4	3	Ω	2	8	128	4	$\overline{4}$	128
12		1	1	$\overline{7}$	1680	2	$\overline{7}$	128	$\overline{7}$	$\overline{2}$	128
16	1	1	1	5	3760	2	$\overline{7}$	128	$\overline{7}$	2	128
19.2	1	1	1	4	4800	2	$\overline{7}$	128	$\overline{7}$	2	128
48	1	4	1	$\overline{7}$	1680	2	7	128	7	2	128

Table 32. PLL Example Configurations

8.3.4.2 Low Frequency Reference Clock

To extend the frequency locking range of the on-chip PLL to an external clock at low frequencies, a clock frequency multiplier is used to generate its output clock with the frequency K times of its input reference clock frequency for the PLL to lock, where K is a 28-bit value of the control register bits {B0_P0_R25_D[3:0], B0_P1_R26, B0_P0R27, B0_P0_R28}. The reference clock source can be selected with the control register bits, B0_P0_R24_D[7:4]. The clock routing for the low frequency clock is shown in [Figure](#page-74-0) 63.

Figure 63. Low-Frequency Clocking

The output clock, HF_REF_CLOCK, is generated by delta-sigma modulation with a high frequency clock, HF_CLK. The source of HF_CLK can be setup by programming the control bits, B0_P0_R24_R[3:0]. If the onchip high frequency oscillator clock, HF_OSC_CLK, is selected as the source, it is recommended to calibrate the oscillator clock by following the proper calibration procedure before turning on the clock multiplier.

The HF_OSC_CLK can have large device-to-device variation of its default frequency. For proper functioning, the HF_OSC_CLK can be calibrated with respect to the LFR_CLKIN. This calibration happens at power-up of the block when this feature is enabled (HF_OSC_CLK is used by any other function). By default this calibration is enabled and if so desired can be disabled by writing B0_P0_R29_D5 = '0'. For calibrating the HF_OSC_CLK the 26-bit ratio of frequencies (Desired HF_OSC_CLK freq / Frequency of LFR_CLKIN) can be programmed in B0_P0_R29_D[1:0], B0_P0_R30_D[7:0], B0_P0_R31_D[7:0], and B0_P0_R32_D[7:0]. This ratio must be programmed before enabling this block. Also, the LFR_CLKIN must be present when the HF_OSC_CLK is enabled, and the LFR CLKIN frequency should be less than 50 kHz. This calibration is an approximate calibration, and the frequency of HF_OSC_CLK will approximately equal Programmed Ratio * LFR_CLKIN frequency. The error can be approx ±7MHz. The desired frequency should ideally be kept between 50MHz and 57.5MHz for good audio performance. Once the calibration is over, the calibrated clock will be available for use by other blocks. The HF_OSC_CLK has an additional programmability by which this block can be used even when AVDD1_18 supply is not powered up. This can be useful when a free running clock is required when AVDD1 18 is not powered as no other analog blocks may be powered up. This feature can be controlled by B0_P0_R29_D6.

For a better quality of the PLL clock, the clock multiplier output should be set at higher frequency by choosing a higher multiplication value of K, if there are multiple options. But the multiplied frequency should not be higher than $\%$ times of HF_REF_CLK frequency and the frequency has to be within the PLL locking range, 10-20MHz for D≠0 and 512kHz – 20MHz for D=0. To select HF_REF_CLK as the PLL reference, B0_P0_R5_D[5:2] should be set as '0110'.

8.3.5 Reference Voltage

All audio data converters require a DC reference voltage. The TLV320AIC3268 achieves its low-noise performance by internally generating a low-noise reference voltage. This reference voltage is generated using a band-gap circuit with a good PSRR performance. This audio converter reference voltage must be filtered externally using a minimum 1μF capacitor connected from the VREF_AUDIO terminal to analog ground (VSS).

To achieve low power consumption, this audio reference block is powered down when all analog blocks inside the device are powered down. In this condition, the VREF_AUDIO terminal is 3-stated. On powerup of any analog block, the audio reference block is also powered up and the VREF_AUDIO terminal settles to its steadystate voltage after the settling time (a function of the decoupling capacitor on the VREF_AUDIO terminal). This time is approximately equal to 1 second when using a 1μF decoupling capacitor. In the event that a faster power-

up is required, either the audio reference block can be kept powered up (even when no other analog block is powered up) by programming B0_P1_R122_D2 = 1. However, in this case, an additional 100μA of current from AVDDx_18 is consumed. Additionally, to achieve a faster powerup, a fast-charge option is also provided where the charging time can be controlled between 40ms and 120ms by programming B0_P1_R122_D[1:0]. By default, the fast charge option is enabled.

In addition, the TLV320AIC3268 can also generate a separate 1.25V DC reference which is utilized by the SAR ADC for measurement. This SAR reference voltage must also be filtered externally using a minimum 1µF capacitor connected from the VREF_SAR terminal to analog ground (AVSS).

To achieve low power consumption, this SAR reference block is powered down by default when SAR conversations are not occurring. The system could utilize this reference voltage outside of SAR ADC conversions by powering it continuously by programming B0 P3 R6 D5 =0.

8.3.6 SAR ADC

This section describes how to use the SAR ADC for the functions:

- Temperature measurement
- Battery measurement
- Auxiliary voltage measurement

The analog inputs of the TLV320AIC3268 are shown in [Figure](#page-75-0) 64.

Figure 64. Simplified Diagram of the SAR ADC Analog Input Section

The ADC is controlled by an ADC control register (B0_P3_R3_D[7:0]). Several modes of operation are possible, depending on the bits set in the control register. Channel selection, scan operation, resolution, and conversion rate may all be programmed through this register. These modes are outlined in the following sections for each type of analog input. The results of conversions made are stored in the appropriate result register.

The SAR ADC can be powered down forcefully by writing to B0_P3_R2_D7. Overall SAR configuration and mode is controlled by writing to B0_P3_R3_D[7:0].

Voltage Reference

The TLV320AIC3268 can use an internal voltage reference of 1.25V or an external reference through the reference control register (B0_P3_R6). The internal reference voltage should only be used in the single-ended mode for battery monitoring, for temperature measurement, and for using the auxiliary inputs.

The TLV320AIC3268 may use an external voltage reference (B0_P3_R6). In many systems, a 2.5V reference is supplied; however, this device supports a reference voltage up to the AVDDx_18 level. The external reference should be a low-noise signal and accordingly, depending on the application, it might need some R-C filtering at the VREF_SAR terminal.

This voltage reference should only be used in the single-ended mode for measuring the auxiliary inputs (IN1L/AUX1, IN1R/AUX2, and VBAT).

Variable Resolution

The TLV320AIC3268 provides three different resolutions for the ADC: 8, 10, or 12 bits. Lower resolutions are often practical for measurements such as system voltages. Performing the conversions at lower resolution reduces the amount of time it takes for the ADC to complete its conversion process, which lowers power consumption. The ADC resolution can be programmed by writing to B0_P3_R2_D[6:5].

Data Format

The TLV320AIC3268 output data is unsigned binary format and can be read from two 8-bit registers over the Control interface (SPI or I²C). The SAR ADC's output data is MSB zero appended to make a 16-bit word.

8.3.6.1 Conversion Clock and Conversion Time

The TLV320AIC3268 contains an internal oscillator (LF_OSC_CLK), which is used to drive the state machines inside the device that perform the many functions of the part. MCLK is also available as a high frequency clock source. The clock source (LF_OSC_CLK or divided down MCLK) is selected by writing to B0_P3_R17_D7. When using the high frequency clock source MCLK, the clock divider B0_P3_R17_D[6:0] should be programmed to result in output clock pulses to be larger than 40ns. This clock (ADC_SAR_Clock) is further divided down to provide a clock to run the SAR ADC. The division ratio for this clock is set by writing to B0_P3_R2_D[4:3]. The ability to change the conversion clock rate allows the user to choose the optimal value for the resolution, speed, and power. If the division value is used as 1 , the ADC is limited to 8-bit resolution, division value of 2 is suitable for 10-bit resolution; 12-bit resolution requires the division value to be set as 4 or 8(recommended).

Similarly, the internal delay timers operate on clocks derived from either the LF_OSC_CLK or MCLK. The clock selection is controlled by writing to B0_P0_R23_D7. When using MCLK as the clock source the clock divider B0_P0_R23_D[6:0] should be programmed to have the resultant output clock to be approximately 1MHz in frequency. To avoid asynchronous issues, the system should use the same value for both B0_P0_R23_D7 and B0_P3_R17_D7.

Details for clock selection can be seen in [Figure](#page-76-0) 65.

Figure 65. SAR ADC and Interval Timer Clock Select

Regardless of the conversion clock speed, the internal clock runs nominally at 8.2 MHz. The conversion time of the TLV320AIC3268 depends on several functions. While the conversion clock speed plays an important role in the time it takes for a conversion to complete, a certain number of internal clock cycles are needed for proper sampling of the signal. Conversion time can vary, depending on the mode in which the TLV320AIC3268 is used. Throughout this document, internal and conversion clock cycles are used to describe the times that many functions take to execute. Considering the total system design, these times must be taken into account by the user.

The ADC uses either the internal MCLK signal or the internal oscillator for the SAR conversions.

8.3.6.2 Data Available - INT1 or INT2 Programmed as DATA_AVA

The interrupt signals INT1 and INT2 can be programmed by writing to B0_P0_R50_D[7:6] (INT1) or B0_P0_R50_D[5:4] (INT2). These terminals function as the DATA_AVA signal. To enable the SAR data available interrupt, B0_P3_R3_D[1:0] must be programmed to '01'. The DATA_AVA signal and interrupts INT1 and INT2 can be mapped to GPIO1, GPIO2, GPO1, DOUT1, WCLK2, BCLK2, or DOUT2.

8.3.6.3 Temperature Measurement

In some applications, such as battery charging, a measurement of ambient temperature is required. The temperature measurement technique used in the TLV320AIC3268 relies on the characteristics of a semiconductor junction operating at a fixed current level. The forward diode voltage (V_j) has a well-defined characteristic versus temperature. The ambient temperature can be predicted in applications by knowing the 25 \degree C value of the V_i voltage and then monitoring the variation of that voltage as the temperature changes.

The TLV320AIC3268 offers two modes of temperature measurement. The first mode requires a single reading to predict the ambient temperature. A diode, as shown in [Figure](#page-77-0) 66, is used during this measurement cycle. This voltage is typically 600 mV at 25°C with a 20μA current through it. The absolute value of this diode voltage can vary a few millivolts. The temperature coefficient of this voltage is typically 2 mV/°C. During the final test of the end product, the diode voltage at a known room temperature is stored in nonvolatile memory. Further calibration can be done to calculate the precise temperature coefficient of the particular device. This method has a temperature resolution of approximately 0.4°C/LSB and accuracy of approximately ±3°C with two-temperature calibration. [Figure](#page-78-0) 67 and [Figure](#page-78-1) 68 show typical plots with single and two-temperature calibration, respectively.

Figure 66. Functional Block Diagram of Temperature-Measurement Mode

Figure 67. Typical Plot of Single-Measurement Method After Calibrating for Offset at Room Temperature

Figure 68. Typical Plot of Single-Measurement Method After Calibrating for Offset and Gain at Two Temperatures

The second mode uses a two-measurement (differential) method. This mode requires a second conversion with a current 82 times larger. The voltage difference between the first (TEMP1) and second (TEMP2) conversion, using 82 times the bias current, is represented by:

$$
V_{(Temp1 - Temp2)} = \frac{kT}{q} \times ln(N)
$$

where

N is the current ratio $= 82$ k = Boltzmann's constant (1.38054 × 10[−]²³ electron volts/Kelvin) q = the electron charge (1.602189 \times 10⁻¹⁹ C)

 $T =$ the temperature in degrees Kelvin

The equation for the relation between differential code and temperature may vary slightly from device to device and can be calibrated at final system test by the user. This method provides resolution of approximately 2°C/LSB and accuracy of approximately ±6°C after calibrating at room temperature. A plot of typical calibration error for this method is shown in [Figure](#page-79-0) 69.

Figure 69. Typical Plot of Differential Measurement Method After Calibrating for Offset and Gain at Two Temperatures

The TLV320AIC3268 supports programmable auto-temperature measurement mode, which can be enabled by setting B0_P3_R19_D4. In this mode, the TLV320AIC3268 can auto-start the temperature measurement after a programmable interval. The user can program minimum and maximum threshold values through a register. In the case of temperature measurements, these thresholds are controlled in B0_P3_R30-R33. If the measurement goes outside the threshold range, the TLV320AIC3268 sets a flag in read-only B0_P3_R21, which is cleared after the flag is read. The TLV320AIC3268 can also be configured to send an active-high interrupt over INT1 or INT2 by setting bits in B0_P0_R50. The duration of the interrupt is approximately 2 ms, if B0_P0_R51_D[7:6] = '00' or B0_P0_R51_D[5:4] = '00', or these interrupt signals can be configured for alternate output signals. See Interrupt [Generation](#page-89-0) and Diagnostic Flags for more details on interrupt generation.

Temperature measurement can only be done in host-controlled mode.

8.3.6.4 Auxiliary Battery-Voltage Measurement for VBAT

The TLV320AIC3268 can be used to measure battery voltage up to 6V. This measurement can made using the VBAT terminal, which has a voltage divider (divide by 5), as seen in [Figure](#page-75-0) 64. This analog prescaler is available on the terminal to allow higher voltages to be measured by the SAR ADC. This battery measurement function is supported in 8-bit, 10-bit, and 12-bit modes.

To enable the battery-voltage measurement mode, write a '0110' to B0_P3_R3_D[5:2].

Because the ADC code is 1/5 of the actual voltage value applied at VBAT, the correct value can be found by multiplying the ADC code by 5. For low voltages of VREF_SAR, this function can support voltages from 0 to (5 x VREF_SAR), where the upper voltage limit for VBAT is 6V, and is also limited by the value listed in the *Absolute Maximum Ratings* table in the TLV320AIC3268 data sheet.

In the battery-voltage measurement mode, the conversion results in an ADC output code of B, where the voltage at the input terminal (VBAT) can be calculated as:

$$
V_{BAT} = \frac{B}{2^N} \times (5 \times VREF_SAR)
$$

(19)

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where:

N is the programmed resolution of the SAR ADC.

VREF SAR is the applied external reference voltage.

For an example of a script for battery voltages on VBAT, download the Example Scripts zip file from the TLV320AIC3268 product page.

8.3.6.5 Auxiliary Voltage Measurements

The TLV320AIC3268 can be used to measure voltage on IN1L/AUX1 and IN1R/AUX2 terminals. This voltage measurement function is supported in 8-bit, 10-bit, and 12-bit modes.

For IN1L/AUX1 and IN1R/AUX2:

If the conversion results in an ADC output code of B, then the voltage at the input terminals (IN1L/AUX1 and IN1R/AUX2) can be calculated as:

$$
V_{PIN} = \frac{B}{2^N} \times VREF_SAR
$$

(20)

where:

N is the programmed resolution of the SAR ADC.

VREF_SAR is the applied external reference voltage.

For an example of a script for reading voltages on IN1L, download the Example Scripts zip file from the TLV320AIC3268 product page.

The TLV320AIC3268 supports a special mode of conversion on IN1L/AUX1 and IN1R/AUX2 terminals called the resistance measurement mode. This mode is useful for measuring the value of an external resistance connected to this terminal. Conversion for IN1L/AUX1 input works in resistance measurement mode by writing '1' to B0_P3_R19_D2 and conversion for IN1R/AUX2 works in resistance measurement mode by writing '1' to B0_P3_R19_D1.

Figure 70. Resistance measurement mode for IN1L/AUX1 and IN1R/AUX2

The resistance measurement mode works in two ways. In Internal Bias Resistance measurement mode the value of external resistance R_{ext} is measured as a ratio with internal bias resistance which is nominally of 50kΩ value. The value of R_{ext} is calculated by [Equation](#page-80-0) 21. This value is useful to measure external resistance without any external components. The internal bias resistance will however exhibit device to device variation and should thus be used where a very high accuracy measurement is not required.

 R_{ext} =ADCOUT.50000/(2^N-ADCOUT)

- (1) Where N is the resolution of the SAR ADC
- (2) Where bias resistor is 50kΩ. The equation will scale for a different value of bias resistor. (21)

The resistance measurement mode can be also enabled to work with external bias resistor by writing '1' to B0_P3_R19_D0. When using this mode, an external bias resistor of 50kΩ should be connected between VREF_SAR and IN1L/AUX1 or IN1R/AUX2 terminals. This mode enables a higher accuracy conversion by using a higher accuracy bias resistance. The value of R_{ext} is determined by [Equation](#page-80-0) 21.

8.3.6.6 Auto Scan

If making voltage measurements on one or many of the inputs from IN1L/AUX1, IN1R/AUX2, VBAT and TEMP1 (or TEMP2) is desired on a continuous basis, then the auto-scan mode can be used. This mode causes the TLV320AIC3268 to sample and convert each of selected inputs in a repetetive manner with programmed intervals. The set of inputs that can be selected are programmed by configuring B0_P3_R19_D[7:4]. Auto scan can be set up by writing "1001" to B0_P3_R3_D[5:2]. Programming B0_P3_R15_D[3:0], allows a programmable interval delay to be introduced between successive conversions of the set of inputs selected for covnersion.

See *Conversion Time Calculations for the TLV320AIC3268*, [Host-Controlled](#page-85-0) Auto Scan Mode for conversion-time calculations and timing diagrams.

8.3.6.7 Port Scan

If making voltage measurements on the inputs IN1L/AUX1, IN1R/AUX2, and VBAT is desired on a periodic basis, then the port-scan mode can be used. This mode causes the TLV320AIC3268 to sample and convert each of the auxiliary inputs once. At the end of this cycle, all of the auxiliary result registers contain the updated values. Thus, with one write to the TLV320AIC3268, the host can cause three different measurements to be made. Port scan can be set up by writing "1011" to B0_P3_R3_D[5:2].

See *Conversion Time Calculations for the TLV320AIC3268*, *Port-Scan Operation*, [Port-Scan](#page-86-0) Operation, for conversion-time calculations and timing diagrams.

8.3.6.8 Buffer Mode

The TLV320AIC3268 supports a programmable buffer mode for all conversions (VBAT, IN1L/AUX1, IN1R/AUX2, TEMP1, TEMP2). Buffer mode is implemented using a circular FIFO with a depth of 64. The number of interrupts required to be serviced by a host processor can be reduced significantly in buffer mode. Buffer mode can be enabled using B0_P3_R13_D7.

Converted data is automatically written into the FIFO. To control the writing, reading and interrupt process, a write pointer (WRPTR), a read pointer (RDPTR), and a trigger pointer (TGPTR) are used. The read pointer always shows the location that is read next. The write pointer indicates the location in which the next converted data is to be written. The trigger pointer indicates the location at which an interrupt is generated if the write pointer reaches that location. Trigger level is the number of the data values needed to be present in the FIFO before generating an interrupt. [Figure](#page-81-0) 71 shows the case when the trigger level is programmed as 32. On resetting the buffer mode, RDPTR moves to location 1, WRPTR moves to location 1, and TGPTR moves to a location equal to the programmed trigger level.

The user can select the input or input sequence to be converted by writing to B0_P3_R3_D[5:2]. The converted values are written in a predefined sequence to the circular buffer. The user has flexibility to program a specific trigger level in order to choose the configuration which best fits the application. When the number of converted data values written in FIFO becomes equal to the programmed trigger level, then the device generates an interrupt signal on INT1 or INT2. In buffer mode, the user should program this terminal as Data Available. In buffer mode, conversions (VBAT, IN1L/AUX1, IN1R/AUX2, TEMP1, TEMP2) are allowed only in host-controlled mode.

Buffer mode can be used in single-shot conversion or continuous-conversion mode.

