

# **TLE94106EL**





#### **Features**

- Six half bridge power outputs
- Very low power consumption in sleep mode
- 3.3V / 5V compatible inputs with hysteresis
- · All outputs with overload and short circuit protection
- Independently diagnosable outputs (overcurrent, open load)
- · Open load diagnostics in ON-state for all high-side and low-side
- Outputs with selectable open load thresholds (HS1, HS2)
- 16-bit Standard SPI interface with daisy chain and in-frame response capability for control and diagnosis
- · Fast diagnosis with the global error flag
- PWM capable outputs for frequencies 80Hz, 100Hz and 200Hz with 8-bit duty cycle resolution
- · Overtemperature pre-warning and protection
- · Over- and Undervoltage lockout
- Cross-current protection

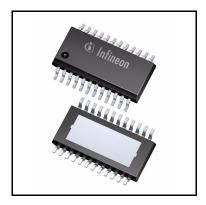
# **Applications**

- HVAC Flap DC motors
- · Monostable and bistable Relays
- Side mirror x-y adjustment and mirror fold
- LEDs

# **Description**

The TLE94106EL is a protected six-fold half-bridge driver designed especially for automotive motion control applications such as Heating, Ventilation and Air Conditioning (HVAC) flap DC motor control. It is part of a larger family offering half-bridge drivers from three outputs to twelve outputs with direct interface or SPI interface.

The half bridge drivers are designed to drive DC motor loads in sequential or parallel operation. Operation modes forward (cw), reverse (ccw), brake and high impedance are controlled from a 16-bit SPI interface. It offers diagnosis features such as short circuit, open load, power supply failure and overtemperature detection. In combination with its low quiescent current, this device is attractive among others for automotive applications. The small fine pitch exposed pad package, PG-SSOP-24, provides good thermal performance and reduces PCB-board space and costs.





Туре	Package	Marking
TLE94106EL	PG-SSOP-24	TLE94106EL

# Table 1 Product Summary

Normal Operating Voltage	$V_{S}$	5.5 18 V
Extended Operating Voltage	$V_{S(EXT)}$	18 20 V
Logic Supply Voltage	$V_{DD}$	3.0 5.5 V
Maximum Supply Voltage for Load Dump Protection	$V_{S(LD)}$	40 V
Minimum Overcurrent Threshold	I <sub>SD</sub>	0.9 A
Maximum On-State Path Resistance at T <sub>j</sub> = 150°C	R <sub>DSON(total)_HSx+LSy</sub>	1.8 + 1.8 Ω
Typical Quiescent Current at T <sub>j</sub> = 85°C	I <sub>SQ</sub>	0.1 μΑ
Maximum SPI Access Frequency	f <sub>SCLK</sub>	5 MHz



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#### **Pin Configuration**

# **1** Pin Configuration

# 1.1 Pin Assignment

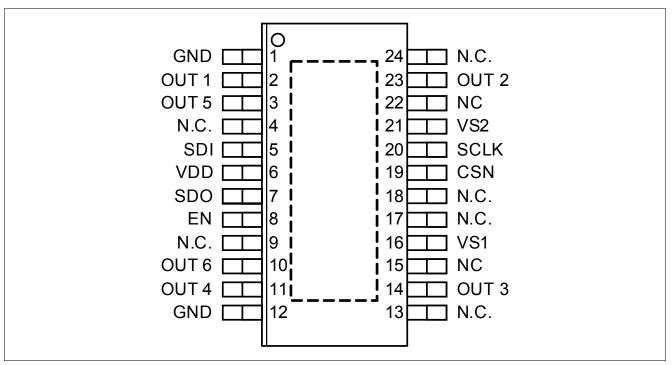


Figure 1 Pin Configuration TLE94106EL

# 1.2 Pin Definitions and Functions

Pin	Symbol	Function
1	GND	Ground. All ground pins should be externally connected together.
2	OUT 1	Power half-bridge 1
3	OUT 5	Power half-bridge 5
4	N.C.	Not connected. This pin should either be left open or terminated to ground.
5	SDI	Serial data input with internal pull down
6	VDD	Logic supply voltage
7	SDO	Serial data output
8	EN	Enable with internal pull-down; Places device in standby mode by pulling the EN line Low
9	N.C.	Not connected. This pin must either be left open or terminated to ground
10	OUT 6	Power half-bridge 6
11	OUT 4	Power half-bridge 4
12	GND	Ground. All ground pins should be externally connected together.