In single-shot conversion mode, once the number of data values written reaches the programmed trigger level, the TLV320AIC3268 generates an interrupt and waits for the user to start reading. As soon as the user starts reading the first data value from the last converted set, the TLV320AIC3268 clears the interrupt and starts a new set of conversions, and the trigger pointer is incremented by the programmed trigger level. An interrupt is generated again when the trigger condition is satisfied.

In continuous-conversion mode, once the number of data values written reaches the programmed trigger level, the TLV320AIC3268 generates an interrupt. It immediately starts a new set of conversions, and the trigger pointer is incremented by the programmed trigger level. An interrupt is cleared either by writing the next converted data value into the FIFO or by starting to read from the FIFO.

Depending on how the user is reading data, the FIFO can become empty or full. If the user is trying to read data even if the FIFO is empty, then RDPTR keeps pointing to same location. If the FIFO becomes full, then the next location is overwritten with newly converted data values, and the read pointer is incremented by one.

While reading the FIFO, the TLV320AIC3268 provides FIFO-empty and -full status flags along with the data. The user can also read a status flag from B0_P3_R13_D[1:0]. See [Table](#page-82-1) 33 for buffer-mode control and Table 34 for buffer-mode 16-bit read-data format.

(1) To enable buffer mode, write a 1 to B0_P3_R13_D7.

Table 34. Buffer Mode 16-Bit Read Data Format (B0_P252_R1 and B0_P252_R2)

Table 34. Buffer Mode 16-Bit Read Data Format (B0_P252_R1 and B0_P252_R2) (continued)

8.3.6.8.1 Buffer Mode Access through I ²C for TLV320AIC3268

To enable faster data access, SPI protocol is preferred, but if I^2C is required, note the following.

- In continuous buffer mode:
	- Only one measurement type, that is choice of IN1L, IN1R, VBAT, TEMP1 or TEMP2, can be used.
- In single-shot mode:
	- Multiple measurement types can be stored in the buffer consecutively.
	- The I²C read must completely empty the buffer. In other words, the number of bytes read must be equal to the trigger-level multiplied by 2 (for 2 bytes per converted data). If the buffer is empty, this will be reflected by bit B0_P252_R1_D6=1 in the last measurement read.
	- The I²C read must empty the buffer in a single call. Note that some I²C drivers may break auto-increment instructions into multiple, smaller calls. This can cause the SAR buffer to return invalid data, so the SAR trigger level must be less than or equal to the max I^2C auto-increment size divided by 2.
- If 64 elements (128 bytes) are read, the last byte will be invalid data since I^2C allows a maximum of 127 bytes.

8.3.6.9 Reading AUX Data in Non-Buffer Mode From SPI

Reading from the TLV320AIC3268 is done by using the protocol called out in [Figure](#page-83-0) 72.

Figure 72. 16-Bit Data-Read Timing, 24 Clocks per 16-Bit Data Read, 8-Bit Bus Interface

This protocol uses a 24-clock sequence to get a 16-bit data read. Set the INT1 or INT2 interrupt for monitoring the data-available status by writing '01' to B0_P3_R3_D[1:0]. Reading is normally done when the interrupt is low (data is available for reading). Status from the ADC conversion can be read from B0_P3_R9. If bit D6 is 0, then the ADC is actively converting, so a BUSY status is read. If bit D5 is set, then some data is now available for reading. Next, reading from a status register on B0_P3_R10 lets us know if data is available for IN1L/AUX1, IN1R/AUX2, or VBAT. If bit D7 is set, then IN1L/AUX1 data can be read. If bit D6 is set, then IN1R/AUX2 data can be read. If bit D5 is set, then VBAT data can be read.

The first 7 bits in the read sequence are for the first register address of the two sequential 8-bit registers. The next bit is high, which specifies that a read operation follows; then the 16 remaining clocks are used to get the 16-bit data that is read out in the order of D15–D0. The register address specified in the first seven clocks of the 24-clock sequence reads out as bits D15–D8, where D15 is the MSB of the byte, then the register number is incremented by 1 and the data is read from D7–D0, where D7 is the MSB of that byte. (For reading data for IN1L/AUX1, use B0 P3 R54 and B0 P3 R55; for reading data for IN1R/AUX2, use B0 P3 R56 and B0_P3_R57; and for reading data for VBAT, use B0_P3_R58 and B0_P3_R59.) From this cycle, the first 16-bit data word has been read. This sequence can be repeated to read further values of IN1L/AUX1, IN1R/AUX2, and VBAT data.

8.3.6.10 Auto Threshold Detect

The SAR ADC in TLV320AIC3268 has a special auto threshold detect feature where the device can detect a conversion exceeding the pre-programmed minimum and maximum thresholds for the input. Upon such a detection the TLV320AIC3268 can interrupt the host processor. This feature is useful as it allows the host to be immediately informed about out of range conditions, without the host having to poll for converted values. Upon receiving an interrupt, the host can read the flag register to find the input which exceeded the maximum or minimum threshold. The register settings for Auto Threshold Detect feature are described in [Table](#page-84-0) 35.

Table 35. Threshold Detect Settings

8.3.6.11 Conversion Time Calculations for the TLV320AIC3268

This section discusses conversion time calculations for temperature, auxiliary, and battery measurements for TLV320AIC3268.

The timing signals can be programmed by B0_P3_R3. INT1 or INT2 can be programmed as \overline{DATA} AVA by programming B0_P0_R50_D[7:4]. DATA_AVA can also be sent to GPIO1, GPIO2, GPO1, DOUT1, WCLK2, BCLK2, or DOUT2.

Table 36. Parameters for calculating Conversion Times

Table 36. Parameters for calculating Conversion Times (continued)

8.3.6.11.1 Host-Controlled Scan Mode

The time needed to make one single conversion for VBAT, IN1L/AUX1, IN1R/AUX2, TEMP1 (or TEMP2) is given by

 ${\mathsf T}_{\mathsf{cycle}}$ = ${\mathsf N}_{\mathsf{avg}}$.{(${\mathsf N}_{\mathsf{bits}}$ + 1). ${\mathsf t}_{\mathsf{conv}}$ + (${\mathsf n}_1$ + 13). ${\mathsf t}_{\mathsf{clk}}$ + ${\mathsf N}_{\mathsf{avg}}$. ${\mathsf n}_2. {\mathsf t}_{\mathsf{clk}}$ + (${\mathsf n}_3$ + 17). ${\mathsf t}_{\mathsf{clk}}$

- (1) This equation is valid if B0_P3_R18_D[6:5] = 00, which means SAR data update is not kept on hold for reading converted data.
- (2) The programmable delay t_{REF} scales accordingly based on the actual divider setting and time period of the clock used to generate this. See the respective control register settings to understand the scale factors.

8.3.6.11.2 Host-Controlled Auto Scan Mode

The time needed for one cycle of Auto Scan Mode is given by:

- ${\sf T}_{\sf cycle}{=}{\sf N}_{\sf inp}.{\sf N}_{\sf avg}.(({\sf N}_{\sf bits}+1).{\sf t}_{\sf conv} + ({\sf n}_1+13).{\sf t}_{\sf clk}) + {\sf N}_{\sf avg}.{\sf n}_2.{\sf t}_{\sf clk} + {\sf N}_{\sf inp}.9. {\sf t}_{\sf clk} + ({\sf n}_3+{\sf n}_4).{\sf t}_{\sf clk}+{\sf t}_{\sf del}$
	- (1) This equation is valid if B0_P3_R18_D[6:5] = 00, which means SAR data update is not kept on hold for reading converted data.
	- (2) This equation is valid only from the second conversion onward.

(3) All the programmable delays, t_{DEL} and t_{REF} , scale accordingly based on the actual divider setting and time period of the clock used to generate this. See the respective control register settings to understand the scale factors.

Figure 74. Host-Controlled Auto Scan Mode

8.3.6.11.3 Port-Scan Operation

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The time needed to complete one set of Port-Scan conversions is given by:

 $T_{\text{cycle}} = 3.N_{\text{avg}}.\{(N_{\text{bits}} + 1).t_{\text{conv}} + (n_1 + 13).t_{\text{clk}} + (n_3 + 35).t_{\text{clk}}\}$

- (1) This equation is valid if B0_P3_R18_D[6:5] = 00, which means SAR data update is not kept on hold for reading converted data.
- (2) The programmable delay t_{REF} scales based on the actual divider setting and time period of the clock used to generate this. See the respective control register settings to understand the scale factors.

Figure 75. Host-Controlled Port Scan Mode

8.3.7 Headset Detection

The TLV320AIC3268 includes extensive capability to monitor a headphone, microphone, or headset jack to find if a plug has been inserted into the jack, and if inserted find the nature of headset or headphone inserted in the jack. The device also includes the capability to detect a button press for actions such as starting a call with headset button press. The figures below show the circuit configuration to enable this feature for stereo headphones and stereo headset with microphone and button, as well as mono headset with and without microphone. It is recommended to use IN1L or IN1R for external headset microphones.

Figure 76. Jack Connections for Detection of Stereo Headsets

Figure 77. Jack Connections for Detection of Mono Headsets

This feature is enabled by programming B0_P0_R67_D7. In order to avoid false detections due to mechanical vibrations in headset jacks or microphone buttons, a debounce function is provided for glitch rejection. For the case of headset insertion/removal, a debounce function with a range of 16ms - 512ms is provided. This can be programmed via B0_P0_R67_D[4:2]. For improved button-press detection, the debounce function has a range of 8ms to 32ms by programming B0_P0_R67_D[1:0].

The TLV320AIC3268 also provides feedback to user when a button press, or a headset insertion/removal event is detected through register readable flags as well as an interrupt on the IO terminals. The value in B0_P0_R46_D[5:4] provides the instantaneous state of button press and headset insertion. B0_P0_R44_D5 is a sticky (latched) flag that is set when the button-press event is detected. B0_P0_R44_D4 is a sticky flag that is set when the headset insertion or removal event is detected. These sticky flags are set by the event occurrence, and are reset only when read. This requires polling B0_P0_R44. To avoid polling and the associated overhead, the TLV320AIC3268 also provides an interrupt feature where the events can trigger the INT1 and/or INT2 interrupts. These interrupt events can be routed to one of the digital output terminals. Please see Interrupt [Generation](#page-89-0) and [Diagnostic](#page-89-0) Flags for details on interrupts (INT1 and INT2) and Interrupt [Generation](#page-89-0) and Diagnostic Flags for details on digital terminal routing.

As shown in [Figure](#page-87-0) 76 and [Figure](#page-87-1) 77, the TLV320AIC3268 not only detects a headset insertion event, but also distinguishes between the different headsets inserted, such as stereo headphones, stereo cellular headsets with microphone, mono headsets with microphone, and mono headset without microphone. After the headsetdetection event, the user can read B0_P0_R37_D[5:4] and B0_P0_R37_D[1:0] to determine the type of headset inserted.

Headset Type	Microphone Detection	Headset Detection
Stereo Headphones without Microphone	$B0_P0_R37_D[5:4] = 01$	B0 P0 R37 D[1:0] = 10
Stereo Headset with Microphone	B0 P0 R37 D[5:4] = 11	B0 P0 R37 D[1:0] = 10
Mono Headset without Microphone	$B0_P0_R37_D[5:4] = 01$	$B0 \cdot P0 \cdot R37 \cdot D[1:0] = 01$
Mono Headset with Microphone	$B0_P0_R37_D[5:4] = 11$	$B0 \cdot P0 \cdot R37 \cdot D[1:0] = 01$

Table 37. Headset Detection Types

For proper detection of these different types, it is important to follow the guidelines in [Table](#page-88-0) 38.

Table 38. Detection Specifications for Microphone, Button, Headset

The headset detection block requires AVDDx_18 and AVDD3_33 to be powered. In addition, the weak connection of AVDDx_18 to DVDD_18 should be disabled (B0_P1_R1_D3="0"), and External Analog Supplies should be enabled (B0_P1_R1_D2="0"). The headset detection feature in the TLV320AIC3268 is achieved with a very low power overhead, requiring less than 30μA of additional current from AVDDx_18 supplies.

8.3.8 Interrupt Generation and Diagnostic Flags

The TLV320AIC3268 an trigger interrupts to the host processor for events that require host processor intervention. This avoids polling the status-flag registers continuously. The TLV320AIC3268 has two defined interrupts; INT1 and INT2 that can be configured by programming Page 0, Register 48 and 49. A user can configure the interrupts INT1 and INT2 to be triggered by one or many events such as:

- Headset Detection
- **Button Press**
- Noise Detected by AGC
- Over-current Condition in Headphones
- Data Overflow in ADC and DAC Processing Blocks and Filters
- Over-temperature Condition in Speaker Drivers
- SAR ADC Data Available or Exceeding Threshold

Each of these INT1 and INT2 interrupts can be routed to output terminals like GPIO1, GPIO2, GPO1, DOUT1, WCLK2, BCLK2, and DOUT2 by configuring B0_P4_R67-R96. [Table](#page-89-1) 40 displays how to individually configure the INT1 or INT2 interrupts.

Table 40. Register Settings for Interrupt Routing

These interrupt signals can either be configured as a single pulse, a series of pulses, or a change in output level by programming B0_P0_R51_D[7:6] and B0_P0_R51_D[5:4]. If the user configures the interrupts as a series of pulses, the events will trigger the start of pulses that will stop when the flag registers in B0_P0_R42, B0_P0_R44, B0_P0_R45 are read by the user to determine the cause of the interrupt. Similarly, if the user configures the interrupts as an active-high, level-based interrupt generated from these sticky flags, the interrupt port will reset low when the flag registers in B0_P0_R42, B0_P0_R44, B0_P0_R45 are read by the user.

When the interrupts are configured for multiple events simultaneously, the user can read associated flags to determine the triggering events. Sometimes upon reading the Primary Flag Registers, additional Secondary Flag Registers may have to be read back to resolve the triggering event. For example, on receiving an interrupt, the user can read Primary Flag Registers B0_P0_R44 and B0_P0_R45. If B0_P0_R44_D[7] is '1', it indicates an over-current condition on one of HPL, RECP or SPK drivers has happened. To determine which of the three drivers had an over-current condition, the Secondary Flag Register B0_P1_R69 should be read back. [Table](#page-89-2) 41 provides details of Primary Flag Registers and Secondary Flag Registers which can be used to resolve events causing the interrupts.

Table 41. Flags for Interrupt

Table 41. Flags for Interrupt (continued)

In addition to the interrupt flags, the TLV320AIC3268 features additional status flags which are very helpful for diagnostics since they enable reporting of the status of various internal blocks of the device. [Table](#page-90-0) 42 provides list of the other flags available in TLV320AIC3268.

Table 42. Miscellaneous Flags

Table 42. Miscellaneous Flags (continued)

Sticky Flags are useful for reporting events which could be intermittent. These flags are set by the triggering events like over-current detect but are reset only when the flag register is read back the user. Status Flags on the other hand are useful for reporting events which are steady state in nature like power-up status of a block. The flag value reflects the triggering event's status when the register is being read by the user.

8.3.9 Interfaces

8.3.9.1 Control Interfaces

The TLV320AIC3268 control interface supports SPI or I²C communication protocols. For SPI, the SPI_SELECT terminal should be tied high; for I²C, SPI_SELECT should be tied low. It is not recommended to change the state of SPI_SELECT during device operation.

8.3.9.1.1 I ²C Control Mode

The TLV320AIC3268 supports the I²C control protocol, and will respond by default (I2C_ADDR_SCLK grounded) to the 7-bit I²C address of 0011000. With the one I²C address terminal, I2C_ADDR_SCLK, the device can be configured to respond to one of two 7-bit I²C addresses, 0011000 or 0011001. The full 8-bit I²C address can be calculated as:

8-Bit I ²C Address = "001100" + I2C_ADDR_SCLK + R/W

Example: to write to the TLV320AIC3268 with I2C_ADDR_SCLK = 1 the 8-Bit I^2C Address is "001100" + I2C_ADDR_SCLK + R/W = "00110010" = 0x32

¹²C is a two-wire, open-drain interface supporting multiple devices and masters on a single bus. Devices on the ²C bus only drive the bus lines LOW by connecting them to ground; they never drive the bus lines HIGH. Instead, the bus wires are pulled HIGH by pullup resistors, so the bus wires are HIGH when no device is driving them LOW. This way, two devices cannot conflict; if two devices drive the bus simultaneously, there is no driver contention.

Communication on the I²C bus always takes place between two devices, one acting as the master and the other acting as the slave. Both masters and slaves can read and write, but slaves can only do so under the direction of the master. Some I²C devices can act as masters or slaves, but the TLV320AIC3268 can only act as a slave device.

An I²C bus consists of two lines, SDA and SCL. SDA carries data, and the SCL signal provides the clock. All data is transmitted across the I²C bus in groups of eight bits. To send a bit on the I²C bus, the SDA line is driven to the appropriate level while SCL is LOW (a LOW on SDA indicates the bit is zero, while a HIGH indicates the bit is one).

Once the SDA line has settled, the SCL line is brought HIGH, then LOW. This pulse on the SCL line clocks the SDA bit into the receiver's shift register.

The I²C bus is bidirectional: the SDA line is used both for transmitting and receiving data. When a master reads from a slave, the slave drives the data line; when a master sends to a slave, the master drives the data line.

Most of the time the bus is idle, no communication is taking place, and both lines are HIGH. When communication is taking place, the bus is active. Only master devices can start communication on the bus. Normally, the data line is only allowed to change state while the clock line is LOW. If the data line changes state while the clock line is HIGH, it is either a START condition or its counterpart, a STOP condition. A START condition is when the clock line is HIGH and the data line goes from HIGH to LOW. A STOP condition is when the clock line is HIGH and the data line goes from LOW to HIGH.

After the master issues a START condition, it sends a byte that selects the slave device for communication. This byte is called the address byte. Each device on an $I^2\dot{C}$ bus has a unique 7-bit address to which it responds. (Slaves can also have 10-bit addresses; see the I²C specification for details.) The master sends an address in the address byte, together with a bit that indicates whether it wishes to read from or write to the slave device.

Every byte transmitted on the I²C bus, whether it is address or data, is acknowledged with an acknowledge bit. When a master has finished sending a byte (eight data bits) to a slave, it stops driving SDA and waits for the slave to acknowledge the byte. The slave acknowledges the byte by pulling SDA LOW. The master then sends a clock pulse to clock the acknowledge bit. Similarly, when a master has finished reading a byte, it pulls SDA LOW to acknowledge this to the slave. It then sends a clock pulse to clock the bit. (Remember that the master always drives the clock line.)

A not-acknowledge is performed by simply leaving SDA HIGH during an acknowledge cycle. If a device is not present on the bus, and the master attempts to address it, it will receive a not−acknowledge because no device is present at that address to pull the line LOW.

When a master has finished communicating with a slave, it may issue a STOP condition. When a STOP condition is issued, the bus becomes idle again. A master may also issue another START condition. When a START condition is issued while the bus is active, it is called a repeated START condition.

The TLV320AIC3268 can also respond to and acknowledge a General Call, which consists of the master issuing a command with a slave address byte of 00H. This feature is disabled by default, but can be enabled via B0_P0_R115_D5.

In the case of an $I²C$ register write, if the master does not issue a STOP condition, then the device enters autoincrement mode. So in the next eight clocks, the data on SDA is treated as data for the next incremental register.

Similarly, in the case of an I²C register read, after the device has sent out the 8-bit data from the addressed register, if the master issues a ACKNOWLEDGE, the slave takes over control of SDA bus and transmit for the next 8 clocks the data of the next incremental register.