#### **TLE94106EL**



#### **Pin Configuration**

Pin	Symbol	Function
13	N.C.	Not connected. This pin should either be left open or terminated to ground (e.g. for layout compatibility with TLE94112EL, TLE94110EL or TLE94108EL).
14	OUT 3	Power half-bridge 3
15	N.C.	Not connected. This pin should either be left open or terminated to ground.
16	VS1	Main supply voltage for power half bridges. VS1 should be externally connected to VS2.
17	N.C.	Not connected. This pin should either be left open or terminated to ground.
18	N.C.	Not connected. This pin should either be left open or terminated to ground.
19	CSN	Chip select Not input with internal pull up
20	SCLK	Serial clock input with internal pull down
21	VS2	Main supply voltage for power half bridges. VS1 should be externally connected to VS2.
22	N.C.	Not connected. This pin should either be left open or terminated to ground.
23	OUT 2	Power half-bridge 2
24	N.C.	Not connected. This pin should either be left open or terminated to ground (e.g. for layout compatibility with TLE94112EL, TLE94110EL or TLE94108EL).
EDP	-	Exposed Die Pad; For cooling and EMC purposes only - not usable as electrical ground. Electrical ground must be provided by pins 1,12.1)

<sup>1)</sup> The exposed die pad at the bottom of the package allows better heat dissipation from the device via the PCB. The exposed pad (EP) must be either left open or connected to GND. It is recommended to connect EP to GND for best EMC and thermal performance.



#### **Block Diagram**

# 2 Block Diagram

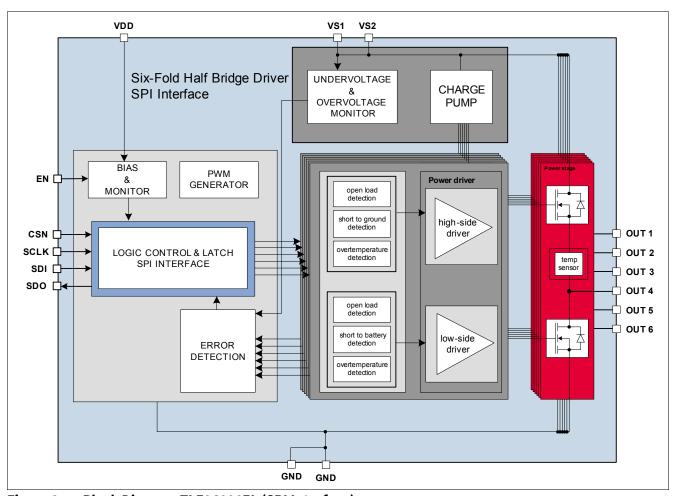


Figure 2 Block Diagram TLE94106EL (SPI Interface)



#### **Block Diagram**

# 2.1 Voltage and current definition

Figure 3 shows terms used in this datasheet, with associated convention for positive values.