8.3.9.1.2 SPI Digital Interface

In the SPI control mode, the TLV320AIC3268 uses the terminals SCL_SSZ as \overline{SS} , I2C_ADDR_SCLK as SCLK, MISO GPO1 as MISO, SDA MOSI as MOSI; a standard SPI port with clock polarity setting of 0 (typical microprocessor SPI control bit CPOL = 0) and clock phase setting of 1 (typical microprocessor SPI control bit CPHA = 1). The SPI port allows full-duplex, synchronous, serial communication between a host processor (the master) and peripheral devices (slaves). The SPI master (in this case, the host processor) generates the synchronizing clock (driven onto SCLK) and initiates transmissions. The SPI slave devices (such as the TLV320AIC3268) depend on a master to start and synchronize transmissions. A transmission begins when initiated by an SPI master. The byte from the SPI master begins shifting in on the slave MOSI terminal under the control of the master serial clock (driven onto SCLK). As the byte shifts in on the MOSI terminal, a byte shifts out on the MISO terminal to the master shift register.

The TLV320AIC3268 interface is designed so that with a clock-phase bit setting of 1 (typical microprocessor SPI control bit CPHA = 1), the master begins driving its MOSI terminal and the slave begins driving its MISO terminal on the first serial clock edge. The SSZ terminal can remain low between transmissions; however, the TLV320AIC3268 only interprets the first 8 bits transmitted after the falling edge of SSZ as a command byte, and the next 8 bits as a data byte only if writing to a register. Reserved register bits should be written to their default values. The TLV320AIC3268 is entirely controlled by registers. Reading and writing these registers is accomplished by an 8-bit command sent to the MOSI terminal of the part prior to the data for that register. The command is structured as shown in [Table](#page-94-0) 43. The first 7 bits specify the address of the register which is being written or read, from 0 to 127 (decimal). The command word ends with an R/W bit, which specifies the direction

of data flow on the serial bus. In the case of a register write, the R/W bit should be set to 0. A second byte of data is sent to the MOSI terminal and contains the data to be written to the register. Reading of registers is accomplished in a similar fashion. The 8-bit command word sends the 7-bit register address, followed by the R/W bit = 1 to signify a register read is occurring. The 8-bit register data is then clocked out of the part on the MISO terminal during the second 8 SCLK clocks in the frame.

8.3.9.2 Digital Audio Interfaces

The TLV320AIC3268 features three digital audio data serial interfaces, or audio buses. These three interfaces can be run simultaneously, thereby enabling reception and transmission of digital audio for or to three separate devices. A common example of usage of multiple digital audio serial interfaces is to allow connections with application processor, bluetooth chipset, digital input Class-D amplifiers and so forth. By utilizing the TLV320AIC3268 as the center of the audio processing in a portable audio system, mixing of voice and music audio is greatly simplified. In addition, the miniDSP can be utilized to greatly enhance the portable device experience by providing advanced audio processing to both communication and media audio streams simultaneously.

Figure 82. Typical Multiple Connections to Three Audio Serial Interfaces

Each audio bus on the TLV320AIC3268 is very flexible, including left or right-justified data options, support for ¹²S or PCM protocols, programmable data length options, a TDM mode for multichannel operation, very flexible master or slave configurability for each bus clock line, and the ability to communicate with multiple devices within a system directly.

Each of the three audio buses of the TLV320AIC3268 can be configured for left or right-justified, I²S, DSP, or TDM modes of operation, where communication with PCM interfaces is supported within the TDM mode. These modes are all MSB-first, with data width programmable as 16, 20, 24, or 32 bits. In addition, the word clock and bit clock can be independently configured in either Master or Slave mode, for flexible connectivity to a wide variety of processors. The word clock is used to define the beginning of a frame, and may be programmed as either a pulse or a square-wave signal. The frequency of this clock corresponds to the maximum of the selected ADC and DAC sampling frequencies. When configuring an audio interface for six-wire mode, the ADC and DAC paths can operate based on separate word clocks.

The bit clock is used to clock in and clock out the digital audio data across the serial bus. When in Master mode, this signal can be programmed to generate variable clock pulses by controlling the bit-clock divider. The number of bit-clock pulses in a frame may need adjustment to accommodate various word-lengths as well as to support the case when multiple TLV320AIC3268s may share the same audio bus. When configuring an audio interface for six-wire mode, the ADC and DAC paths can operate based on separate bit clocks.

The TLV320AIC3268 also includes a feature to offset the position of start of data transfer with respect to the word-clock. This offset can be controlled in terms of number of bit-clocks.

The TLV320AIC3268 also has the feature of inverting the polarity of the bit-clock used for transferring the audio data as compared to the default clock polarity used. This feature can be used independently of the mode of audio interface chosen.

The TLV320AIC3268 further includes programmability to 3-state the DOUT line during all bit clocks when valid data is not being sent. By combining this capability with the ability to program at what bit clock in a frame the audio data begins, time-division multiplexing (TDM) can be accomplished, enabling the use of multiple codecs on a single audio serial data bus. When the audio serial data bus is powered down while configured in master mode, the terminals associated with the interface are put into a 3-state output condition.

By default, when the word-clocks and bit-clocks are generated by the TLV320AIC3268, these clocks are active only when the codec (ADC, DAC or both) are powered up within the device. This is done to save power. However, it also supports a feature when both the word clocks and bit-clocks can be active even when the codec is powered down. This is useful when using the TDM mode with multiple codecs on the same bus, or when wordclock or bit-clocks are used in the system as general-purpose clocks.

The TLV320AIC3268 contains advanced Digital Audio interfaces features to enable:

- Connections of Multiple Digital Audio interfaces
- 6-wire Digital Audio interfaces for separate uplink/downlink clocks or ADC/DAC clocks

• Multi-channel, Multiple terminal operation

8.3.9.2.1 Connecting Multiple Audio Digital Interfaces

The TLV320AIC3268 enables connections to multiple audio data buses. [Figure](#page-95-0) 82 shows a typical example of utilizing the digital terminals on the device to connect to four separate 4-wire digital audio buses, with up to three of these 4-wire buses receiving and sending digital audio data simultaneously. This configuration can be utilized when using I²C for control of the device. If only 3 total audio interface connections are needed (that is, a fourth audio bus does not need to be muxed into Audio Serial Interface 1), either I²C or SPI control can be used. (Further details on SPI control and terminals utilized can be found in SPI Digital [Interface](#page-93-0) and [Table](#page-112-0) 47.)

To configure each of the three audio serial interfaces, both the audio interface and the terminals should be set up for appropriate routing of the signals. Audio Serial Interface 1 configuration registers are located in B0_P4_R1- R16 and B0_P4_R49-R52. Audio Serial Interface 2 configuration registers are located in B0_P4_R17-R32 and B0_P4_R53-R54. Audio Serial Interface 3 configuration registers are located in B0_P4_R33-R48 and B0_P4_R55-R56. The terminal muxing registers are located in B0_P4_R65-R96. [Table](#page-96-0) 44 displays the appropriate register settings needed to implement the Audio Serial Interface configuration found in [Figure](#page-95-0) 82.

Table 44. Register Settings for Typical Multiple Audio Digital Interface Connections

Table 44. Register Settings for Typical Multiple Audio Digital Interface Connections (continued)

Since each interface can be configured separately as master or slave, the appropriate settings are displayed for both possible configurations for each of the three audio serial interfaces. When in master mode, the bit clock and work clock source can be derived from a variety of sources, and more details on the possible sources of these clocks can be found in the Clock Generation and PLL section.

8.3.9.2.2 Six-Wire Digital Audio Interface

The six-wire audio interface mode allows independent configuration of receive and transmist word and bit clocks for the device. The TLV320AIC3268 supports six-wire audio interface on ASI1 and ASI2. The six-wire inteface mode is available in all interface formats such as I2S, LJF, RJF, DSP and PCM modes. Due to terminal limitations, only one of ASI1 or ASI2 can operate in six-wire mode at a time.

When ASI1 operates in six-wire mode, WCLK1 functions as DAC WCLK or receive word clock, BCLK1 functions as DAC_BCLK or receive bit clock. The tranmist word clock and transmit bit clock function can get supported on any of the pair of terminals amongst GPIO1, GPIO2, GPIO3 or GPIO4.

Figure 83. Six-Wire Audio Serial Interface with ASI1

When ASI2 operates in six-wire mode, WCLK1 functions as DAC_WCLK or receive word clock, BCLK1 functions as DAC_BCLK or receive bit clock. The tranmist word clock and transmit bit clock function can get supported on any of the pair of terminals amongst GPIO1, GPIO2, GPIO3 or GPIO4.

The details of register settings to enable six-wire interface mode are shown in [Table](#page-118-0) 51.

8.3.9.2.3 Multiple Channel, Multiple Terminal Setup

The TLV320AIC3268 also enables connections of up to four stereo pairs (8 total channels) of input and output data on Audio Serial Interface 1. These eight bidirectional channels are all synchronized to a single word clock (WCLK1) and bit clock (BCLK1). [Figure](#page-98-0) 85 displays a typical configuration for this multi-channel setup.

Figure 85. Multi-channel, Multi-Terminal Inputs and Outputs to Audio Serial Interface 1

(1) BCLK2, WCLK2, DIN2 and DOUT2 terminals are with respect to IOVDD2 supply therefore using with ASI-1 in multiterminal 8-ch mode requires either shorting IOVDD1 and IOVDD2 with same voltage level or putting on-board level-shifter for the above signals.

The configuration shown in [Figure](#page-98-0) 85 can be used with either I²C or SPI for control interface. In this configuration ASI2 and ASI3 cannot be used. Other combinations with reduced channels or I²C only control interface allow simultaneous use of ASI2 or ASI3 by choosing different set of terminals. For details of various functions supported on terminals, see [Table](#page-112-0) 47.

8.3.9.2.4 Audio Formats

Each Audio Serial Interface supports left or right-justified, I²S, DSP, or mono PCM modes. In addition, timedivision multiplexing (TDM) can be implemented in each of these formats to enable multi-channel operation.

8.3.9.2.4.1 Right Justified Mode

Audio Serial Interface 1 can be put into Right Justified Mode by programming B0_P4_R1_D[7:5] = 010. Audio Serial Interface 2 can be put into Right Justified Mode by programming B0_P4_R17_D[7:5] = 010. Audio Serial Interface 3 can be put into Right Justified Mode by programming B0_P4_R33_D[7:5] = 010. In right-justified mode, the LSB of the left channel is valid on the rising edge of the bit clock preceding the falling edge of the word clock. Similarly, the LSB of the right channel is valid on the rising edge of the bit clock preceding the rising edge of the word clock.

Figure 86. Timing Diagram for Right-Justified Mode

For Right-Justified mode, the number of bit-clocks per frame should be greater than twice the programmed wordlength of the data.

8.3.9.2.4.2 Left Justified Mode

Audio Serial Interface 1 can be put into Left Justified Mode by programming B0_P4_R1_D[7:5] = 011. Audio Serial Interface 2 can be put into Left Justified Mode by programming B0_P4_R17_D[7:5] = 011. Audio Serial Interface 3 can be put into Left Justified Mode by programming B0_P4_R33_D[7:5] = 011. In left-justified mode, the MSB of the right channel is valid on the rising edge of the bit clock following the falling edge of the word clock. Similarly the MSB of the left channel is valid on the rising edge of the bit clock following the rising edge of the word clock.

Figure 87. Timing Diagram for Left-Justified Mode

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Figure 89. Timing Diagram for Left-Justified Mode with Offset=0 and Inverted Bit Clock

For Left-Justified mode, the programmed offset value should be less than the number of bit-clocks per frame by at least the programmed word-length of the data.

8.3.9.2.4.3 I ²S Mode

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Audio Serial Interface 1 can be put into 1^2S Mode by programming B0_P4_R1_D[7:5] = 000. Audio Serial Interface 2 can be put into I²S Mode by programming B0_P4_R17_D[7:5] = 000. Audio Serial Interface 3 can be put into I²S Mode by programming B0_P4_R33_D[7:5] = 000. In I²S mode, the MSB of the left channel is valid on the second rising edge of the bit clock after the falling edge of the word clock. Similarly the MSB of the right channel is valid on the second rising edge of the bit clock after the rising edge of the word clock.

[TLV320AIC3268](http://www.ti.com/product/tlv320aic3268?qgpn=tlv320aic3268)

Figure 92. Timing Diagram for I ²S Mode with Offset=0 and Bit Clock Inverted

For I²S mode, the programmed offset value should be less than the number of bit-clocks per frame by at least the programmed word-length of the data.

8.3.9.2.4.4 DSP Mode

Audio Serial Interface 1 can be put into DSP Mode by programming B0_P4_R1_D[7:5] = 001. Audio Serial Interface 2 can be put into DSP Mode by programming $B0_P4_R17_D[7:5] = 001$. Audio Serial Interface 3 can be put into DSP Mode by programming $B0P4-R33D[7:5] = 001$. In DSP mode, the rising edge of the word clock starts the data transfer with the left channel data first and immediately followed by the right channel data. Each data bit is valid on the falling edge of the bit clock.

Figure 93. Timing Diagram for DSP Mode

Figure 95. Timing Diagram for DSP Mode with Offset = 0 and Bit Clock Inverted

For DSP mode, the number of bit-clocks per frame should be greater than twice the programmed word-length of the data. Also the programmed offset value should be less than the number of bit-clocks per frame by at least the programmed word-length of the data.

8.3.9.2.4.5 Mono PCM Mode

Audio Serial Interface 1 can be put into Mono PCM Mode by programming B0_P4_R1_D[7:5] = 100. Audio Serial Interface 2 can be put into DSP Mode by programming B0_P4_R17_D[7:5] = 100. Audio Serial Interface 3 can be put into DSP Mode by programming B0_P4_R33_D[7:5] = 100. In mono PCM mode, the rising edge of the word clock starts the data transfer of the single channel of data. Each data bit is valid on the falling edge of the bit clock.

Figure 96. Timing Diagram for Mono PCM Mode

Figure 97. Timing Diagram for Mono PCM Mode with Offset=2

Figure 98. Timing Diagram for Mono PCM Mode with Offset=2 and Bit Clock Inverted

For mono PCM mode, the programmed offset value should be less than the number of bit-clocks per frame by at least the programmed word-length of the data.

8.3.9.2.5 Multi-channel Configurations

The TLV320AIC3268 can utilize TDM techniques to enable several multi-channel system scenarios. First, multiple codecs can transmit/receive on a single digital audio interface bus. Second, multiple stereo pairs can be sent and received by a single TLV320AIC3268 on a single 4-wire digital audio interface bus. Lastly, up to 4 individual stereo data pairs to/from the TLV320AIC3268 can be routed to individual DIN and DOUT lines in the system which are synchronized to a single BCLK and WCLK.

8.3.9.2.5.1 Single Host, Multiple Audio Codecs

Using the offset programmability and the DOUT line 3-state feature, the TLV320AIC3268 enables the flexibility where multiple TLV320AIC3268 devices can be interfaced together and can communicate to a host/multimedia processor using a single digital audio serial interface. [Figure](#page-103-0) 99 displays a typical configuration where M devices are connected to a single host processor.

Figure 99. Interfacing Multiple TLV320AIC3268 Devices Using Single I ²S Interface

By changing the programmable offset for each device, the bit clock in each frame where the data begins can be changed, and the serial data output driver (DOUT) also can be programmed to a 3-state mode during all bit clocks except when valid data is being put onto the bus. This allows other codecs to be programmed with different offsets and to drive their data onto the same DOUT line, just in a different slot. For incoming data, the codec simply ignores data on the bus except where it is expected based on the programmed offset.

Figure 100. DSP Timing for Multiple Devices Interfaced Together, Sequential Left/Right Pairs

The digital audio serial interface timing diagram for the interface in [Figure](#page-103-0) 99 is shown in [Figure](#page-104-0) 100. In this particular configuration, the TLV320AIC3268 (or any other TLV320AIC32x codec) is programmed for DSP mode with N-bit word length per channel. The offset programmed for the Codec-1 is 0, for Codec-2 it is 2N, and likewise, the offset programmed for the Codec-M is (M-1) x 2N. In this TDM mode, the number of bit-clocks per frame should be greater than M*2N. The TLV320AIC3268 allows a maximum offset of 255 bit clocks, and this enables connections of up to 4 codecs for 32-bit stereo data and 8 codecs for 16-bit stereo data.

For each of the three individual Digital Audio interfaces, this offset controls when data is received and sent by these interfaces. For Audio Serial Interface 1, this offset can be set to a value in the range of 0 to 255 bit clocks by programming B0_P4_R2. For Audio Serial Interface 2, this offset can be set to a value in the range of 0 to 255 bit clocks by programming B0_P4_R18. For Audio Serial Interface 3, this offset can be set to a value in the range of 0 to 255 bit clocks by programming B0_P4_R34. When utilized in DSP mode, each of these offsets will set the start of the left channel, with the right channel data immediately following the LSB of the left channel.

8.3.9.2.5.1.1 Time Slot Mode

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In addition, Audio Serial Interface 1 can also control the offset of the right channel with respect to the end of the left channel of data. This is achieved by enabling Time Slot Mode (setting B0_P4_R8_D0) and configuring the Right Channel Offset 2 (in the range of 0 to 255 bit clocks) in B0_P4_R3. Thus, the Right Channel Offset 2 control allows us to place the right channel anywhere in the frame after the left channel, and this functionality can be utilized in each of the audio formats (DSP, left or right-justified, or l²S).

Figure 101. DSP Timing for Multiple Devices Interfaced Together, Grouped Left Channels and Right Channels

By utilizing Time Slot Mode, the individual left and right channels can be grouped together, as shown in [Figure](#page-104-1) 101. Assuming each channel contains N bits in this example, Codec-1 would have an offset1=0 and offset2=M*N, Codec-2 would have an offset1=N and offset2=M*N, and likewise, Codec-M would have an offset1=(M-1)*N and offset2=M*N.

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8.3.9.2.5.2 Multiple Channel Operation, Single Data Lines (Audio Serial Interface 1)

The TLV320AIC3268 can receive and send multiple stereo pairs on a single 4-wire digital audio interface bus. This particularly useful when sending multi-channel audio data to the miniDSP for stereo downmix and playback over the integrated stereo headphones, speakers, or line-outs. Alternatively, the host could utilize the audio miniDSP engine as a multi-channel audio co-processor. This multi-channel operation is only available on Audio Serial Interface 1, and this is enabled by increasing the number of available channels in B0_P4_R4_D[7:6] to greater than 1 stereo pair. By increasing the number of stereo pairs, the interface essentially lengthens the data length for each channel. Thus, the first half of the X channels are interpreted as Left Channels, while the second half of X channels are interpreted as Right Channels. Once these channels are inside the miniDSP, they can be interpreted as any channel for surround processing. [Figure](#page-105-0) 102 shows the timing for X channels of data utilizing DOUT1 and DIN1 data lines.

Figure 102. DSP Timing for Multi-channel Mode, Single DOUT and DIN Lines

Because Audio Serial Interface 1 interprets the first X/2 channels as Left data, the last X/2 "Right" channels can be shifted utilizing Time Slot Mode. [Figure](#page-105-1) 103 shows how, in DSP mode, the start of the first X/2 channels can be delayed by one bit clock (by setting offset1=1 in B0_P4_R2), while the last X/2 "Right" channels can be delayed by two bit clocks after the end of the first X/2 "Left" channels.

For this multi-channel DSP mode, the number of bit-clocks per frame should be greater than M times the programmed word-length of the data, where M is the total number of channels set in B0_P4_R4_D[7:6]. Also the sum of the two programmed offset values should be less than the number of bit-clocks per frame by at least M times the programmed word-length of the data.

Figure 103. DSP Timing for Multi-channel Mode, Single DOUT and DIN Lines

8.3.9.2.5.3 Multiple Channel Operation, Multiple Data Lines (Audio Serial Interface 1)

The TLV320AIC3268 can receive or send up to 4 individual stereo data pairs can be routed to individual DIN and DOUT lines in the system which are synchronized to a single BCLK and WCLK. This multi-channel, multiterminal operation is only available on Audio Serial Interface 1. The multi-terminal mode is enabled by setting B0_P4_R6_D7 to 1. In addition to routing the channels to/from the interface, the individual terminals also need to be configured (refer to [Table](#page-112-0) 46 and Table 47 for possible digital terminal muxing setups). Just as in the multichannel, single-terminal case, the audio serial interface should configure the appropriate number of channels by writing to B0_P4_R4_D[7:6]. [Figure](#page-106-0) 104 shows an example of multi-channel, multi-terminal mode using 4 stereo data pairs (8 channels) in DSP format.

Figure 104. DSP Timing for Multi-channel Mode, Four Data Lines

For this multi-channel DSP mode, the number of bit-clocks per frame should be greater than twice the programmed word-length of the data. Also, any programmed offset1 value (for shift of start of left channel) should be less than the number of bit-clocks per frame by at least twice the programmed word-length of the data.

Figure 105. DSP Timing for Multi-channel Mode, Time Slot Mode, Four Data Lines with Offset1=1 and Offset2=2

By enabling Time Slot Mode, the start of the left and right channels on each data line can be controlled by offset1 and offset2. In other words, offset1 would control the start of all four left channels in [Figure](#page-107-0) 105, and offset2 would delay the start of all right channels after the end of the left channels' LSB. For this multi-channel, multiterminal DSP mode, the number of bit-clocks per frame should be greater than twice the programmed wordlength of the data. Also the sum of the two programmed offset values should be less than the number of bitclocks per frame by at least twice the programmed word-length of the data.

П $\vert \ \vert$ **WORD LEFT CHANNELS RIGHT CHANNELS CLOCK (1, 3, 5, 7) (2, 4, 6, 8)** $| \ |$ **BIT CLOCK DATA N N N N N N N N N IN 1, 2 1 0** $3 \mid 2 \mid 1 \mid 0$ **3 - - - - - - - - - 3 DATA 3 3 2 2 1 2 1 1 3 OUT 1 RD1(n)** $LD_1(n)$ **d** $RD_1(n)$ **d** $Q_1(n+1)$ **DATA N N N N N N N N N IN 2, 2 1 0** $3 \mid 2 \mid 1 \mid 0$ **3 - 3 - -** $\overline{}$ **- - - - - - DATA 1 2 3 1 2 3 2 3 1 OUT 2 LD2(n)** $RD_2(n)$ **LD**₂ $(n+1)$ **DATA N N N N N N N N N IN 3, 3 2 1 0** $3 \mid 2 \mid 1 \mid 0$ **3 - - -** $\overline{}$ **- - - - - - DATA 2 3 3 3 1 1 2 1 2 OUT 3 LD3(n) RD**₃(n) \longrightarrow **DATA N N N N N N N N N IN 4,** $3 \mid 2 \mid 1 \mid 0$ **2 1 0 3** $\overline{}$ **3 -** $\vert \ \ \vert$ **- - - - - - - - DATA 1 3 1 3 3 2 2 1 2 OUT 4** $-LD_4(n) RD_4(n)$ **LD**₄ $(n+1)$ ►

Figure 106. I ²S Timing for Multi-channel Mode, Four Data Lines

On Audio Serial Interface 1, any format (DSP, left or right-justified, or I²S) can be utilized in multi-channel, multiterminal mode. [Figure](#page-108-0) 106 shows an example of multi-channel, multi-terminal mode using 4 stereo data pairs (8 channels) in I²S format.

Figure 107. I ²S Timing for Multi-channel Mode, Four Data Lines with Offset1=1

For I²S multi-channel, multi-terminal mode, the programmed offset value should be less than the number of bitclocks per frame by at least the programmed word-length of the data.

8.3.10 miniDSP

The TLV320AIC3268 features two fully programmable miniDSP cores. The first miniDSP core is tightly coupled to the ADC, the second miniDSP core is tightly coupled to the DAC. The algorithms for the miniDSP must be loaded into the device after power up. The miniDSPs have direct access to the digital stereo audio stream on the ADC and on the DAC side, offering the possibility for advanced, very-low group delay DSP algorithms. Each miniDSP can run up to 1229 instructions on every audio sample at a 48kHz sample rate. The two cores can run fully synchronized and can exchange data. The miniDSPs in TLV320AIC3268 enable advanced sound enhancement algorithms on an audio device.

The TLV320AIC3268 features two fully programmable miniDSP cores and three ASI ports. The miniDSP_A is capable of generating 8 channels of audio data, which can either be routed to the ASIs to be output from the device or routed back to the miniDSP_D for a loopback function. Similarly the miniDSP_D can take in audio data from multiple ASIs or the miniDSP_A.

The miniDSP_A can generate 8 channels of data called the miniDSP_A_DataOutput[1:8]. The miniDSP_A_DataOutput[1,2,3,...,8] are also referred to as miniDSP_A_DataOutput[L1,R1,L2,...,R4]. When the device is used in pre-programmed PRB modes only miniDSP_A_DataOutput[1:2] are generated by the device for stereo modes and only miniDSP A DataOutput[1] is generated in the mono modes.

The miniDSP_D features 3 input ports called miniDSP_D_DataInput_1, miniDSP_D_DataInput_2 and
miniDSP D DataInput 3. The input port miniDSP D DataInput 1 features 8 channels, miniDSP_D_DataInput_3. The input port miniDSP_D_DataInput_1 features 8 channels, miniDSP_D_DataInput_1[1:8], also referred to as miniDSP_D_DataInput_1[L1,R1,...,R4]. In the pre-programmed stereo PRB modes only miniDSP_D_DataInput_1[1:2] are processed and other input ports are ignored. Similarly for the pre-programmed mono PRB modes only miniDSP_D_DataInput[1] is processed. The input port miniDSP_D_DataInput_2 features 2 channels called miniDSP_D_DataInput_2[1:2] also referred to as

miniDSP_D_DataInput_2[L1,R1]. The input port miniDSP_D_DataInput_3 also features 2 channels of audio data called miniDSP_D_DataInput_3[1:2] also referred to as miniDSP_D_DataInput_3[L1,R1]. Signal routing to miniDSP_D_DataInput_1[1:8] is controlled by programming B0_P4_R118_D[5:4]. Signal routing to miniDSP_D_DataInput_2[1,2] is controlled by programming B0_P4_R118_D[3:2]. Signal routing to miniDSP_D_DataInput_3[1,2] is controlled by programming B0_P4_R118_D[1:0].

The audio serial port ASI1, can take in 8 channels of audio data. By default the miniDSP_A_DataOutput[1:8] is routed to ASI1_DataOutput. By programming B0_P4_R7_D[2:0], the default programming can be changed to one of ASI1_DataInput[1:8], ASI2_DataInput[1:2] or ASI3_DataInput[1:2] to acheive ASI to ASI loopback. Similarly the serial output of ASI1 is routed to ASI1 DOUT by default but by programming B0 P4 R15 D[1:0], this can be changed to route ASI1 DIN, ASI2 DIN or ASI3 DIN to achieve terminal-to-terminal loopback between ASIs. The ASI1's parallel output ASI1 DataInput can be routed to any of the miniDSP D input ports.

The audio serial port ASI2 and ASI3, can take in 2 channels of audio data each. By default the inputs to ASI2_DataOutput[1:2] and ASI3_DataOutput[1:2] are disabled. By programming B0_P4_R23_D[2:0], one of miniDSP A DataOutput[1:2], miniDSP A DataOutput[3:4], ASI1 DataInput[1:2], ASI2 DataInput[1:2] or ASI3_DataInput[1:2] can be routed to ASI2_DataOutput[1:2]. Similarly by programming B0_P4_R39_D[2:0], one of miniDSP_A_DataOutput[1:2], miniDSP_A_DataOutput[5:6], ASI1_DataInput[1:2], ASI2_DataInput[1:2] or ASI3_DataInput[1:2] can be routed to ASI3_DataOutput[1:2]. The ASI2_DataInput and ASI3_DataInput can be routed to any of the miniDSP_D input ports.

The serial output ASI2_DOUT can be configured to route serial output of ASI2 or loop back ASI1_DIN, ASI2_DIN or ASI3_DIN. This feature is controlled by configuring B0_P4_R31_D[1:0]. The serial output of ASI3_DOUT can be configured to route serial output of ASI3 or loop back ASI1_DIN, ASI2_DIN or ASI3_DIN. This feature is controlled by configuring B0_P4_R47_D[1:0].

The signal routing between ASIs and miniDSPs is shown in [Figure](#page-110-0) 108.

Figure 108. Audio Routing Between ASI ports,÷ miniDSP

8.3.11 Device Connections

8.3.11.1 Digital Terminals

Only a small number of digital terminals are dedicated to a single function; whenever possible, the digital terminals have a default function, and also can be reprogrammed to cover alternative functions for various applications.

The fixed-function terminals are hardware-control terminals RESET and SPI_SELECT terminal. Depending on the state of SPI_SELECT, four terminals SCL_SSZ, SDA_MOSI, MISO_GPO1, and I2C_ADDR_SCLK are configured for either I²C or SPI protocol. Only in I²C mode, I2C_ADDR_SCLK provide two possible I²C addresses for the TLV320AIC3268, while this terminal receives the SPI SCLK when the device is set to SPI mode.

Other digital IO terminals can be configured for various functions via register control.

8.3.11.2 Analog Terminals

Analog functions can also be configured to a large degree. For minimum power consumption, analog blocks are powered down by default. The blocks can be powered up with fine granularity according to the application needs.

The possible analog routings of analog input terminals to ADCs and output amplifiers as well as the routing from DACs to output amplifiers can be seen in the Analog Routing Diagram.

8.3.11.3 Multifunction Terminals

[Table](#page-111-0) 46 and [Table](#page-112-0) 47 show the possible allocation of terminals for specific functions. The PLL input, for example, can be programmed to be any of 7 terminals (MCLK, BCLK1, DIN1, BCLK2,GPIO1, GPIO2, GPIO3).

Table 46. Multifunction Terminal Assignments for Terminals SDA_MOSI, SCL_SSZ, I2C_ADDR_SCLK,MISO_GPO1,MCLK, WCLK1, BCLK1, DIN1, and DOUT1

Table 46. Multifunction Terminal Assignments for Terminals SDA_MOSI, SCL_SSZ, I2C_ADDR_SCLK,MISO_GPO1,MCLK, WCLK1, BCLK1, DIN1, and DOUT1 (continued)

(1) E: The terminal is **exclusively** used for this function, no other function can be implemented with the same terminal (such as if DOUT1 has been allocated for General Purpose Output, it cannot be used as the INT1 output at the same time)

(2) D: The terminal is the **default** selection for the function

(3) S: This terminal can be **simultaneously** used with other functions marked S for the same terminal. (such as MCLK terminal could be chosen to drive the PLL, ADC Clock, DAC Clock, CDIV Clock, LFR Clock, HF Clock, and REF_1MHz_CLK inputs simultaneously)

(4) MISO_GPO1 can only be utilized for functions defined in this table when part utilizes I²C for control. In SPI mode, this terminal serves as MISO.

Table 47. Multifunction Terminal Assignments for Terminals WCLK2, BCLK2, DIN2, DOUT2, GPIO1, GPIO2, GPIO3, GPIO4 and GPIO5

Table 47. Multifunction Terminal Assignments for Terminals WCLK2, BCLK2, DIN2, DOUT2, GPIO1, GPIO2, GPIO3, GPIO4 and GPIO5 (continued)

- (1) E: The terminal is **exclusively** used for this function, no other function can be implemented with the same terminal (such as if DOUT1 has been allocated for General Purpose Output, it cannot be used as the INT1 output at the same time)
- (2) D: The terminal is the **default** selection for the function
- (3) S: This terminal can be **simultaneously** used with other functions marked S for the same terminal. (such as MCLK terminal could be chosen to drive the PLL, ADC Clock, DAC Clock, CDIV Clock, LFR Clock, HF Clock, and REF_1MHz_CLK inputs simultaneously)

See Register Settings for [Multifunction](#page-114-3) Terminals for details on register configuration of multi-function terminals.

8.3.11.3.1 Register Settings for Multifunction Terminals

[Table](#page-114-4) 48 summarizes the register settings that must be applied to configure the terminal assignments for general inputs and outputs, interrupts, clocking outputs, and digital microphones. In [Table](#page-114-4) 48, the letter/number combination represents the row and the column number from and in bold type.

Please be aware that more settings may be necessary to obtain a full functionality matching the application requirement.

Table 48. Multifunction Terminal Register Configuration - General Inputs/Outputs, Interrupts

Table 48. Multifunction Terminal Register Configuration - General Inputs/Outputs, Interrupts (continued)

[Table](#page-115-0) 49 summarizes the register settings that must be applied to configure the terminal assignments for clocking inputs and outputs from the device. In [Table](#page-115-0) 49, the letter/number combination represents the row and the column number from and in bold type.

Please be aware that more settings may be necessary to obtain a full functionality matching the application requirement.

Table 49. Multifunction Terminal Register Configuration - Clocking Inputs/Outputs

	Description	Required Register Setting		Description	Required Register Setting
AI ₅	PLL_CLKIN input on MCLK	B0 P0 R5 D[5:2]=0000;	AI7	PLL CLKIN input on BCLK1	B0 P0 R5 D[5:2]=0001;
AI8	PLL CLKIN input on DIN1	B0 P0 R5 D[5:2]=0011; B0_P4_R68_D[6:5]=01;	AI11	PLL CLKIN input on BCLK2	B0 P0 R5 D[5:2]=0100; B0_P4_R70_D[5:2]=0010;
AI14	PLL CLKIN input on GPIO1	B0_P0_R5_D[5:2]=0010; B0_P4_R86_D[6:2]=00001;	AI15	PLL CLKIN input on GPIO2	B0_P0_R5_D[5:2]=0111; B0_P4_R87_D[6:2]=00001;
AI16	PLL CLKIN input on GPIO3	B0 P0 R5 D[5:2]=0101; B0 P4 R88 D[6:2]=00001;	AJ5	ADC CLKIN input on MCLK	B0 P0 R4 D[3:0]=0000;
AJ7	ADC CLKIN input on BCLK1	B0_P0_R4_D[3:0]=0001;	AJ11	ADC CLKIN input on BCLK2	B0_P0_R4_D[3:0]=0100; B0_P4_R70_D[5:2]=0010;
AJ14	ADC CLKIN input on GPIO1	B0 P0 R4 D[3:0]=0010; B0 P4 R86 D[6:2]=00001;	AJ15	ADC CLKIN input on GPIO2	B0 P0 R4 D[3:0]=1001; B0 P4 R87 D[6:2]=00001;
AJ16	ADC CLKIN input on GPIO3	B0_P0_R4_D[3:0]=0101; B0_P4_R88_D[6:2]=00001;	AK ₅	DAC CLKIN input on MCLK	B0 P0 R4 D[7:4]=0000;
AK7	DAC CLKIN input on BCLK1	B0 P0 R4 D[7:4]=0001;	AK11	DAC CLKIN input on BCLK2	B0 P0 R4 D[7:4]=0100; B0 P4 R70 D[5:2]=0010;
AK14	DAC CLKIN input on GPIO1	B0_P0_R4_D[7:4]=0010; B0 P4 R86 D[6:2]=00001;	AK15	DAC CLKIN input on GPIO2	B0_P0_R4_D[7:4]=1001; B0_P4_R87_D[6:2]=00001;
AK16	DAC CLKIN input on GPIO3	B0_P0_R4_D[7:4]=0101; B0 P4 R88 D[6:2]=00001;	AL ₅	CDIV CLKIN input on MCLK	B0 P0 R21 D[4:0]=0000;
AL7	CDIV CLKIN input on BCLK1	B0 P0 R21 D[4:0]=0001;	AL8	CDIV CLKIN input on DIN1	B0 P0 R21 D[4:0]=0010; B0_P4_R68_D[6:5]=01;
AL ₁₁	CDIV_CLKIN input on BCLK2	B0 P0 R21 D[4:0]=1000; B0_P4_R70_D[5:2]=0010;	AL ₁₆	CDIV_CLKIN input on GPIO3	B0 P0 R21 D[4:0]=1001; B0_P4_R88_D[6:2]=00001;
AM ₅	LFR CLKIN input on MCLK	B0_P0_R24_D[7:4]=0000;	AM ₆	LFR CLKIN input on WCLK1	B0_P0_R24_D[7:4]=0001;
AM10	LFR CLKIN input on WCLK2	B0 P0 R24 D[7:4]=0011; B0_P4_R69_D[5:2]=0010;	AM11	LFR CLKIN input on BCLK2	B0 P0 R24 D[7:4]=0100; B0_P4_R70_D[5:2]=0010;
AM12	LFR CLKIN input on DIN2	B0 P0 R24 D[7:4]=0110; B0 P4 R72 D[6:5]=01;	AM14	LFR CLKIN input on GPIO1	B0 P0 R24 D[7:4]=0010; B0 P4 R86 D[6:2]=00001;
AM15	LFR CLKIN input on GPIO2	B0 P0 R24 D[7:4]=1000; B0_P4_R87_D[6:2]=00001;	AM16	LFR CLKIN input on GPIO3	B0 P0 R24 D[7:4]=0101; B0_P4_R88_D[6:2]=00001;
AN ₅	HF CLK input on MCLK	B0_P0_R24_D[3:0]=0000;	AO ₅	REF_1MHz_CLK input on MCLK	B0_P0_R23_D[7]=1;
AP4	CLKOUT output on MISO	B0 P4 R96 D[4:1]=0011; Configure B0 P0 R21 D[4:0] and B0_P0_R22_D[7:0]	AP ₆	CLKOUT output on WCLK1	B0 P4 R65 D[5:2]=0100; Configure B0_P0_R21_D[4:0] and B0_P0_R22_D[7:0]
AP9	CLKOUT output on DOUT1	B0 P4 R67 D[4:1]=0011; Configure B0_P0_R21_D[4:0] and B0 P0 R22 D[7:0]	AP10	CLKOUT output on WCLK2	B0_P4_R69_D[5:2]=0100; Configure B0_P0_R21_D[4:0] and B0 P0 R22 D[7:0]
AP11	CLKOUT output on BCLK2	B0 P4_R70_D[5:2]=0100; Configure B0 P0 R21 D[4:0] and B0 P0 R22 D[7:0]	AP14	CLKOUT output on GPIO1	B0 P4 R86 D[6:2]=00100; Configure B0 P0 R21 D[4:0] and B0 P0 R22 D[7:0]
AP15	CLKOUT output on GPIO2	B0_P4_R87_D[6:2]=00100; Configure B0 P0 R21 D[4:0] and B0_P0_R22_D[7:0]			

[Table](#page-116-0) 50 summarizes the register settings that must be applied to configure the terminal assignments for digital microphone feature in the device. In [Table](#page-116-0) 50, the letter/number combination represents the row and the column number from and in bold type.

Please be aware that more settings may be necessary to obtain a full functionality matching the application requirement.

Table 50. Multifunction Terminal Register Configuration - Digital Microphone

[Table](#page-118-0) 51 summarizes the register settings that must be applied to configure the terminal assignments for the audio serial interfaces. In [Table](#page-118-0) 51, the letter/number combination represents the row and the column number from [Table](#page-111-0) 46 and [Table](#page-112-0) 47 in bold type.

Please be aware that more settings may be necessary to obtain a full audio serial interface definition matching the application requirement (for example B0_P4_R1-R16 for Audio Serial Interface 1, B0_P4_R17-R32 for Audio Serial Interface 2, and B0_P4_R33-R48 for Audio Serial Interface 3).