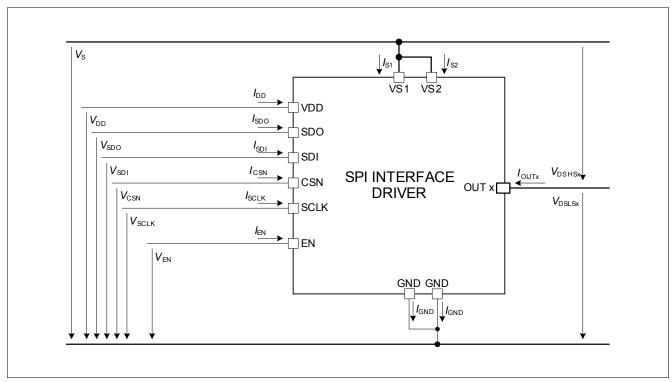


Figure 3 Voltage and Current Definition



# 3 General Product Characteristics

# 3.1 Absolute Maximum Ratings

Table 2 Absolute Maximum Ratings<sup>1)</sup> $T_i = -40$ °C to +150°C

Parameter	Symbol		Value	s	Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Voltages	- 1	- 1	<b>"</b>	11	T.		
Supply voltage	V <sub>S</sub>	-0.3	_	40	V	$V_{\rm S} = V_{\rm S1} = V_{\rm S2}$	P_4.1.1
Supply Voltage Slew Rate	dV <sub>s</sub> /dt	-	-	10	V/µs	V <sub>S</sub> increasing and decreasing <sup>1)</sup>	P_4.2.2
Power half-bridge output voltage	V <sub>OUT</sub>	-0.3	-	40	V	0 V < V <sub>OUT</sub> < V <sub>S</sub>	P_4.1.2
Logic supply voltage	$V_{DD}$	-0.3	_	5.5	V	0 V < V <sub>S</sub> < 40 V	P_4.1.3
Logic input voltages (SDI, SCLK, CSN, EN)	$V_{\rm SDI}, \ V_{\rm SCLK}, \ V_{\rm CSN}, V_{\rm EN}$	-0.3	_	VDD	V	0 V < V <sub>S</sub> < 40 V 0 V < V <sub>DD</sub> < 5.5V	P_4.1.4
Logic output voltage (SDO)	$V_{SDO}$	-0.3	-	VDD	V	0 V < V <sub>S</sub> < 40 V 0 V < V <sub>DD</sub> < 5.5V	P_4.1.5
Currents							
Continuous Supply Current for $V_{S1}$	I <sub>S1</sub>	0	_	1.5	Α	_	P_4.1.6
Continuous Supply Current for $V_{S2}$	I <sub>S2</sub>	0	_	1.5	Α	-	P_4.1.7
Current per GND pin	I <sub>GND</sub>	0	_	2.0	Α	-	P_4.1.14
Output Currents	I <sub>OUT</sub>	-2.0	_	2.0	Α	-	P_4.1.15
Temperatures							
Junction temperature	$T_{\rm j}$	-40	-	150	°C	-	P_4.1.8
Storage temperature	$T_{\rm stg}$	-50	-	150	°C	-	P_4.1.9
ESD Susceptibility	•	*	•	•	•		•
ESD susceptibility OUTn and VSx pins versus GND. All other pins grounded.	V <sub>ESD</sub>	-8	_	8	kV	JEDEC HBM <sup>1)2)</sup>	P_4.1.10
ESD susceptibility all pins	V <sub>ESD</sub>	-2	-	2	kV	JEDEC HBM <sup>1)2)</sup>	P_4.1.11
ESD susceptibility all pins	V <sub>ESD</sub>	-500	-	500	V	CDM <sup>1)3)</sup>	P_4.1.12
ESD susceptibility corner pins	$V_{ESD}$	-750	_	750	٧	CDM <sup>1)3)</sup>	P_4.1.13

<sup>1)</sup> Not subject to production test, specified by design

#### **Notes**

1. Stresses above the ones listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

<sup>2)</sup> ESD susceptibility, "JEDEC HBM" according to ANSI/ ESDA/ JEDEC JS001 (1.5 kΩ, 100pF)

<sup>3)</sup> ESD susceptibility, Charged Device Model "CDM" according JEDEC JESD22-C101

#### **TLE94106EL**



#### **General Product Characteristics**

2. Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.



# 3.2 Functional Range

Table 3 Functional Range

Parameter	Symbol		Values	5	Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Supply voltage range for normal operation	V <sub>S(nor)</sub>	5.5	-	18	V	-	P_4.2.1
Extended supply voltage range	$V_{S(ext)}$	18	-	20	٧	1)2)	P_4.2.7
Logic supply voltage range for normal operation	$V_{ m DD}$	3.0	-	5.5	V	-	P_4.2.3
Logic input voltages (SDI, SCLK, CSN, EN)	$V_{\rm SDI}$ , $V_{\rm SCLK}$ , $V_{\rm CSN}$ , $V_{\rm EN}$	-0.3	-	5.5	V	-	P_4.2.4
Junction temperature	$T_{\rm j}$	-40	-	150	°C		P_4.2.5

<sup>1)</sup> Not subject to production test, specified by design.

Note:

Within the normal functional range the IC operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the related electrical characteristics table.

<sup>2)</sup> In the extended supply range, the device is still functional. However, deviations of the specified electrical characteristics are possible.