Table 51. Multifunction Terminal Register Configuration - Audio Serial Interfaces

Table 51. Multifunction Terminal Register Configuration - Audio Serial Interfaces (continued)

8.4 Device Functional Modes

Functions Modes Details

-
- Differential Input Mode
- High Input Impedance Mode
- Low Input Impedance Mode
-
- Digital Microphone/PDM Input

Analog Inputs **Fixed Analog Inputs** Single Ended Input Mode **See ADC Signal [Routing](#page-42-0) for details.**

Audio Inputs **Audio Inputs** Audio Inputs **Allection for** Analog Inputs **See Digital [Microphone](#page-53-0) Function for** details.

Device Functional Modes (continued)

8.5 Programming

To enable the TLV320AIC3268 in a particular application, it needs to be comfigured or programmed. [Initialization](#page-246-0) [Setup](#page-246-0) describes various configurations required to enable the device.

To enable use of miniDSP in configurable modes, PurePath tools are provided. Please contact Texas Instruments for more details.

8.6 Register Maps

8.6.1 Register Map Summary

Table 52. Summary of Register Map

Copyright © 2014, Texas Instruments Incorporated *Submit [Documentation](http://www.go-dsp.com/forms/techdoc/doc_feedback.htm?litnum=SLAS953A&partnum=TLV320AIC3268) Feedback* 125

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8.6.2 Book 0 Page 0

Book 0 / Page 0 / Register 0: Page Select Register - 0x00 / 0x00 / 0x00 (B0_P0_R0)

Book 0 / Page 0 / Register 1: Software Reset Register - 0x00 / 0x00 / 0x01 (B0_P0_R1)

Book 0 / Page 0 / Register 2-3: Reserved Registers - 0x00 / 0x00 / 0x02-0x03 (B0_P0_R2-3)

Book 0 / Page 0 / Register 4: Clock Control Register 1, Clock Input Multiplexers - 0x00 / 0x00 / 0x04 **(B0_P0_R4)**

Book 0 / Page 0 / Register 4: Clock Control Register 1, Clock Input Multiplexers - 0x00 / 0x00 / 0x04 **(B0_P0_R4) (continued)**

Book 0 / Page 0 / Register 5: Clock Control Register 2, PLL Input Multiplexer - 0x00 / 0x00 / 0x05 **(B0_P0_R5)**

Book 0 / Page 0 / Register 6: Clock Control Register 3, PLL P and R Values - 0x00 / 0x00 / 0x06 **(B0_P0_R6)**

Book 0 / Page 0 / Register 7: Clock Control Register 4, PLL J Value - 0x00 / 0x00 / 0x07 (B0_P0_R7)

Book 0 / Page 0 / Register 8: Clock Control Register 5, PLL D Values (MSB) - 0x00 / 0x00 / 0x08 **(B0_P0_R8)**

Book 0 / Page 0 / Register 9: Clock Control Register 6, PLL D Values (LSB) - 0x00 / 0x00 / 0x09 **(B0_P0_R9)**

Book 0 / Page 0 / Register 10: Clock Control Register 7, PLL_CLKIN Divider - 0x00 / 0x00 / 0x0A **(B0_P0_R10)**

Book 0 / Page 0 / Register 11: Clock Control Register 8, NDAC Divider Values - 0x00 / 0x00 / 0x0B **(B0_P0_R11)**

Book 0 / Page 0 / Register 11: Clock Control Register 8, NDAC Divider Values - 0x00 / 0x00 / 0x0B **(B0_P0_R11) (continued)**

Book 0 / Page 0 / Register 12: Clock Control Register 9, MDAC Divider Values - 0x00 / 0x00 / 0x0C **(B0_P0_R12)**

Book 0 / Page 0 / Register 13: DAC OSR Control Register 1, MSB Value - 0x00 / 0x00 / 0x0D (B0_P0_R13)

Book 0 / Page 0 / Register 14: DAC OSR Control Register 2, LSB Value - 0x00 / 0x00 / 0x0E (B0_P0_R14)

Book 0 / Page 0 / Register 15-17: Reserved Registers - 0x00 / 0x00 / 0x0F-0x11 (B0_P0_R15-17)

Book 0 / Page 0 / Register 18: Clock Control Register 10, NADC Values - 0x00 / 0x00 / 0x12 (B0_P0_R18)

Book 0 / Page 0 / Register 19: Clock Control Register 11, MADC Values - 0x00 / 0x00 / 0x13 (B0_P0_R19)

Book 0 / Page 0 / Register 20: ADC Oversampling (AOSR) Register - 0x00 / 0x00 / 0x14 (B0_P0_R20)

Book 0 / Page 0 / Register 21: CLKOUT MUX - 0x00 / 0x00 / 0x15 (B0_P0_R21)

Book 0 / Page 0 / Register 22: Clock Control Register 12, CLKOUT M Divider Value - 0x00 / 0x00 / 0x16 **(B0_P0_R22)**

Book 0 / Page 0 / Register 23: Timer clock - 0x00 / 0x00 / 0x17 (B0_P0_R23)

Book 0 / Page 0 / Register 24: Low Frequency Clock Generation Control - 0x00 / 0x00 / 0x18 (B0_P0_R24)

Book 0 / Page 0 / Register 25: High Frequency Clock Generation Control 1 - 0x00 / 0x00 / 0x19 **(B0_P0_R25)**

Book 0 / Page 0 / Register 25: High Frequency Clock Generation Control 1 - 0x00 / 0x00 / 0x19 **(B0_P0_R25) (continued)**

Book 0 / Page 0 / Register 26: High Freguency Clock Generation Control 2 - 0x00 / 0x00 / 0x1A **(B0_P0_R26)**

Book 0 / Page 0 / Register 27: High Frequency Clock Generation Control 3 - 0x00 / 0x00 / 0x1B **(B0_P0_R27)**

Book 0 / Page 0 / Register 28: High Frequency Clock Generation Control 4 - 0x00 / 0x00 / 0x1C **(B0_P0_R28)**

Book 0 / Page 0 / Register 29: High Frequency Clock Trim Control 1 - 0x00 / 0x00 / 0x1D (B0_P0_R29)

Book 0 / Page 0 / Register 30: High Frequency Clock Trim Control 2 - 0x00 / 0x00 / 0x1E (B0_P0_R30)

XAS **ISTRUMENTS**

Book 0 / Page 0 / Register 31: High Frequency Clock Trim Control 3 - 0x00 / 0x00 / 0x1F (B0_P0_R31)

Book 0 / Page 0 / Register 32: High Frequency Clock Trim Control 4 - 0x00 / 0x00 / 0x20 (B0_P0_R32)

Book 0 / Page 0 / Register 33-35: Reserved Registers - 0x00 / 0x00 / 0x21-0x23 (B0_P0_R33-35)

Book 0 / Page 0 / Register 36: ADC Flag Register - 0x00 / 0x00 / 0x24 (B0_P0_R36)

Book 0 / Page 0 / Register 37: DAC Flag Register - 0x00 / 0x00 / 0x25 (B0_P0_R37)

Book 0 / Page 0 / Register 38: DAC Flag Register - 0x00 / 0x00 / 0x26 (B0_P0_R38)

Book 0 / Page 0 / Register 39-41: Reserved Registers - 0x00 / 0x00 / 0x27-0x29 (B0_P0_R39-41)

Book 0 / Page 0 / Register 42: Sticky Flag Register 1 - 0x00 / 0x00 / 0x2A (B0_P0_R42)

Book 0 / Page 0 / Register 43: Reserved Register - 0x00 / 0x00 / 0x2B (B0_P0_R43)

Book 0 / Page 0 / Register 44: Sticky Flag Register 2 - 0x00 / 0x00 / 0x2C (B0_P0_R44)

Book 0 / Page 0 / Register 44: Sticky Flag Register 2 - 0x00 / 0x00 / 0x2C (B0_P0_R44) (continued)

Book 0 / Page 0 / Register 45: Sticky Flag Register 3 - 0x00 / 0x00 / 0x2D (B0_P0_R45)

Book 0 / Page 0 / Register 46-47: Reserved Register - 0x00 / 0x00 / 0x2E-0x2F (B0_P0_R46-47)

Book 0 / Page 0 / Register 48: INT1 Interrupt Control - 0x00 / 0x00 / 0x30 (B0_P0_R48)

Book 0 / Page 0 / Register 48: INT1 Interrupt Control - 0x00 / 0x00 / 0x30 (B0_P0_R48) (continued)

Book 0 / Page 0 / Register 49: INT2 Interrupt Control - 0x00 / 0x00 / 0x31 (B0_P0_R49)

Book 0 / Page 0 / Register 50: SAR Control 1 - 0x00 / 0x00 / 0x32 (B0_P0_R50)

Book 0 / Page 0 / Register 51: Interrupt Format Control Register - 0x00 / 0x00 / 0x33 (B0_P0_R51)

Book 0 / Page 0 / Register 52-59: Reserved Registers - 0x00 / 0x00 / 0x34-0x3B (B0_P0_R52-59)

Book 0 / Page 0 / Register 60: DAC Processing Block and miniDSP Power Control - 0x00 / 0x00 / 0x3C **(B0_P0_R60)**

Book 0 / Page 0 / Register 61: ADC Processing Block Control - 0x00 / 0x00 / 0x3D (B0_P0_R61)

Book 0 / Page 0 / Register 62: Reserved Register - 0x00 / 0x00 / 0x3E (B0_P0_R62)

Book 0 / Page 0 / Register 63: Primary DAC Power and Soft-Stepping Control - 0x00 / 0x00 / 0x3F **(B0_P0_R63)**

Book 0 / Page 0 / Register 64: Primary DAC Master Volume Configuration - 0x00 / 0x00 / 0x40 (B0_P0_R64)

Book 0 / Page 0 / Register 65: Primary DAC Left Volume Control Setting - 0x00 / 0x00 / 0x41 (B0_P0_R65)

Book 0 / Page 0 / Register 66: Primary DAC Right Volume Control Setting - 0x00 / 0x00 / 0x42 **(B0_P0_R66)**

Book 0 / Page 0 / Register 67: Headset Detection - 0x00 / 0x00 / 0x43 (B0_P0_R67)

Book 0 / Page 0 / Register 68: Reserved Register - 0x00 / 0x00 / 0x44 (B0_P0_R68)

Book 0 / Page 0 / Register 69: Reserved Register - 0x00 / 0x00 / 0x45 (B0_P0_R69)

Book 0 / Page 0 / Register 70: Reserved Register - 0x00 / 0x00 / 0x46 (B0_P0_R70)

Book 0 / Page 0 / Register 71-80: Reserved Registers - 0x00 / 0x00 / 0x47-0x50 (B0_P0_R71-80)

Book 0 / Page 0 / Register 81: ADC Channel Power Control - 0x00 / 0x00 / 0x51 (B0_P0_R81)

Book 0 / Page 0 / Register 82: ADC Fine Gain Volume Control - 0x00 / 0x00 / 0x52 (B0_P0_R82)

Book 0 / Page 0 / Register 83: Left ADC Volume Control - 0x00 / 0x00 / 0x53 (B0_P0_R83)

Book 0 / Page 0 / Register 83: Left ADC Volume Control - 0x00 / 0x00 / 0x53 (B0_P0_R83) (continued)

Book 0 / Page 0 / Register 84: Right ADC Volume Control - 0x00 / 0x00 / 0x54 (B0_P0_R84)

Book 0 / Page 0 / Register 85: ADC Phase Control - 0x00 / 0x00 / 0x55 (B0_P0_R85)

Book 0 / Page 0 / Register 86: Left AGC Control 1 - 0x00 / 0x00 / 0x56 (B0_P0_R86)

Book 0 / Page 0 / Register 86: Left AGC Control 1 - 0x00 / 0x00 / 0x56 (B0_P0_R86) (continued)

Book 0 / Page 0 / Register 87: Left AGC Control 2 - 0x00 / 0x00 / 0x57 (B0_P0_R87)

Book 0 / Page 0 / Register 88: Left AGC Control 3 - 0x00 / 0x00 / 0x58 (B0_P0_R88)

Book 0 / Page 0 / Register 89: Left AGC Attack Time - 0x00 / 0x00 / 0x59 (B0_P0_R89)

Book 0 / Page 0 / Register 89: Left AGC Attack Time - 0x00 / 0x00 / 0x59 (B0_P0_R89) (continued)

Book 0 / Page 0 / Register 90: Left AGC Decay Time - 0x00 / 0x00 / 0x5A (B0_P0_R90)

Book 0 / Page 0 / Register 91: Left AGC Noise Debounce - 0x00 / 0x00 / 0x5B (B0_P0_R91)

Book 0 / Page 0 / Register 92: Left AGC Signal Debounce - 0x00 / 0x00 / 0x5C (B0_P0_R92)

Book 0 / Page 0 / Register 92: Left AGC Signal Debounce - 0x00 / 0x00 / 0x5C (B0_P0_R92) (continued)

Book 0 / Page 0 / Register 93: Left AGC Gain - 0x00 / 0x00 / 0x5D (B0_P0_R93)

Book 0 / Page 0 / Register 94: Right AGC Control 1 - 0x00 / 0x00 / 0x5E (B0_P0_R94)

Book 0 / Page 0 / Register 95: Right AGC Control 2 - 0x00 / 0x00 / 0x5F (B0_P0_R95)

Book 0 / Page 0 / Register 95: Right AGC Control 2 - 0x00 / 0x00 / 0x5F (B0_P0_R95) (continued)

Book 0 / Page 0 / Register 96: Right AGC Control 3 - 0x00 / 0x00 / 0x60 (B0_P0_R96)

Book 0 / Page 0 / Register 97: Right AGC Attack Time - 0x00 / 0x00 / 0x61 (B0_P0_R97)

Book 0 / Page 0 / Register 98: Right AGC Decay Time - 0x00 / 0x00 / 0x62 (B0_P0_R98)

Book 0 / Page 0 / Register 98: Right AGC Decay Time - 0x00 / 0x00 / 0x62 (B0_P0_R98) (continued)

Book 0 / Page 0 / Register 99: Right AGC Noise Debounce - 0x00 / 0x00 / 0x63 (B0_P0_R99)

Book 0 / Page 0 / Register 100: Right AGC Signal Debounce - 0x00 / 0x00 / 0x64 (B0_P0_R100)

Book 0 / Page 0 / Register 101: Right AGC Gain - 0x00 / 0x00 / 0x65 (B0_P0_R101)

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Book 0 / Page 0 / Register 113-114: Reserved Registers - 0x00 / 0x00 / 0x71-0x72 (B0_P0_R113-114)

Book 0 / Page 0 / Register 115: I2C Interface Miscellaneous Control - 0x00 / 0x00 / 0x73 (B0_P0_R115)

Book 0 / Page 0 / Register 116-119: Reserved Registers - 0x00 / 0x00 / 0x74-0x77 (B0_P0_R116-119)

Book 0 / Page 0 / Register 120: miniDSP Control Register access - 0x00 / 0x00 / 0x78 (B0_P0_R120)

Book 0 / Page 0 / Register 121-126: Reserved Registers - 0x00 / 0x00 / 0x79-0x7E (B0_P0_R121-126)

Book 0 / Page 0 / Register 127: Book Selection Register - 0x00 / 0x00 / 0x7F (B0_P0_R127)

8.6.3 Book 0 Page 1

Book 0 / Page 1 / Register 0: Page Select Register - 0x00 / 0x01 / 0x00 (B0_P1_R0)

Book 0 / Page 1 / Register 1: Power Configuration Register - 0x00 / 0x01 / 0x01 (B0_P1_R1)

Book 0 / Page 1 / Register 2: Reserved Register - 0x00 / 0x01 / 0x02 (B0_P1_R2)

Book 0 / Page 1 / Register 3: Left DAC PowerTune Configuration Register - 0x00 / 0x01 / 0x03 (B0_P1_R3)

Book 0 / Page 1 / Register 4: Right DAC PowerTune Configuration Register - 0x00 / 0x01 / 0x04 **(B0_P1_R4)**

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Book 0 / Page 1 / Register 10: Receiver Output Driver Control - 0x00 / 0x01 / 0x0A (B0_P1_R10)

Book 0 / Page 1 / Register 10: Receiver Output Driver Control - 0x00 / 0x01 / 0x0A (B0_P1_R10) (continued)

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Book 0 / Page 1 / Register 12: Receiver Output Driver De-Pop Control - 0x00 / 0x01 / 0x0C (B0_P1_R12)

Book 0 / Page 1 / Register 12: Receiver Output Driver De-Pop Control - 0x00 / 0x01 / 0x0C (B0_P1_R12) (continued)

Book 0 / Page 1 / Register 13-16: Reserved Registers - 0x00 / 0x01 / 0x0D-0x10 (B0_P1_R13-16)

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Book 0 / Page 1 / Register 18: Left ADC PGA to Left Mixer Amplifier (MAL) Volume Control - 0x00 / 0x01 / **0x12 (B0_P1_R18)**

Book 0 / Page 1 / Register 18: Left ADC PGA to Left Mixer Amplifier (MAL) Volume Control - 0x00 / 0x01 / **0x12 (B0_P1_R18) (continued)**

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Book 0 / Page 1 / Register 22: Lineout Amplifier Control 1 - 0x00 / 0x01 / 0x16 (B0_P1_R22) (continued)

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Book 0 / Page 1 / Register 24-26: Reserved - 0x00 / 0x01 / 0x18-0x1A (B0_P1_R24-26)

Book 0 / Page 1 / Register 27: Headphone Amplifier Control 1 - 0x00 / 0x01 / 0x1B (B0_P1_R27)

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0001: Volume Control = -32.7 dB

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Book 0 / Page 1 / Register 28: Headphone Amplifier Control 2 - 0x00 / 0x01 / 0x1C (B0_P1_R28) (continued)

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Book 0 / Page 1 / Register 29: Headphone Amplifier Control 3 - 0x00 / 0x01 / 0x1D (B0_P1_R29) (continued)

0001: Volume Control = -32.7 dB

Book 0 / Page 1 / Register 29: Headphone Amplifier Control 3 - 0x00 / 0x01 / 0x1D (B0_P1_R29) (continued)

Book 0 / Page 1 / Register 30: Reserved Register - 0x00 / 0x01 / 0x1E (B0_P1_R30)

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Book 0 / Page 1 / Register 32: HPR Driver Volume Control - 0x00 / 0x01 / 0x20 (B0_P1_R32)

Book 0 / Page 1 / Register 32: HPR Driver Volume Control - 0x00 / 0x01 / 0x20 (B0_P1_R32) (continued)

Book 0 / Page 1 / Register 33: Reserved Register - 0x00 / 0x01 / 0x21 (B0_P1_R33)

Book 0 / Page 1 / Register 34: Reserved Register - 0x00 / 0x01 / 0x22 (B0_P1_R34)

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Book 0 / Page 1 / Register 36: Receiver Amplifier Control 1 - 0x00 / 0x01 / 0x24 (B0_P1_R36) (continued)

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Г

Book 0 / Page 1 / Register 37: Receiver Amplifier Control 2 - 0x00 / 0x01 / 0x25 (B0_P1_R37) (continued)

Book 0 / Page 1 / Register 37: Receiver Amplifier Control 2 - 0x00 / 0x01 / 0x25 (B0_P1_R37) (continued)

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Book 0 / Page 1 / Register 38: Receiver Amplifier Control 3 - 0x00 / 0x01 / 0x26 (B0_P1_R38) (continued)

Book 0 / Page 1 / Register 38: Receiver Amplifier Control 3 - 0x00 / 0x01 / 0x26 (B0_P1_R38) (continued)

Book 0 / Page 1 / Register 39: Receiver Amplifier Control 4 - 0x00 / 0x01 / 0x27 (B0_P1_R39)

Book 0 / Page 1 / Register 39: Receiver Amplifier Control 4 - 0x00 / 0x01 / 0x27 (B0_P1_R39) (continued)