#### 3.3 Thermal Resistance

Table 4 Thermal Resistance TLE94106EL

Parameter	Symbol Values				Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Junction to Case, $T_A = -40$ °C	$R_{\rm thjC\_cold}$	_	7	_	K/W	1)	
Junction to Case, T <sub>A</sub> = 85°C	R <sub>thjC_hot</sub>	_	9	_	K/W	1)	
Junction to ambient, $T_A = -40^{\circ}C$ (1s0p, minimal footprint)	R <sub>thjA_cold_</sub>	_	80	-	K/W	1) 2)	
Junction to ambient, $T_A = 85$ °C (1s0p, minimal footprint)	R <sub>thjA_hot_m</sub>	_	85	-	K/W	1) 2)	
Junction to ambient, $T_A = -40^{\circ}\text{C}$ (1s0p, 300mm2 Cu)	R <sub>thjA_cold_3</sub>	_	51	-	K/W	1) 3)	
Junction to ambient, $T_A = 85^{\circ}C$ (1s0p, 300mm2 Cu)	R <sub>thjA_hot_30</sub>	_	58	-	K/W	1) 3)	
Junction to ambient, $T_A = -40^{\circ}\text{C}$ (1s0p, 600mm2 Cu)	R <sub>thjA_cold_6</sub>	_	51	-	K/W	1) 4)	
Junction to ambient, $T_A = 85^{\circ}C$ (1s0p, 600mm2 Cu)	R <sub>thjA_hot_60</sub>	_	58	-	K/W	1) 4)	
Junction to ambient, $T_A = -40^{\circ}\text{C}$ (2s2p)	R <sub>thjA_cold_2</sub>	_	35	-	K/W	1) 5)	
Junction to ambient, $T_A = 85^{\circ}C$ (2s2p)	R <sub>thjA_hot_2s</sub>	_	44	-	K/W	1) 5)	

- 1) Not subject to production test, specified by design.
- 2) Specified  $R_{thJA}$  value is according to JEDEC JESD51-2, -3 at natural convection on FR4 1s0p board; The product (chip + package) was simulated on a 76.2 x 114.3 x 1.5mm board with minimal footprint copper area and 35 $\mu$ m thickness. Ta = -40°C, each channel dissipates 0.2W. Ta = 85°C, each channel dissipates 0.135W.
- 3) Specified  $R_{thJA}$  value is according to JEDEC JESD51-2, -3 at natural convection on FR4 1s0p board; The product (chip + package) was simulated on a 76.2 x 114.3 x 1.5mm board with additional cooling of 300mm2 copper area and 35 $\mu$ m thickness. Ta = -40°C, each channel dissipates 0.2W. Ta = 85°C, each channel dissipates 0.135W.
- 4) Specified  $R_{thJA}$  value is according to JEDEC JESD51-2, -3 at natural convection on FR4 1s0p board; The product (chip + package) was simulated on a 76.2 x 114.3 x 1.5mm board with additional cooling of 600mm2 copper area and 35 $\mu$ m thickness. Ta = -40°C, each channel dissipates 0.2W. Ta = 85°C, each channel dissipates 0.135W.
- 5) Specified  $R_{thJA}$  value is according to JEDEC JESD51-2, -3 at natural convection on FR4 2s2p board; The product (chip + package) was simulated on a 76.2 x 114.3 x 1.5mm board with two inner copper layers (4 x 35 $\mu$ m Cu). Ta = -40°C, each channel dissipates 0.2W. Ta = 85°C, each channel dissipates 0.135W.



#### **Electrical Characteristics** 3.4

Table 5 Electrical Characteristics,  $V_{\rm S}$  =5.5 V to 18 V,  $V_{\rm DD}$  = 3.0V to 5.5V,  $T_{\rm j}$  = -40°C to +150°C, EN= HIGH,  $I_{\rm OUTn}$ = 0 A; Typical values refer to  $V_{\rm DD}$  = 5.0 V,  $V_{\rm S}$  = 13.5 V and  $T_{\rm J}$  = 25 °C unless otherwise specified; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified)

Parameter	Symbol	Values			Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Current Consumption, EN = GN	ID	1	T.	1			
Supply Quiescent current	$I_{SQ}$	-	0.5	2	μΑ	-40°C ≤ <i>T</i> <sub>j</sub> ≤ 85°C	P_4.4.1
Logic supply quiescent current	$I_{\mathrm{DD}_{\mathrm{Q}}}$	_	0.1	1	μΑ	$-40^{\circ}\text{C} \le T_{j} \le 85^{\circ}\text{C}$	P_4.4.2
Total quiescent current	$I_{SQ} + I_{DD_Q}$	-	0.6	3	μΑ	$-40^{\circ}\text{C} \le T_{j} \le 85^{\circ}\text{C}$	P_4.4.3
Current Consumption, EN=HIG	Н						
Supply current	Is	_	0.5	1	mA	Power drivers and power stages are off	P_4.4.4
Supply current	I <sub>S_HSON</sub>	_	4.5	9	mA	All high-sides ON <sup>1)</sup>	P_4.4.101
Logic supply current	I <sub>DD</sub>	-	1.5	3	mA	SPI not active	P_4.4.5
Logic supply current	I <sub>DD_RUN</sub>	_	5	-	mA	SPI 5MHz 3)	P_4.4.6
Total supply current	I <sub>S</sub> +I <sub>DD_RUN</sub>	-	5.5	-	mA	SPI 5MHz 3)	P_4.4.7
Over- and Undervoltage Locko	ut						
Undervoltage Switch ON voltage threshold	V <sub>UV ON</sub>	4.25	-	5.25	V	V <sub>S</sub> increasing	P_4.4.8
Undervoltage Switch OFF voltage threshold	V <sub>UV OFF</sub>	4	-	5.0	V	$V_{\rm S}$ decreasing	P_4.4.9
Undervoltage Switch ON/OFF hysteresis	V <sub>UV HY</sub>	-	0.25	-	V	V <sub>UV ON</sub> - V <sub>UV OFF</sub> 3)	P_4.4.10
Overvoltage Switch OFF voltage threshold	V <sub>OV OFF</sub>	21	-	25	V	V <sub>S</sub> increasing	P_4.4.11
Overvoltage Switch ON voltage threshold	V <sub>OV ON</sub>	20	-	24	V	$V_{\rm S}$ decreasing	P_4.4.12
Overvoltage Switch ON/OFF hysteresis	V <sub>OV HY</sub>	-	1	-	V	V <sub>OV OFF</sub> - V <sub>OV ON</sub> 3)	P_4.4.13
V <sub>DD</sub> Power-On-Reset	$V_{\rm DDPOR}$	2.40	2.70	2.90	V	V <sub>DD</sub> increasing	P_4.4.14
V <sub>DD</sub> Power-Off-Reset	$V_{\rm DD\ POffR}$	2.35	2.65	2.85	V	V <sub>DD</sub> decreasing	P_4.4.15
V <sub>DD</sub> Power ON/OFF hysteresis	V <sub>DD POR HY</sub>	-	0.05	_	V	$V_{\rm DD\ POR}$ - $V_{\rm DD\ POffR}$ 3)	P_4.4.98
Static Drain-source ON-Resista	ance (High-S	ide or L	ow-Sid	le)			
High-Side or Low-Side R <sub>DSON</sub> (all outputs)	R <sub>DSON_HB_25C</sub>	-	850	1200	mΩ	$I_{OUT} = \pm 0.5 \text{ A};$ $T_i = 25 \text{ °C}$	P_4.4.16
High-Side or Low-Side R <sub>DSON</sub> (all outputs)	R <sub>DSON_HB_150</sub>	_	1400	1800	mΩ	$I_{OUT} = \pm 0.5 \text{ A};$ $T_i = 150 ^{\circ}\text{C}$	P_4.4.17



Table 5 Electrical Characteristics,  $V_{\rm S}$  =5.5 V to 18 V,  $V_{\rm DD}$  = 3.0V to 5.5V,  $T_{\rm j}$  = -40°C to +150°C, EN= HIGH,  $I_{\rm OUTn}$  = 0 A; Typical values refer to  $I_{\rm DD}$  = 5.0 V,  $I_{\rm S}$  = 13.5 V and  $I_{\rm J}$  = 25 °C unless otherwise specified; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified) (cont'd)