Book 0 / Page 1 / Register 39: Receiver Amplifier Control 4 - 0x00 / 0x01 / 0x27 (B0_P1_R39) (continued)

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Book 0 / Page 1 / Register 40: Receiver Amplifier Control 5 - 0x00 / 0x01 / 0x28 (B0_P1_R40) (continued)

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Book 0 / Page 1 / Register 45: Speaker Amplifier Control 1 - 0x00 / 0x01 / 0x2D (B0_P1_R45)

Book 0 / Page 1 / Register 46: Speaker Amplifier Control 2 - 0x00 / 0x01 / 0x2E (B0_P1_R46)

READ/ RESET BIT DESCRIPTION WRITE VALUE D6-D0 R/W 111 1111 LOL Output Routed to SPK Driver Volume Control: 0000: Volume Control = 0.00 dB 0001: Volume Control = -0.5 dB 0010: Volume Control = -1.0 dB 0011: Volume Control = -1.5 dB 0100: Volume Control = -2.0 dB 0101: Volume Control = -2.5 dB 0110: Volume Control = -3.0 dB 0111: Volume Control = -3.5 dB 1000: Volume Control = -4.0 dB 1001: Volume Control = -4.5 dB 1010: Volume Control = -5.0 dB 1011: Volume Control = -5.5 dB 1100: Volume Control = -6.0 dB 1101: Volume Control = -6.5 dB 1110: Volume Control = -7.0 dB 1111: Volume Control = -7.5 dB 0000: Volume Control = -8.0 dB 0001: Volume Control = -8.5 dB 0010: Volume Control = -9.0 dB 0011: Volume Control = -9.5 dB 0100: Volume Control = -10.0 dB 0101: Volume Control = -10.5 dB 0110: Volume Control = -11.0 dB 0111: Volume Control = -11.5 dB 1000: Volume Control = -12.0 dB 1001: Volume Control = -12.5 dB 1010: Volume Control = -13.0 dB 1011: Volume Control = -13.5 dB 1100: Volume Control = -14.1 dB 1101: Volume Control = -14.6 dB 1110: Volume Control = -15.1 dB 1111: Volume Control = -15.6 dB 0000: Volume Control = -16.0 dB 0001: Volume Control = -16.5 dB 0010: Volume Control = -17.1 dB 0011: Volume Control = -17.5 dB 0100: Volume Control = -18.1 dB 0101: Volume Control = -18.6 dB 0110: Volume Control = -19.1 dB 0111: Volume Control = -19.6 dB 1000: Volume Control = -20.1 dB 1001: Volume Control = -20.6 dB 1010: Volume Control = -21.1 dB 1011: Volume Control = -21.6 dB 1100: Volume Control = -22.1 dB 1101: Volume Control = -22.6 dB 1110: Volume Control = -23.1 dB 1111: Volume Control = -23.6 dB 0000: Volume Control = -24.1 dB 0001: Volume Control = -24.6 dB 0010: Volume Control = -25.1 dB 0011: Volume Control = -25.6 dB 0100: Volume Control = -26.1 dB 0101: Volume Control = -26.6 dB 0110: Volume Control = -27.1 dB 0111: Volume Control = -27.6 dB 1000: Volume Control = -28.1 dB 1001: Volume Control = -28.6 dB 1010: Volume Control = -29.1 dB 1011: Volume Control = -29.6 dB 1100: Volume Control = -30.1 dB 1101: Volume Control = -30.6 dB 1110: Volume Control = -31.1 dB 1111: Volume Control = -31.6 dB 0000: Volume Control = -32.1 dB 0001: Volume Control = -32.7 dB 0010: Volume Control = -33.1 dB

Book 0 / Page 1 / Register 46: Speaker Amplifier Control 2 - 0x00 / 0x01 / 0x2E (B0_P1_R46) (continued)

Book 0 / Page 1 / Register 46: Speaker Amplifier Control 2 - 0x00 / 0x01 / 0x2E (B0_P1_R46) (continued)

Book 0 / Page 1 / Register 47: Speaker Amplifier Control 3 - 0x00 / 0x01 / 0x2F (B0_P1_R47)

READ/ RESET BIT DESCRIPTION WRITE VALUE D6-D0 R/W 111 1111 LOR Output Routed to SPK_RIGHT_CH_IN Volume Control: 0000: Volume Control = 0.00 dB 0001: Volume Control = -0.5 dB 0010: Volume Control = -1.0 dB 0011: Volume Control = -1.5 dB 0100: Volume Control = -2.0 dB 0101: Volume Control = -2.5 dB 0110: Volume Control = -3.0 dB 0111: Volume Control = -3.5 dB 1000: Volume Control = -4.0 dB 1001: Volume Control = -4.5 dB 1010: Volume Control = -5.0 dB 1011: Volume Control = -5.5 dB 1100: Volume Control = -6.0 dB 1101: Volume Control = -6.5 dB 1110: Volume Control = -7.0 dB 1111: Volume Control = -7.5 dB 0000: Volume Control = -8.0 dB 0001: Volume Control = -8.5 dB 0010: Volume Control = -9.0 dB 0011: Volume Control = -9.5 dB 0100: Volume Control = -10.0 dB 0101: Volume Control = -10.5 dB 0110: Volume Control = -11.0 dB 0111: Volume Control = -11.5 dB 1000: Volume Control = -12.0 dB 1001: Volume Control = -12.5 dB 1010: Volume Control = -13.0 dB 1011: Volume Control = -13.5 dB 1100: Volume Control = -14.1 dB 1101: Volume Control = -14.6 dB 1110: Volume Control = -15.1 dB 1111: Volume Control = -15.6 dB 0000: Volume Control = -16.0 dB 0001: Volume Control = -16.5 dB 0010: Volume Control = -17.1 dB 0011: Volume Control = -17.5 dB 0100: Volume Control = -18.1 dB 0101: Volume Control = -18.6 dB 0110: Volume Control = -19.1 dB 0111: Volume Control = -19.6 dB 1000: Volume Control = -20.1 dB 1001: Volume Control = -20.6 dB 1010: Volume Control = -21.1 dB 1011: Volume Control = -21.6 dB 1100: Volume Control = -22.1 dB 1101: Volume Control = -22.6 dB 1110: Volume Control = -23.1 dB 1111: Volume Control = -23.6 dB 0000: Volume Control = -24.1 dB 0001: Volume Control = -24.6 dB 0010: Volume Control = -25.1 dB 0011: Volume Control = -25.6 dB 0100: Volume Control = -26.1 dB 0101: Volume Control = -26.6 dB 0110: Volume Control = -27.1 dB 0111: Volume Control = -27.6 dB 1000: Volume Control = -28.1 dB 1001: Volume Control = -28.6 dB 1010: Volume Control = -29.1 dB 1011: Volume Control = -29.6 dB 1100: Volume Control = -30.1 dB 1101: Volume Control = -30.6 dB 1110: Volume Control = -31.1 dB 1111: Volume Control = -31.6 dB 0000: Volume Control = -32.1 dB 0001: Volume Control = -32.7 dB 0010: Volume Control = -33.1 dB

Book 0 / Page 1 / Register 47: Speaker Amplifier Control 3 - 0x00 / 0x01 / 0x2F (B0_P1_R47) (continued)

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Book 0 / Page 1 / Register 52: Input Select 1 for Left Microphone PGA P-Terminal - 0x00 / 0x01 / 0x34 **(B0_P1_R52) (continued)**

Book 0 / Page 1 / Register 53: Input Select 2 for Left Microphone PGA P-Terminal - 0x00 / 0x01 / 0x35 **(B0_P1_R53)**

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83& Book 0 / Page 1 / Register 56: Input Select 2 for Right Microphone PGA P-Terminal - 0x00 / 0x01 **(B0_P1_R56)**

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8.6.4 Book 0 Page 3

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Book 0 / Page 3 / Register 69: TEMP2 Measurement Data (LSB) - 0x00 / 0x03 / 0x45 (B0_P3_R69)

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Book 0 / Page 3 / Register 70-127: Reserved Registers - 0x00 / 0x03 / 0x46-0x7F (B0_P3_R70-127)

8.6.5 Book 0 Page 4

Book 0 / Page 4 / Register 0: Page Select Register - 0x00 / 0x04 / 0x00 (B0_P4_R0)

Book 0 / Page 4 / Register 1: ASI1, Audio Bus Format Control Register - 0x00 / 0x04 / 0x01 (B0_P4_R1)

Book 0 / Page 4 / Register 2: ASI1, Left Ch_Offset_1 Control Register - 0x00 / 0x04 / 0x02 (B0_P4_R2)

Book 0 / Page 4 / Register 3: ASI1, Right Ch_Offset_2 Control Register - 0x00 / 0x04 / 0x03 (B0_P4_R3)

Book 0 / Page 4 / Register 4: ASI1, Channel Setup Register - 0x00 / 0x04 / 0x04 (B0_P4_R4)

Book 0 / Page 4 / Register 4: ASI1, Channel Setup Register - 0x00 / 0x04 / 0x04 (B0_P4_R4) (continued)

Book 0 / Page 4 / Register 5: ASI1, ADC Audio Bus Format Control Register - 0x00 / 0x04 / 0x05 **(B0_P4_R5)**

Book 0 / Page 4 / Register 6: Audio Serial Interface 1, Multi-Pin Mode - 0x00 / 0x04 / 0x06 (B0_P4_R6)

Book 0 / Page 4 / Register 7: ASI1, ADC Input Control - 0x00 / 0x04 / 0x07 (B0_P4_R7)

Book 0 / Page 4 / Register 8: ASI1, DAC Output Control - 0x00 / 0x04 / 0x08 (B0_P4_R8)

Book 0 / Page 4 / Register 9: ASI1, Control Register 9, ADC Slot Tristate Control - 0x00 / 0x04 / 0x09 **(B0_P4_R9)**

Book 0 / Page 4 / Register 10: ASI1, WCLK and BCLK Control Register - 0x00 / 0x04 / 0x0A (B0_P4_R10)

Book 0 / Page 4 / Register 10: ASI1, WCLK and BCLK Control Register - 0x00 / 0x04 / 0x0A **(B0_P4_R10) (continued)**

Book 0 / Page 4 / Register 11: ASI1, Bit Clock N Divider Input Control - 0x00 / 0x04 / 0x0B (B0_P4_R11)

Book 0 / Page 4 / Register 12: ASI1, Bit Clock N Divider - 0x00 / 0x04 / 0x0C (B0_P4_R12)

Book 0 / Page 4 / Register 13: ASI 1, Word Clock N Divider - 0x00 / 0x04 / 0x0D (B0_P4_R13)

Book 0 / Page 4 / Register 13: ASI 1, Word Clock N Divider - 0x00 / 0x04 / 0x0D (B0_P4_R13) (continued)

Book 0 / Page 4 / Register 14: ASI1, BCLK and WCLK Output - 0x00 / 0x04 / 0x0E (B0_P4_R14)

Book 0 / Page 4 / Register 15: ASI1, Data Output - 0x00 / 0x04 / 0x0F (B0_P4_R15)

Book 0 / Page 4 / Register 16: ASI1, ADC Word Clock and Bit Clock Control - 0x00 / 0x04 / 0x10 **(B0_P4_R16)**

Book 0 / Page 4 / Register 16: ASI1, ADC Word Clock and Bit Clock Control - 0x00 / 0x04 / 0x10 **(B0_P4_R16) (continued)**

Book 0 / Page 4 / Register 17: ASI2, Audio Bus Format Control Register - 0x00 / 0x04 / 0x11 (B0_P4_R17)

Book 0 / Page 4 / Register 18: ASI2, Data Offset Control Register - 0x00 / 0x04 / 0x12 (B0_P4_R18)

Book 0 / Page 4 / Register 19-20: Reserved Registers - 0x00 / 0x04 / 0x13-0x14 (B0_P4_R19-20)

Book 0 / Page 4 / Register 21: ASI2, ADC Audio Bus Format Control Register - 0x00 / 0x04 / 0x15 **(B0_P4_R21)**

Book 0 / Page 4 / Register 21: ASI2, ADC Audio Bus Format Control Register - 0x00 / 0x04 / 0x15 **(B0_P4_R21) (continued)**

Book 0 / Page 4 / Register 22: Reserved Register - 0x00 / 0x04 / 0x16 (B0_P4_R22)

Book 0 / Page 4 / Register 23: ASI2, ADC Input Control - 0x00 / 0x04 / 0x17 (B0_P4_R23)

Book 0 / Page 4 / Register 24: ASI 2, DAC Output Control - 0x00 / 0x04 / 0x18 (B0_P4_R24)

Book 0 / Page 4 / Register 25: Reserved Register - 0x00 / 0x04 / 0x19 (B0_P4_R25)

Book 0 / Page 4 / Register 26: ASI2, Word Clock and Bit Clock Control Register - 0x00 / 0x04 / 0x1A **(B0_P4_R26)**

Book 0 / Page 4 / Register 26: ASI2, Word Clock and Bit Clock Control Register - 0x00 / 0x04 / 0x1A **(B0_P4_R26) (continued)**

Book 0 / Page 4 / Register 27: ASI2, Bit Clock N Divider Input Control - 0x00 / 0x04 / 0x1B (B0_P4_R27)

Book 0 / Page 4 / Register 28: ASI2, Bit Clock N Divider - 0x00 / 0x04 / 0x1C (B0_P4_R28)

Book 0 / Page 4 / Register 29: ASI2, Word Clock N Divider - 0x00 / 0x04 / 0x1D (B0_P4_R29)

Book 0 / Page 4 / Register 30: ASI2, Bit Clock and Word Clock Output - 0x00 / 0x04 / 0x1E (B0_P4_R30)

Book 0 / Page 4 / Register 31: ASI2, Data Output - 0x00 / 0x04 / 0x1F (B0_P4_R31)

Book 0 / Page 4 / Register 32: ASI2, ADC Word Clock and Bit Clock Control - 0x00 / 0x04 / 0x20 **(B0_P4_R32)**

Book 0 / Page 4 / Register 32: ASI2, ADC Word Clock and Bit Clock Control - 0x00 / 0x04 / 0x20 **(B0_P4_R32) (continued)**

Book 0 / Page 4 / Register 33: ASI3, Audio Bus Format Control Register - 0x00 / 0x04 / 0x21 (B0_P4_R33)

Book 0 / Page 4 / Register 34: ASI3, Data Offset Control Register - 0x00 / 0x04 / 0x22 (B0_P4_R34)

Book 0 / Page 4 / Register 35-36: Reserved Registers - 0x00 / 0x04 / 0x23-0x24 (B0_P4_R35-36)

Book 0 / Page 4 / Register 37: Reserved Register - 0x00 / 0x04 / 0x25 (B0_P4_R37)

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Book 0 / Page 4 / Register 39: ASI3, ADC Input Control - 0x00 / 0x04 / 0x27 (B0_P4_R39)

Book 0 / Page 4 / Register 40: ASI3, DAC Output Control - 0x00 / 0x04 / 0x28 (B0_P4_R40)

Book 0 / Page 4 / Register 41: Reserved Register - 0x00 / 0x04 / 0x29 (B0_P4_R41)

Book 0 / Page 4 / Register 42: ASI3, Word Clock and Bit Clock Control Register - 0x00 / 0x04 / 0x2A **(B0_P4_R42)**

Book 0 / Page 4 / Register 43: ASI3, Bit Clock N Divider Input Control - 0x00 / 0x04 / 0x2B (B0_P4_R43)

Book 0 / Page 4 / Register 43: ASI3, Bit Clock N Divider Input Control - 0x00 / 0x04 / 0x2B **(B0_P4_R43) (continued)**

Book 0 / Page 4 / Register 44: ASI3, Bit Clock N Divider - 0x00 / 0x04 / 0x2C (B0_P4_R44)

Book 0 / Page 4 / Register 45: ASI3, Word Clock N Divider - 0x00 / 0x04 / 0x2D (B0_P4_R45)

Book 0 / Page 4 / Register 46: ASI3, Bit Clock and Word Clock Output - 0x00 / 0x04 / 0x2E (B0_P4_R46)

Book 0 / Page 4 / Register 46: ASI3, Bit Clock and Word Clock Output - 0x00 / 0x04 / 0x2E **(B0_P4_R46) (continued)**

Book 0 / Page 4 / Register 47: ASI3, Data Output - 0x00 / 0x04 / 0x2F (B0_P4_R47)

Book 0 / Page 4 / Register 48: Reserved Register. - 0x00 / 0x04 / 0x30 (B0_P4_R48)

Book 0 / Page 4 / Register 49: ASI1 L1, R1 Input Control - 0x00 / 0x04 / 0x31 (B0_P4_R49)

Book 0 / Page 4 / Register 50: ASI1 L2, R2 Input Control - 0x00 / 0x04 / 0x32 (B0_P4_R50)

Book 0 / Page 4 / Register 51: ASI1 L3, R3 Input Control - 0x00 / 0x04 / 0x33 (B0_P4_R51)

Book 0 / Page 4 / Register 51: ASI1 L3, R3 Input Control - 0x00 / 0x04 / 0x33 (B0_P4_R51) (continued)

Book 0 / Page 4 / Register 52: ASI1 L4, R4 Input Control - 0x00 / 0x04 / 0x34 (B0_P4_R52)

Book 0 / Page 4 / Register 53: Reserved Register - 0x00 / 0x04 / 0x35 (B0_P4_R53)

Book 0 / Page 4 / Register 54: ASI2, DIN Input Multiplexer Control - 0x00 / 0x04 / 0x36 (B0_P4_R54)

Book 0 / Page 4 / Register 55: ASI3, Word Clock and Bit Clock Input Multiplexer Control - 0x00 / 0x04 / **0x37 (B0_P4_R55)**

Book 0 / Page 4 / Register 56: ASI3, DIN Input Multiplexer Control - 0x00 / 0x04 / 0x38 (B0_P4_R56)

Book 0 / Page 4 / Register 57-64: Reserved Registers - 0x00 / 0x04 / 0x39-0x40 (B0_P4_R57-64)

Book 0 / Page 4 / Register 65: WCLK1 (Input or Output) Pin Control - 0x00 / 0x04 / 0x41 (B0_P4_R65)

Book 0 / Page 4 / Register 66: Reserved Register - 0x00 / 0x04 / 0x42 (B0_P4_R66)

Book 0 / Page 4 / Register 67: DOUT1 (Output) Pin Control - 0x00 / 0x04 / 0x43 (B0_P4_R67)

Book 0 / Page 4 / Register 68: DIN1 (Input) Pin Control - 0x00 / 0x04 / 0x44 (B0_P4_R68)

Book 0 / Page 4 / Register 68: DIN1 (Input) Pin Control - 0x00 / 0x04 / 0x44 (B0_P4_R68) (continued)

Book 0 / Page 4 / Register 69: WCLK2 (Input or Output) Pin Control - 0x00 / 0x04 / 0x45 (B0_P4_R69)

Book 0 / Page 4 / Register 70: BCLK2 (Input or Output) Pin Control - 0x00 / 0x04 / 0x46 (B0_P4_R70)

Book 0 / Page 4 / Register 71: DOUT2 (Output) Pin Control - 0x00 / 0x04 / 0x47 (B0_P4_R71)

Book 0 / Page 4 / Register 72: DIN2 (Input) Pin Control - 0x00 / 0x04 / 0x48 (B0_P4_R72)

Book 0 / Page 4 / Register 73-74: Reserved Register - 0x00 / 0x04 / 0x49-0x4A (B0_P4_R73-74)

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Book 0 / Page 4 / Register 76: Reserved Register - 0x00 / 0x04 / 0x4C (B0_P4_R76)

Book 0 / Page 4 / Register 77-85: Reserved Registers - 0x00 / 0x04 / 0x4D-0x55 (B0_P4_R77-85)