Parameter	Symbol		Values		Unit	Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
High-Side R <sub>DSON</sub> (HS1 and HS2 in LED mode)	R <sub>DSON_HI_HB_</sub> 25C	-	950	1300	mΩ	$I_{OUT} = \pm 0.1 \text{ A};$ $I_{j} = 25 \text{ °C}$	P_4.4.18
High-Side R <sub>DSON</sub> (HS1 and HS2 in LED mode)	R <sub>DSON_HI_HB_</sub>	_	1500	2000	mΩ	$I_{OUT} = \pm 0.1 \text{ A};$ $I_{j} = 150 \text{ °C}$	P_4.4.19
Output Protection and Diagno	sis of high-si	ide (HS)	chann	els of h	alf-brid	lge output	
HS Overcurrent Shutdown Threshold	I <sub>SD_HS</sub>	-1.4	-1.1	-0.9	A	See <b>Figure 7</b>	P_4.4.89
Difference between shutdown and limit current	I <sub>LIM_HS</sub> - I <sub>SD_HS</sub>	-1.2	-0.6	0	А	$ I_{LIM\_HS}  \ge  I_{SD\_HS} $ See <b>Figure 7</b>	P_4.4.21
Overcurrent Shutdown filter time	$t_{\sf dSD\_HS}$	15	19	23	μs	3)	P_4.4.22
Open Load Detection Current	I <sub>OLD1_HS</sub>	-15	-8	-3	mA	-	P_4.4.23
Open Load Detection filter time		2000	3000	4000	μs	3)	P_4.4.24
Open Load Detection Current for LED mode (HS1 & HS2)	I <sub>OLD2_HS1,2</sub>	-3.2	-2	-0.5	mA	Bit OL_SEL_HS1 = 1, OL_SEL_HS2 = 1	P_4.4.25
Open Load Detection filter time for LED mode (HS1 & HS2)	t <sub>OLD2_HS1,2</sub>	100	200	300	μs	Bit OL_SEL_HS1 = 1, OL_SEL_HS2 = 1; 3)	P_4.4.26
Output Protection and Diagno	sis of low-sid	de (LS)	channe	ls of ha	lf-bridg	ge output	
LS Overcurrent Shutdown Threshold	I <sub>SD_LS</sub>	0.9	1.1	1.4	А	Figure 8	P_4.4.104
Difference between shutdown and limit current	I <sub>LIM_LS</sub> - I <sub>SD_LS</sub>	0	0.6	1.2	A	3) I <sub>LIM_LS</sub> ≥ I <sub>SD_LS</sub> Figure 8	P_4.4.28
Overcurrent Shutdown filter time	$t_{dSD\_LS}$	15	19	23	μs	3)	P_4.4.29
Open Load Detection Current	I <sub>OLD_LS</sub>	3	8	15	mA	-	P_4.4.30
Open Load Detection filter time	$t_{OLD\_LS}$	2000	3000	4000	μs	3)	P_4.4.31
Outputs OUT(1n) leakage cu	rrent	•					
HS leakage current in off state	I <sub>QLHn_NOR</sub>	-2	-0.5	_	μΑ	V <sub>OUTn</sub> = 0V; EN=High	P_4.4.32
HS leakage current in off state	I <sub>QLHn_SLE</sub>	-2	-0.5	_	μΑ	V <sub>OUTn</sub> = 0V; EN=GND	P_4.4.33
LS Leakage current in off state	I <sub>QLLn_NOR</sub>	_	0.5	2	μΑ	$V_{\text{OUTn}} = V_{\text{S}}$ ; EN=High	P_4.4.34
LS Leakage current in off state	I <sub>QLLn_SLE</sub>	_	0.5	2	μΑ	$V_{\text{OUTn}} = V_{\text{S}}$ ; EN=GND	P_4.4.35
Output Switching Times. See F	igure 9 and	Figure :	LO.				
Slew rate of high-side and low- side outputs	$d_{VOUT}/dt$	0.1	0.45	0.75	V/µs	Resistive load = $100\Omega$ ; $V_S$ =13.5V <sup>4)</sup>	P_4.4.36



Table 5 Electrical Characteristics,  $V_{\rm S}$  = 5.5 V to 18 V,  $V_{\rm DD}$  = 3.0V to 5.5V,  $T_{\rm j}$  = -40°C to +150°C, EN= HIGH,  $I_{\rm OUTn}$  = 0 A; Typical values refer to  $V_{\rm DD}$  = 5.0 V,  $V_{\rm S}$  = 13.5 V and  $T_{\rm J}$  = 25 °C unless otherwise specified; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified) (cont'd)

Parameter	Symbol		Values			Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
Output delay time high side driver on	$t_{\sf dONH}$	5	20	35	μs	Resistive load = 100Ω to GND	P_4.4.37
Output delay time high side driver off	$t_{dOFFH}$	15	45	75	μs	Resistive load = 100Ω to GND	P_4.4.38
Output delay time low side driver on	$t_{\sf dONL}$	5	20	35	μs	Resistive load = 100Ω to VS	P_4.4.39
Output delay time low side driver off	$t_{dOFFL}$	15	45	75	μs	Resistive load = 100Ω to VS	P_4.4.40
Cross current protection time, high to low	$t_{DHL}$	100	130	160	μs	Resistive load = $100\Omega^{3)}$	P_4.4.41
Cross current protection time, low to high	$t_{DLH}$	100	130	160	μs	Resistive load = $100\Omega^{3)}$	P_4.4.42
Input Interface: Logic Input E	N				<b>"</b>		
High-input voltage	V <sub>ENH</sub>	0.7 * V <sub>DD</sub>	_	_	V	-	P_4.4.43
Low-input voltage	V <sub>ENL</sub>	-	-	0.3 * V <sub>DD</sub>	V	-	P_4.4.44
Hysteresis of input voltage	$V_{ENHY}$	_	500	-	mV	3)	P_4.4.45
Pull down resistor	R <sub>PD_EN</sub>	20	40	70	kΩ	$V_{\rm EN} = 0.2 \times V_{\rm DD}$	P_4.4.46
SPI frequency		1					
Maximum SPI frequency	$f_{\rm SPI,max}$	_	-	5.0	MHz	3) 5)	P_4.4.47
SPI INTERFACE: Delay Time fr		g edge to	first D	ata in			
Setup time	$t_{\rm set}$	_	-	150	μs	3) See <b>Figure 14</b>	P_4.4.48
SPI INTERFACE: Input Interface	e, Logic Inp	outs SDI,	SCLK,	CSN			
H-input voltage threshold	$V_{IH}$	0.7 * V <sub>DD</sub>	-	_	V	-	P_4.4.50
L-input voltage threshold	V <sub>IL</sub>	_	-	0.3 * V <sub>DD</sub>	V	-	P_4.4.51
Hysteresis of input voltage	$V_{IHY}$	_	500	-	mV	3)	P_4.4.52
Pull up resistor at pin CSN	R <sub>PU_CSN</sub>	30	50	80	kΩ	$V_{\rm CSN} = 0.7 \times V_{\rm DD}$	P_4.4.53
Pull down resistor at pin SDI, SCLK	R <sub>PD_SDI</sub> , R <sub>PD_SCLK</sub>	20	40	70	kΩ	$V_{\rm SDI}$ , $V_{\rm SCLK} = 0.2 \times V_{\rm DD}$	P_4.4.54
Input capacitance at pin CSN, SDI or SCLK	Cı	-	10	15	pF	$0V < V_{DD} < 5.25V^{3}$	P_4.4.55
Input Interface, Logic Output	SDO			•	•		
H-output voltage level	$V_{SDOH}$	V <sub>DD</sub> - 0.4	V <sub>DD</sub> - 0.2	-	V	I <sub>SDOH</sub> = -1.6 mA	P_4.4.56



Table 5 Electrical Characteristics,  $V_{\rm S}$  = 5.5 V to 18 V,  $V_{\rm DD}$  = 3.0V to 5.5V,  $T_{\rm j}$  = -40°C to +150°C, EN= HIGH,  $I_{\rm OUTn}$  = 0 A; Typical values refer to  $V_{\rm DD}$  = 5.0 V,  $V_{\rm S}$  = 13.5 V and  $T_{\rm J}$  = 25 °C unless otherwise specified; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified) (cont'd)

Parameter	Symbol		Values			Note or	Number
		Min.	Тур.	Max.		<b>Test Condition</b>	
L-output voltage level	$V_{\text{SDOL}}$	_	0.2	0.4	V	I <sub>SDOL</sub> = 1.6 mA	P_4.4.57
Tri-state Leakage Current	I <sub>SDOLK</sub>	-1	-	1	μΑ	$V_{\text{CSN}} = V_{\text{DD}};$ $0V < V_{\text{SDO}} < V_{\text{DD}}$	P_4.4.58
Tri-state input capacitance	$C_{SDO}$	_	10	15	pF	3)	P_4.4.59
Data Input Timing. See Figure	15 and Figu	ıre 17.					
SCLK Period	$t_{pCLK}$	200	-	-	ns	3)	P_4.4.60
SCLK High Time	t <sub>SCLKH</sub>	0.45 * t <sub>pCLK</sub>	_	0.55 * t <sub>pCLK</sub>	ns	3)	P_4.4.61
SCLK Low Time	t <sub>SCLKL</sub>	0.45 * t <sub>pCLK</sub>	_	0.55 * t <sub>pCLK</sub>	ns	3)	P_4.4.62
SCLK Low before CSN Low	$t_{BEF}$	125	_	_	ns	3)	P_4.4.63
CSN Setup Time	$t_{\text{lead}}$	250	_	_	ns	3)	P_4.4.64
SCLK Setup Time	$t_{\text{lag}}$	250	_	_	ns	3)	P_4.4.65
SCLK Low after CSN High	$t_{BEH}$	125	_	_	ns	3)	P_4.4.66
SDI Setup Time	t <sub>SDI_setup</sub>	30	_	-	ns	3)	P_4.4.67
SDI Hold Time	t <sub>SDI_hold</sub>	30	_	-	ns	3)	P_4.4.68
Input Signal Rise Time at pin SDI, SCLK, CSN	t <sub>rIN</sub>	-	_	50	ns	3)	P_4.4.69
Input Signal Fall Time at pin SDI, SCLK, CSN	t <sub>fIN</sub>	-	-	50	ns	3)	P_4.4.70
Delay time from EN falling edge to standby mode	$t_{DMODE}$	-	-	8	μs	3)	P_4.4.71
Minimum CSN High Time	$t_{CSNH}$	5	_	-	μs	3)	P_4.4.72
Data Output Timing. See Figur		,					
SDO Rise Time	$t_{rSDO}$	_	30	80	ns	C <sub>load</sub> = 40pF <sup>3)</sup>	P_4.4.73
SDO Fall Time	$t_{fSDO}$	_	30	80	ns	$C_{load} = 40pF^{3)}$	P_4.4.74
SDO Enable Time after CSN falling edge	t <sub>ENSDO</sub>	-	_	75	ns	Low Impedance 3)	P_4.4.75
SDO Disable Time after CSN rising edge	$t_{ extsf{DISSDO}}$	-	-	75	ns	High Impedance 3)	P_4.4.76
Duty cycle of incoming clock at SCLK	duty <sub>SCLK</sub>	45	-	55	%	3)	P_4.4.77
SDO Valid Time for $V_{DD} = 3.3V$	t <sub>VASDO3</sub>	-	70	95	ns	$V_{SDO} < 0.2 \times V_{DD}$ $V_{SDO} > 0.8 \times V_{DD}$ $C_{load} = 40 \text{pF}^{3}$	P_4.4.78