Book 0 / Page 4 / Register 86: GPIO1 (Input or Output) Pin Control - 0x00 / 0x04 / 0x56 (B0_P4_R86)

Book 0 / Page 4 / Register 87: GPIO2 (Input or Output) Pin Control - 0x00 / 0x04 / 0x57 (B0_P4_R87)

Book 0 / Page 4 / Register 87: GPIO2 (Input or Output) Pin Control - 0x00 / 0x04 / 0x57 **(B0_P4_R87) (continued)**

Book 0 / Page 4 / Register 88: GPIO3 (Input or Output) Pin Control - 0x00 / 0x04 / 0x58 (B0_P4_R88)

Book 0 / Page 4 / Register 89: GPIO4 (Input or Output) Pin Control - 0x00 / 0x04 / 0x59 (B0_P4_R89)

Book 0 / Page 4 / Register 89: GPIO4 (Input or Output) Pin Control - 0x00 / 0x04 / 0x59 **(B0_P4_R89) (continued)**

Book 0 / Page 4 / Register 90: GPIO5 (Input or Output) Pin Control - 0x00 / 0x04 / 0x5A (B0_P4_R90)

Book 0 / Page 4 / Register 91: Reserved Register - 0x00 / 0x04 / 0x5B (B0_P4_R91)

Book 0 / Page 4 / Register 92-95: Reserved Registers - 0x00 / 0x04 / 0x5C-0x5F (B0_P4_R92-95)

Book 0 / Page 4 / Register 96: MISO_GPO1 (Output) Pin Control - 0x00 / 0x04 / 0x60 (B0_P4_R96)

Book 0 / Page 4 / Register 96: MISO_GPO1 (Output) Pin Control - 0x00 / 0x04 / 0x60 (B0_P4_R96) (continued)

Book 0 / Page 4 / Register 97-99: Reserved Registers - 0x00 / 0x04 / 0x61-0x63 (B0_P4_R97-99)

Book 0 / Page 4 / Register 100: Digital Microphone Clock Control - 0x00 / 0x04 / 0x64 (B0_P4_R100)

Book 0 / Page 4 / Register 101: Digital Microphone 1 Input Pin Control - 0x00 / 0x04 / 0x65 (B0_P4_R101)

Book 0 / Page 4 / Register 101: Digital Microphone 1 Input Pin Control - 0x00 / 0x04 / 0x65 **(B0_P4_R101) (continued)**

Book 0 / Page 4 / Register 102: Digital Microphone 2 Input Pin Control - 0x00 / 0x04 / 0x66 (B0_P4_R102)

Book 0 / Page 4 / Register 103: Reserved Register - 0x00 / 0x04 / 0x67 (B0_P4_R103)

Book 0 / Page 4 / Register 104: Bit-Bang Output - 0x00 / 0x04 / 0x68 (B0_P4_R104)

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XAS

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Book 0 / Page 4 / Register 107: Bit-Bang Input - 0x00 / 0x04 / 0x6B (B0_P4_R107)

Book 0 / Page 4 / Register 108-112: Reserved Registers - 0x00 / 0x04 / 0x6C-0x70 (B0_P4_R108-112)

Book 0 / Page 4 / Register 113: Bit-Bang miniDSP Output Control - 0x00 / 0x04 / 0x71 (B0_P4_R113)

Book 0 / Page 4 / Register 114: Reserved Register - 0x00 / 0x04 / 0x72 (B0_P4_R114)

Book 0 / Page 4 / Register 115: ASI1, ADC Bit Clock and ADC Word Clock Output - 0x00 / 0x04 / 0x73 **(B0_P4_R115)**

Book 0 / Page 4 / Register 115: ASI1, ADC Bit Clock and ADC Word Clock Output - 0x00 / 0x04 / 0x73 **(B0_P4_R115) (continued)**

Book 0 / Page 4 / Register 116: ASI2, ADC Bit Clock and ADC Word Clock Output - 0x00 / 0x04 / 0x74 **(B0_P4_R116)**

Book 0 / Page 4 / Register 117: ASI3, ADC Bit Clock and ADC Word Clock Output - 0x00 / 0x04 / 0x75 **(B0_P4_R117)**

Book 0 / Page 4 / Register 118: miniDSP Data Port Control - 0x00 / 0x04 / 0x76 (B0_P4_R118)

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Book 0 / Page 4 / Register 118: miniDSP Data Port Control - 0x00 / 0x04 / 0x76 (B0_P4_R118) (continued)

Book 0 / Page 4 / Register 119: Digital Audio Engine Synchronization Control - 0x00 / 0x04 / 0x77 **(B0_P4_R119)**

Book 0 / Page 4 / Register 120-127: Reserved Registers - 0x00 / 0x04 / 0x78-0x7F (B0_P4_R120-127)

8.6.6 Book 0 Page 252

Book 0 / Page 252 / Register 0: Page Select Register - 0x00 / 0xFC / 0x00 (B0_P252_R0)

Book 0 / Page 252 / Register 1: SAR Buffer Mode Data (MSB) and Buffer Flags - 0x00 / 0xFC / 0x01 **(B0_P252_R1)**

Book 0 / Page 252 / Register 1: SAR Buffer Mode Data (MSB) and Buffer Flags - 0x00 / 0xFC / 0x01 **(B0_P252_R1) (continued)**

Book 0 / Page 252 / Register 2: SAR Buffer Mode Data (LSB) - 0x00 / 0xFC / 0x02 (B0_P252_R2)

Book 0 / Page 252 / Register 3-127: Reserved Registers - 0x00 / 0xFC / 0x03-0x7F (B0_P252_R3-127)

8.6.7 Book 20 Page 0

Book 20 / Page 0 / Register 0: Page Select Register - 0x14 / 0x00 / 0x00 (B20_P0_R0)

Book 20 / Page 0 / Register 1-126: Reserved Registers - 0x14 / 0x00 / 0x01-0x7E (B20_P0_R1-126)

Book 20 / Page 0 / Register 127: Book Selection Register - 0x14 / 0x00 / 0x7F (B20_P0_R127)

8.6.8 Book 20 Page 1-26

Book 20 / Page 1-26 / Register 0: Page Select Register - 0x14 / 0x01-0x1A / 0x00 (B20_P1-26_R0)

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EXAS

Book 20 / Page 1-26 / Register 1-7: Reserved Registers - 0x14 / 0x01-0x1A / 0x01-0x07 (B20_P1-26_R1-7)

Book 20 / Page 1-26 / Register 8-127: ADC Fixed Coefficients C(0:767) - 0x14 / 0x01-0x1A / 0x08-0x7F **(B20_P1-26_R8-127)**

8.6.9 Book 40 Page 0

Book 40 / Page 0 / Register 0: Page Select Register - 0x28 / 0x00 / 0x00 (B40_P0_R0)

Book 40 / Page 0 / Register 1: ADC Adaptive CRAM Configuration Register - 0x28 / 0x00 / 0x01 **(B40_P0_R1)**

Book 40 / Page 0 / Register 2-126: Reserved Registers - 0x28 / 0x00 / 0x02-0x7E (B40_P0_R2-126)

Book 40 / Page 0 / Register 127: Book Selection Register - 0x28 / 0x00 / 0x7F (B40_P0_R127)

8.6.10 Book 40 Page 1-17

Book 40 / Page 1-17 / Register 0: Page Select Register - 0x28 / 0x01-0x11 / 0x00 (B40_P1-17_R0)

Book 40 / Page 1-17 / Register 1-7: Reserved Registers - 0x28 / 0x01-0x11 / 0x01-0x07 (B40_P1-17_R1-7)

Book 40 / Page 1-17 / Register 8-127: ADC Adaptive Coefficients C(0:509) - 0x28 / 0x01-0x11 / 0x08-0x7F **(B40_P1-17_R8-127)**

8.6.11 Book 40 Page 18

Book 40 / Page 18 / Register 0: Page Select Register - 0x28 / 0x12 / 0x00 (B40_P18_R0)

Book 40 / Page 18 / Register 1-7: Reserved Registers - 0x28 / 0x12 / 0x01-0x07 (B40_P18_R1-7)

Book 40 / Page 18 / Register 8-15: ADC Adaptive Coefficients C(510:511) - 0x28 / 0x12 / 0x08-0x0F (B40_P18_R8-15)

Book 40 / Page 18 / Register 16-127: Reserved Registers - 0x28 / 0x12 / 0x10-0x7F (B40_P18_R16-127)

8.6.12 Book 60 Page 0

Book 60 / Page 0 / Register 0: Page Select Register - 0x3C / 0x00 / 0x00 (B60_P0_R0)

Book 60 / Page 0 / Register 1-126: Reserved Registers - 0x3C / 0x00 / 0x01-0x7E (B60_P0_R1-126)

Book 60 / Page 0 / Register 127: Book Selection Register - 0x3C / 0x00 / 0x7F (B60_P0_R127)

8.6.13 Book 60 Page 1-35

Book 60 / Page 1-35 / Register 0: Page Select Register - 0x3C / 0x01-0x23 / 0x00 (B60_P1-35_R0)

Book 60 / Page 1-35 / Register 1-7: Reserved Registers - 0x3C / 0x01-0x23 / 0x01-0x07 (B60_P1-35_R1-7)

Book 60 / Page 1-35 / Register 8-127: DAC Fixed Coefficients C(0:1023) - 0x3C / 0x01-0x23 / 0x08-0x7F **(B60_P1-35_R8-127)**

8.6.14 Book 80 Page 0

Book 80 / Page 0 / Register 0: Page Select Register - 0x50 / 0x00 / 0x00 (B80_P0_R0)

Book 80 / Page 0 / Register 1: DAC Adaptive Coefficient Bank 1 Configuration Register - 0x50 / 0x00 / **0x01 (B80_P0_R1)**

Book 80 / Page 0 / Register 2-126: Reserved Registers - 0x50 / 0x00 / 0x02-0x7E (B80_P0_R2-126)

Book 80 / Page 0 / Register 127: Book Selection Register - 0x50 / 0x00 / 0x7F (B80_P0_R127)

8.6.15 Book 80 Page 1-17

Book 80 / Page 1-17 / Register 0: Page Select Register - 0x50 / 0x01-0x11 / 0x00 (B80_P1-17_R0)

Book 80 / Page 1-17 / Register 1-7: Reserved Registers - 0x50 / 0x01-0x11 / 0x01-0x07 (B80_P1-17_R1-7)

Book 80 / Page 1-17 / Register 8-127: DAC Adaptive Coefficient Bank 1 C(0:509) - 0x50 / 0x01-0x11 / 0x08-**0x7F (B80_P1-17_R8-127)**

8.6.16 Book 80 Page 18

Book 80 / Page 18 / Register 0: Page Select Register - 0x50 / 0x12 / 0x00 (B80_P18_R0)

Book 80 / Page 18 / Register 1-7: Reserved Registers - 0x50 / 0x12 / 0x01-0x07 (B80_P18_R1-7)

Book 80 / Page 18 / Register 8-15: DAC Adaptive Coefficient Bank 1 C(510:511) - 0x50 / 0x12 / 0x08-0x0F **(B80_P18_R8-15)**

Book 80 / Page 18 / Register 16-127: Reserved Registers - 0x50 / 0x12 / 0x10-0x7F (B80_P18_R16-127)

8.6.17 Book 82 Page 0

Book 82 / Page 0 / Register 0: Page Select Register - 0x52 / 0x00 / 0x00 (B82_P0_R0)

Book 82 / Page 0 / Register 1: DAC Adaptive Coefficient Bank 2 Configuration Register - 0x52 / 0x00 / **0x01 (B82_P0_R1)**

Book 82 / Page 0 / Register 2-126: Reserved Registers - 0x52 / 0x00 / 0x02-0x7E (B82_P0_R2-126)

Book 82 / Page 0 / Register 127: Book Selection Register - 0x52 / 0x00 / 0x7F (B82_P0_R127)

8.6.18 Book 82 Page 1-17

Book 82 / Page 1-17 / Register 0: Page Select Register - 0x52 / 0x01-0x11 / 0x00 (B82_P1-17_R0)

Book 82 / Page 1-17 / Register 1-7: Reserved Registers - 0x52 / 0x01-0x11 / 0x01-0x07 (B82_P1-17_R1-7)

Book 82 / Page 1-17 / Register 8-127: DAC Adaptive Coefficient Bank 2 C(0:509) - 0x52 / 0x01-0x11 / 0x08-**0x7F (B82_P1-17_R8-127)**

8.6.19 Book 82 Page 18

Book 82 / Page 18 / Register 0: Page Select Register - 0x52 / 0x12 / 0x00 (B82_P18_R0)

Book 82 / Page 18 / Register 1-7: Reserved Registers - 0x52 / 0x12 / 0x01-0x07 (B82_P18_R1-7)

Book 82 / Page 18 / Register 8-15: DAC Adaptive Coefficient Bank 2 C(510:511) - 0x52 / 0x12 / 0x08-0x0F **(B82_P18_R8-15)**

EXAS NSTRUMENTS

Book 82 / Page 18 / Register 16-127: Reserved Registers - 0x52 / 0x12 / 0x10-0x7F (B82_P18_R16-127)

8.6.20 Book 100 Page 0

Book 100 / Page 0 / Register 0: Page Select Register - 0x64 / 0x00 / 0x00 (B100_P0_R0)

Book 100 / Page 0 / Register 1-46: Reserved Registers - 0x64 / 0x00 / 0x01-0x2E (B100_P0_R1-46)

Book 100 / Page 0 / Register 47: Non-Programmable Override Options - 0x64 / 0x00 / 0x2F (B100_P0_R47)

Book 100 / Page 0 / Register 48: ADC miniDSP_A Instruction Control Register 1 - 0x64 / 0x00 / 0x30 **(B100_P0_R48)**

Book 100 / Page 0 / Register 49: ADC miniDSP_A Instruction Control Register 2 - 0x64 / 0x00 / 0x31 **(B100_P0_R49)**

Book 100 / Page 0 / Register 50: ADC miniDSP_A CIC Input and Decimation Ratio Control Register - 0x64 **/ 0x00 / 0x32 (B100_P0_R50)**

Book 100 / Page 0 / Register 51-59: Reserved Registers - 0x64 / 0x00 / 0x33-0x3B (B100_P0_R51-59)

Book 100 / Page 0 / Register 60: ADC miniDSP_A Secondary CIC Input Control - 0x64 / 0x00 / 0x3C **(B100_P0_R60)**

Book 100 / Page 0 / Register 61: miniDSP_A to Audio Serial Interface Handoff Control - 0x64 / 0x00 / 0x3D **(B100_P0_R61)**

Book 100 / Page 0 / Register 61: miniDSP_A to Audio Serial Interface Handoff Control - 0x64 / 0x00 / 0x3D **(B100_P0_R61) (continued)**

Book 100 / Page 0 / Register 62-126: Reserved Registers - 0x64 / 0x00 / 0x3E-0x7E (B100_P0_R62-126)

Book 100 / Page 0 / Register 127: Book Selection Register - 0x64 / 0x00 / 0x7F (B100_P0_R127)

8.6.21 Book 100 Page 1-52

Book 100 / Page 1-52 / Register 0: Page Select Register - 0x64 / 0x01-0x34 / 0x00 (B100_P1-52_R0)

Book 100 / Page 1-52 / Register 1-7: Reserved Registers - 0x64 / 0x01-0x34 / 0x01-0x07 (B100_P1-52_R1- 7)

Book 100 / Page 1-52 / Register 8-127: miniDSP_A Instructions - 0x64 / 0x01-0x34 / 0x08-0x7F (B100_P1- 52_R8-127)

8.6.22 Book 120 Page 0

Book 120 / Page 0 / Register 0: Page Select Register - 0x78 / 0x00 / 0x00 (B120_P0_R0)

Book 120 / Page 0 / Register 47: Non-Programmable Override Options - 0x78 / 0x00 / 0x2F (B120_P0_R47)

Book 120 / Page 0 / Register 48: DAC miniDSP_D Instruction Control Register 1 - 0x78 / 0x00 / 0x30 **(B120_P0_R48)**

Book 120 / Page 0 / Register 49: DAC miniDSP_D Instruction Control Register 2 - 0x78 / 0x00 / 0x31 **(B120_P0_R49)**

Book 120 / Page 0 / Register 50: DAC miniDSP_D Interpolation Factor Control Register - 0x78 / 0x00 / **0x32 (B120_P0_R50)**

Book 120 / Page 0 / Register 50: DAC miniDSP_D Interpolation Factor Control Register - 0x78 / 0x00 / **0x32 (B120_P0_R50) (continued)**

Book 120 / Page 0 / Register 51-126: Reserved Registers - 0x78 / 0x00 / 0x33-0x7E (B120_P0_R51-126)

Book 120 / Page 0 / Register 127: Book Selection Register - 0x78 / 0x00 / 0x7F (B120_P0_R127)

8.6.23 Book 120 Page 1-103

Book 120 / Page 1-103 / Register 0: Page Select Register - 0x78 / 0x01-0x67 / 0x00 (B120_P1-103_R0)

Book 120 / Page 1-103 / Register 1-7: Reserved Registers - 0x78 / 0x01-0x67 / 0x01-0x07 (B120_P1- 103_R1-7)

Book 120 / Page 1-103 / Register 8-127: miniDSP_D Instructions - 0x78 / 0x01-0x67 / 0x08-0x7F (B120_P1- 103_R8-127)

8.6.24 ADC Coefficients

Table 53. ADC Fixed Coefficient Map

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Table 54. ADC Adaptive Coefficient Buffer-A Map (continued)

Table 55. ADC Adaptive Coefficient Buffer-B Map (continued)

8.6.25 ADC Defaults

Table 56. Default values of ADC Coefficients in Buffers A and B

8.6.26 DAC Coefficients

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Table 57. DAC Fixed Coefficient Map (continued)

Table 58. DAC Adaptive Coefficient Bank 1 Buffer-A Map

Table 59. DAC Adaptive Coefficient Bank 1 Buffer-B Map

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Table 60. DAC Adaptive Coefficient Bank 2 Buffer-A Map

Table 60. DAC Adaptive Coefficient Bank 2 Buffer-A Map (continued)

Table 61. DAC Adaptive Coefficient Bank 2 Buffer-B Map

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8.6.27 DAC Defaults

Table 62. Default values of DAC Coefficients Bank 1 and Bank 2 in Buffers A and B

9 Applications and Implementation

9.1 Application Information

The TLV320AIC3268 is a highly integrated stereo audio codec with integrated miniDSP, mono Speaker amplifier and mutiple digital audio interfaces. It enables many different types of audio platforms having a need for stereo audio record and playback and needing to interface with several other devices in the system over multiple digital audio interfaces.

9.2 Typical Applications

[Figure](#page-242-0) 109 shows a typical circuit configuration for a system utilizing the TLV320AIC3268.

9.2.1 Design Requirements

9.2.1.1 Charge Pump Flying and Holding Capacitor

The TLV320AIC3268 features a built in charge-pump to generate a negative supply rail, VNEG from CPVDD_18. This negative voltage is used by the headphone amplifier to enable driving the output signal biased around ground potential. For proper operation of the charge pump and headphone amplifier, it is recommended that the flying capacitor connected between CPFCP and CPFCM terminals and the holding capacitor connected between VNEG and ground be of X7R type. It is recommended to use 2.2µF as capacitor values. Failure to use X7R type capacitor can result in degraded performance of charge pump and headphone amplifier.

9.2.1.2 Reference Fitlering Capacitor

The TLV320AIC3268 has a built-in bandgap used to generate reference voltages and currents for the device. To acheive high SNR, the reference voltage on VREF AUDIO should be filtered using a 1µF capacitor from VREF_AUDIO terminal to ground.

The built-in SAR ADC in TLV320AIC3268 can operate with either internally generated reference voltage or externally provided reference voltage. When used with internal reference, the reference voltage on VREF_SAR should be filtered with a 1µF capacitor for filtering noise as well as reference stability.

9.2.1.3 Micbias

The TLV320AIC3268 has a built-in bias voltage output for biasing of microphones. No intentional capacitors should be connected directly to either MICBIAS or MICBIAS_EXT outputs for filtering.

9.2.1.4 Power Supply

The TLV320AIC3268 needs several power supplies for it's operation.

The SVDD input is used to power the speaker amplifier. This includes the power required for internal circuits as well as the power delivered to the speaker load. It is recommended that this supply is directly connected to battery, for systems where battery rail is in the range of 2.7V to 5.5V.The peak switching current on this power can exceed 1A. MICBIAS_VDD can be tapped from same source that is used for SVDD.

The AVDDx_18, HVDD_18 and CPVDD_18 power inputs are used to power the analog circuits including analog to digital converters, digital to analog converters, programmable gain amplifiers, headphone amplifiers, charge pump etc. The analog blocks in TLV320AIC3268 have high power supply rejection ratio, however it is recommended that these supplies be powered by well regulated power supplies like low dropout regulators (LDO) for optimal performance. When these power terminals are driven from a common power source, the current drawn from the source will depend upon blocks enabled inside the device. However as an example when all the internal blocks powered are enabled the source should be able to deliver 150mA of current.

The RECVDD 33 powers the receiver amplifier of TLV320AIC3268. When the receiver amplifier is used in differential lineout mode, then the RECVDD 33 could be connected to the same supply as AVDDx 18 terminals. When the receiver amplifier is used to drive mono receiver speakers in BTL mode, the RECVDD_33 supply could consume approximately 150mA of peak current.

The DVDD_18 powers the digital core of TLV320AIC3268, including the minIDSP, audio serial interfaces (ASI), control interfaces (SPI or I2C), clock generation and PLL. The DVDD_18 power can be driven by high efficiency switching regulators or low drop out regulators. When the miniDSP_a and miniDSP_D are enabled in programmable mode and operated at peak frequencies, the supply source should be able to able to deliver approx 100mA of current. When the PRB modes are used instead of programmable miniDSP mode, then the peak current load on DVDD_18 supply source could be approximately 20mA.

The IOVDD1_33 and IOVDD2_33 power the digital input and digital output buffers of TLV320AIC3268. The current consumption of this power depends on configuration of digital terminals as inputs or outputs. When the digital terminals are configured as outputs, the current consumption would depend on switching frequency of the signal and the load on the output terminal, which depends on board design and input capacitance of other devices connected to the signal .

Refer to [Figure](#page-242-0) 109 for recommendations on decoupling capacitors.

where g is the analog PGA gain calculated in linear terms.

9.2.2 Detailed Design Procedure

9.2.2.1 Analog Input Connection

The analog inputs to TLV320AIC3268 should be ac-coupled to the device terminals to allow decoupling of signal source's common mode voltage with that of TLV320AIC3268's common mode voltage. The input coupling capacitor in combination with the selected input impedance of TLV320AIC3268 forms a high-pass filter.

$$
F_c = 1/(2^* \pi^* R_{eq} C_c)
$$
\n
$$
C_c = 1/(2^* \pi^* R_{eq} F_c)
$$
\n(22)

For high fidelity audio recording application it is desirable to keep the cutoff frequency of the high pass filter as low as possible. For single-ended input mode, the equivalent input resistance R_{eq} can be calculated as

$$
(24)
$$

floor(G/6)/ R_{in} (25)

 $R_{eq} = R_{in}^{*} (1 + 2g)/(1+g)$

 $g = 10000 * 2^{floor(G/6)}/R_{in}$

where G is the analog PGA gain programmed in B0_P1_R59-60 (in dB) and R_{in} is the value of the resistor programmed in B0_P1_R52-53 and B0_P1_R55-56 and assumes $R_{in} = R_{cm}$ (as defined in B0_P1_R54 and BO_P1_R57). For differential input mode, R_{eq} can be calculated as:

 $\mathsf{R}_{\text{eq}} = \mathsf{R}_{\text{in}}$ (26)

where R_{in} is the value of the resistor programmed in B0_P1_R52-54 and B0_P1_R55-57, assuming symmetrical inputs.

Figure 110. Analog Input Connection With Pull-down Resistor

When the analog signal is connected to the system through a connector such as audio jack, it is recommended to put a pull-down resistor on the signal as shown in [Figure](#page-244-0) 111. The pulldown resistor helps keep the signal grounded and helps improve noise immunity when no source is connected to the connector. The pulldown resistor value should be chosen large enough to avoid loading of signal source.

Each analog input of the TLV320AIC3268 is capable of handling signal amplitude of 0.5Vrms. If the input signal source can drive signals higher than the maximum value, an external resistor divider network as shown in [Figure](#page-244-0) 111 should be used to attenuate the signal to less than 0.5Vrms before connecting the signal to the device. The resistor values of the network should be chosen to provide desired attenuation as well as [Equation](#page-244-1) 27.

 $R_1|| R_2 << R_{eq}$

Figure 111. Analog Input Connection With Resistor Divider Network

Certain non audio applications require supporting of high impedance sources. The TLV320AIC3268 supports a high-input impedance mode on IN1L/AUX1 and IN1R/AUX2 terminals, for such use cases. See High [Impedance](#page-42-0) Input [Mode](#page-42-0) for more details. For such cases the input coupling capacitor can be eliminated from design.

Whenever any of the analog input terminals IN1L_AUX1, IN2L, IN3L, IN4L, IN1R_AUX2, IN2R, IN3R or IN4R are not used in an application, it is recommended to short the unused input terminals together and connect them to ground using a small capacitor (example 0.1µF). If VBAT is not used in an application, then it can be shorted to ground.

9.2.2.2 Analog Output Connection

The line outputs of the TLV320AIC3268 drive a signal biased around the device common mode voltage. To avoid loading the common mode with the load, it is recommended to connect the single-ended load through an accoupling capacitor. The ac-coupling capacitor in combination with the load impedance forms a high pass filter.

 $F_c = 1/(2[*]π[*]R_LC_c)$ $C_c = 1/(2[*]π[*]R_LF_c)$

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For high fidelity playback, the cutoff frequency of the resultant high-pass filter should be kept low. For example with R_L of 10kΩ, using 1µF coupling capacitor results in a cut-off frequency of 8Hz.

For differential lineout configurations, the load should be directly connected between the differential outputs, with no coupling capacitor.

The TLV320AIC3268 supports headphone in single-ended configuration and drives the signal biased around ground. The headphone load can be directly connected between device terminals and ground.

The TLV320AIC3268 supports BTL load for receiver and speaker amplifier. The load can be directly connected between the differential output terminals of receiver and speaker amplifier.

Whenver any of the analog output terminals LOL, LOR, RECP, RECM, HPL, HPR, SPKP or SPKM are not used in an application, they should be left open or not connected.

9.2.2.3 EMI Passive Devices

The TLV320AIC3268 does not need filters for the functioning of it's speaker amplifier.

However depending on the board layout and system level EMI requirements, an optional EMI filter consisting of ferrite beads and capacitors may be used as shown in and and labelled as optional. The ferrite beads if used should be capable of handing peak currents for desired power delivery into speakers. If EMI passives are not used, ferrite bead should be replaced with a short and the capacitor not installed.

9.2.3 Application Performance Plots

[Figure](#page-245-0) 112 shows the excellent low-distortion performance of the TLV320AIC3268 in a system over the 20Hz to 20kHz audio spectrum.

Conditions: Differential Lineout, R_{load}=10kΩ, CM=0.9V, Input Amplitude=-3dBFS

Figure 112. Total Harmonic Distortion + Noise versus Input Frequency

[Figure](#page-246-0) 113 shows the distortion performance of the TLV320AIC3268 in a system over the input amplitude range.

Conditions: Differential Lineout, R_{load}=10kΩ, CM=0.9V

Figure 113. Total Harmonic Distortion + Noise versus Input Amplitude

9.3 Initialization Setup

The TLV320AIC3268 has multiple power supply domains to support various functions. Before the device is fully functional, all the power supplies must be enabled. The section Power On [Sequence](#page-246-1) describes the supported power supply sequencing for TLV320AIC3268.

9.3.1 Power On Sequence

There are two recommended power sequence possible for TLV320AIC3268:

1) Speaker/Microphone Supplies, then Digital Supplies, then Analog Supplies

2) Speaker/Microphone Supplies, then Digital and Analog Supplies

The first power on sequence is useful if the end system uses separate analog and digital supplies. This is useful to improve the efficiency of the digital rails by using a DC/DC converter, while keeping the analog supplies clean by using a low-dropout regulator (LDO). While it is recommended to separate analog and digital supplies, if all the 1.8V supplies (analog and digital) must be tied together, the second power sequence can be utilized.

9.3.1.1 Power On Sequence 1 - Separate Digital and Analog Supplies

[Figure](#page-247-0) 114 shows a timing diagram for the case where all supplies are provided separately. In such case, the depicted sequence should be used. The dashed lines marked in blue color refer to an internally supplied voltage.

Initialization Setup (continued)

Figure 114. Analog Supplies provided after Digital Supplies

SVDD, SPK_V, and MICBIAS_VDD should be provided first. Next, IOVDD1_33 and IOVDD2_33 should be provided, and DVDD_18 can be provided at the same time as these IOVDDx_33 supplies. Since, by default, DVDD_18 is weakly connected to AVDD1_18 by a 10kΩ resistor, AVDDx_18 and HVDD_18 (it is recommended to connect these five supplies together) will ramp up to the DVDD_18 voltage once DVDD_18 is provided at approximately 5*10k*C_{AVDD} seconds, where C_{AVDD} are the sum of the decoupling capacitors on the AVDDx_18 and HVDD_18 terminals. For C_{AVDD} = 1uF, the charging time is approximately 50ms. Parameter t_{DA} allows analog supplies to be stable before analog supplies are provided. To prevent high currents from DVDD_18 to the 1.8V and 3.3V analog supplies (that is AVDDx_18, HVDD_18, RECVDD_33), these analog supplies cannot be externally driven low by the external power source. This means that the external power source should be either high impedance or have a weak pull-down before being enabled.

Ensure that CPVDD_18 supply is provided either at same time as analog supplies or later. After RESET is released (or a software reset is performed), no register writes should be performed within 1ms.

Table 63. Power Supply Timing Parameters (continued)

9.3.1.2 Power On Sequence 2 - Shared 1.8V Analog Supplies

If desired, the analog supplies (AVDD1_18, AVDD2_18, AVDD4_18, AVDD_18 and HVDD_18) could also be externally supplied at the same time as DVDD_18. In this case, the weak pullup is not utilized. This is shown in [Figure](#page-248-0) 115.

Figure 115. Simultaneous Powerup of Digital and Analog 1.8V Supplies

After RESET is released (or a software reset is performed), no register writes should be performed within 1ms.

Table 64. Power Supply Timing Parameters

Table 64. Power Supply Timing Parameters (continued)

When the TLV320AIC3268 is controlled over a shared I^2C bus, it is recommended that any I^2C communication be attempted only after the IOVDDx_33 and DVDD_18 rails have been powered up and the hardware reset signal applied. If ^{[2}C communication is attempted before this requirement, even with other devices on the same I²C bus, error conditions may occur.

9.3.2 Reset

The TLV320AIC3268 internal logic must be initialized to a known condition for proper device function. To initialize the device in its default operating condition, the hardware reset terminal (RESET) must be pulled low for at least 10ns. For this initialization to work, both the IOVDDx_33 and DVDD_18 supplies must be powered up. It is recommended that while the DVDD 18 supply is being powered up, the RESET terminal be pulled low.

The device can also be reset via software reset. Writing '1' into B0_P0_R1_D0 resets the device. After a device reset, all registers are initialized with default values as listed in the Register Map section.

After hardware or software reset, the TLV320AIC3268 is in a sleep mode, when all the blocks and functions are disabled. In this mode however the control interface, SPI or I²C remains enabled, allowing an external host controller to configure the device. The blocks inside the TLV320AIC3268 must be configured for desired function by writing into the control registers.

Given below are some of the typical configurations that are needed to be done before enabling TLV320AIC3268. However the list of configurations is not unique or comprehensive and the requirements must be determined per the application requirements.

- Wait for Lockout time. See Device Startup [Lockout](#page-249-0) Times for details.
- Configuration of FIFO's. See [Setting](#page-250-0) Device FIFOs for details.
- Configure the device for PRB modes or programmable miniDSP modes for ADC and DAC.
	- For PRB modes, program the coefficients of user programmable filters, including setup adaptive coefficient updates if necessary.
	- For programmable miniDSP mode, configure the instruction and coefficient memory for miniDSP.
- Configure the oversampling rates, AOSR and DOSR, and power tune modes for ADCs and DACs.
- Configure the clock dividers and PLL, if required. See Clock [Generation](#page-63-0) and PLL for details.
- Configure the Audio Serial Interfaces (ASI).
- Configure the device common mode voltages. See Setting Device [Common](#page-250-1) Mode Voltage for details.
- Configure the analog signal routing to ADCs and from DACs, volume controls and power up the relevant blocks.

Some device configurations require wait times after register configuration, to take effect. See [Analog](#page-250-2) and [Reference](#page-250-2) Startup and PLL [Startup](#page-250-3) for details. [Table](#page-90-0) 42 describes internal flags which can be used to check device status in response to device configurations or input signals.

The details of control registers for device configurations are described in [Register](#page-120-0) Maps.

9.3.3 Device Startup Lockout Times

After the TLV320AIC3268 is initialized through hardware reset at power-up or software reset, the internal registers are initialized to default values. This initialization takes place within 1ms after pulling the RESET signal high. During this initialization phase, no register-read or register-write operation should be performed on ADC or DAC coefficient buffers. Also, no block within the codec should be powered up during the initialization phase.

9.3.4 Analog and Reference Startup

The TLV320AIC3268 uses an external VREF_AUDIO terminal for decoupling the reference voltage used for the data converters and other analog blocks. VREF_AUDIO terminal requires a minimum 1µF decoupling capacitor from VREF_AUDIO to VSS. In order for any analog block to be powered up, the Analog Reference block must be powered up. By default, the Analog Reference block will implicitly be powered up whenever any analog block is powered up, or it can be powered up independently. Detailed descriptions of Analog Reference including fast power-up options are provided in [Reference](#page-74-0) Voltage. During the time that the reference block is not completely powered up, subsequent requests for powering up analog blocks (for example PLL) are queued, and executed after the reference power up is complete.

When analog inputs are routed to the ADC PGA, approximately 2ms of wait is required to enable charging up of input coupling capacitor before the routing is enabled.

9.3.5 PLL Startup

Whenever the PLL is powered up, a startup delay of approx 1ms is involved after the power up command of the PLL and before the clocks are available to the codec. This delay is to ensure stable operation of PLL and clockdivider logic.

9.3.6 Setting Device Common Mode Voltage

The TLV320AIC3268 allows the user to set the common mode voltage for analog inputs to 0.75V or 0.9V by programming B0_P1_R8_D2. The input common-mode voltage of 0.9V works best when the analog supply voltage is centered around 1.8V or above, and offers the highest possible performance. For analog supply voltages below 1.8V, a common mode voltage of 0.75V must be used.

Table 65. Input Common Mode voltage and Input Signal Swing

NOTE

The input common mode setting is common for DAC playback and Analog Bypass path

9.3.7 Setting Device FIFOs

The TLV320AIC3268 features FIFO between CIC filters and miniDSP_A and miniDSP_D and DAC modulators. These FIFO's allow ease of programming of signal processing functions within the miniDSP by allowing easier interface with other blocks. However use of these FIFOs adds to group delay in the channel and should be bypassed for delay sensitive applications along with appropriate care taken in miniDSP programming. By default the FIFO between CIC filter and miniDSP_A is disabled and is recommended to be enabled by writing B64 P0_R32_D7 as '1'. Similarly by default the FIFO between miniDSP_D and DAC modulator is disabled and is recommended to be enabled by writing B120_P0_R50_D7 as '1'.

The TLV320AIC3268 has a feature which allows the miniDSP instruction frame to be synchronized with ASI's data frame. For miniDSP D, the synchronization is done with ASI1's data frame by default. However when the TLV320AIC3268 is used with PRB modes for playback, it is recommended to disable miniDSP_D's synchronization with ASI data frame by programming B0_P4_R119_D[7:6] as "11".

9.3.8 Miscellaneous

When enabling headphone amplifiers, it is recommended to program B0_P1_R77_D0 as '1' before powering up headphone amplifier.

10 Power Supply Recommendations

See Power [Supply](#page-243-0) for details about driving power supplies. See Power On [Sequence](#page-246-1) for details on sequencing of power supplies.

11 Layout

11.1 Layout Guidelines

Each system design and PCB layout is unique; layout should be carefully reviewed in the context of a specific PCB design. However, the following guidelines can optimize TLV320AIC3268 performance:

- The TLV320AIC3268 thermal pad also serves as a device ground connection. Connect the thermal pad to the ground plane using multiple VIAS to minimize impedance between device ground and PCB ground. The TLV320AIC3268 has only one ground terminal. The digital ground and analog ground planes on the PCB should be shorted to each other close the TLV320AIC3268 thermal pad.
- The decoupling capacitors for the power supplies should be placed close to the device terminals. [Figure](#page-242-0) 109 shows the recommended decoupling capacitors for the TLV320AIC3268. For SVDD, place the 10µF bulk decoupling capacitor near where battery supply enters the PCB, and place the smaller 0.1µF and 1µF decoupling capacitors closer to the device terminals.
- Place the flying capacitor between CPFCP and CPFCM near the device terminals, with no VIAS between the device terminals and the capacitor. Similarly, keep the decoupling capacitor on VNEG near the device terminal with minimal VIAS between the device terminals, capacitor and PCB ground.
- The TLV320AIC3268 internal voltage references must be filtered using external capacitors. Place the filter capacitors on VREF_AUDIO and VREF_SAR near the device terminals for optimal performance.
- The TLV320AIC3268 reduces crosstalk by a separate ground sense signal for the headphone jack. To optimize crosstalk performance, use a separate trace from the HPVSS_SENSE terminal to the headphonejack ground terminal, with no other ground connections along the length.
- The parasitic capacitance to ground plane should be minimized for SPKP and SPKM signals. Keep the signal routing for SPKP and SPKM as short as possible for optimal performance of the speaker amplifier.
- For analog differential audio signals, the signals should be routed differentially on the PCB for better noise immunity. Avoid crossing of digital and analog signals to avoid undesirable crosstalk.

11.2 Layout Example

12 Device and Documentation Support

12.1 Community Resources

E2E™ Audio [Converters](http://e2e.ti.com/support/data_converters/audio_converters/f/64.aspx) Forum

TI E2E [Community](http://e2e.ti.com/)

12.2 Trademarks

E2E is a trademark of Texas Instruments, Inc..

12.3 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.4 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — *TI Glossary*.

This glossary lists and explains terms, acronyms and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, see the left-hand navigation.

PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check<http://www.ti.com/productcontent>for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

Texas
Instruments

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

TEXAS
INSTRUMENTS

PACKAGE MATERIALS INFORMATION

www.ti.com 11-Apr-2015

*All dimensions are nominal

MECHANICAL DATA

NOTES: All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5-1994. A.

- Β. This drawing is subject to change without notice.
- Quad Flatpack, No-leads (QFN) package configuration. \mathbb{C} .
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

NOTE: A. All linear dimensions are in millimeters

RGC (S-PVQFN-N64)

PLASTIC QUAD FLATPACK NO-LEAD

NOTES:

- A All linear dimensions are in millimeters.
	- This drawing is subject to change without notice. В.
	- C. Publication IPC-7351 is recommended for alternate designs.
	- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <http://www.ti.com>.
	- Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should E. contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
	- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in thermal pad.

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