Table 5 Electrical Characteristics,  $V_S$  = 5.5 V to 18 V,  $V_{DD}$  = 3.0V to 5.5V,  $T_j$  = -40°C to +150°C, EN= HIGH,  $I_{OUTn}$  = 0 A; Typical values refer to  $V_{DD}$  = 5.0 V,  $V_S$  = 13.5 V and  $T_J$  = 25 °C unless otherwise specified; all voltages with respect to ground, positive current flowing into pin (unless otherwise specified) (cont'd)

<b>a</b> ) (conca)						
Symbol	Values			Unit	Note or	Number
	Min.	Тур.	Max.		<b>Test Condition</b>	
t <sub>VASDO5</sub>	_	50	65	ns	$V_{SDO} < 0.2 \times V_{DD}$ $V_{SDO} > 0.8 V_{DD}$ $C_{load} = 40 pF^{3}$	P_4.4.79
				•		
$T_{jW}$	120	140	170	°C	See Figure 11 <sup>3)</sup>	P_4.4.80
$T_{\rm jSD}$	150	175	200	°C	See Figure 11 <sup>3)</sup>	P_4.4.81
$T_{\rm jHYS}$	-	5	-	°C	3)	P_4.4.82
$T_{\rm jSD}/T_{\rm jW}$	1.05	1.20	-	_	3)	P_4.4.83
	$t_{ m VASDO5}$ $T_{ m jW}$ $T_{ m jSD}$ $T_{ m jHYS}$	$\begin{array}{c c} \textbf{Symbol} & & \\ \hline \textbf{Min.} \\ \hline t_{\text{VASDO5}} & - \\ \hline \\ \hline T_{\text{jW}} & 120 \\ \hline T_{\text{jSD}} & 150 \\ \hline \\ \hline T_{\text{jHYS}} & - \\ \hline \end{array}$	Symbol         Value           Min.         Typ. $t_{VASDO5}$ -         50 $T_{jW}$ 120         140 $T_{jSD}$ 150         175 $T_{jHYS}$ -         5	Symbol         Values           Min.         Typ.         Max. $t_{VASDO5}$ -         50         65 $T_{jW}$ 120         140         170 $T_{jSD}$ 150         175         200 $T_{jHYS}$ -         5         -	Symbol         Values         Unit           Min.         Typ.         Max. $t_{VASDO5}$ -         50         65         ns $T_{jW}$ 120         140         170         °C $T_{jSD}$ 150         175         200         °C $T_{jHYS}$ -         5         -         °C	Min.       Typ.       Max.       Test Condition $t_{VASDO5}$ -       50       65       ns $V_{SDO} < 0.2 \times V_{DD}$ $V_{SDO} > 0.8 V_{DD}$ $C_{load} = 40 pF^{3}$ $T_{jW}$ 120       140       170       °C       See Figure 11 (3) $T_{jSD}$ 150       175       200       °C       See Figure 11 (3) $T_{jHYS}$ -       5       -       °C       3)

<sup>1)</sup>  $I_{S \text{ HSON}}$  does not include the load current

<sup>2)</sup> HS1, respectively HS2, is set to LED mode by setting OL\_SEL\_HS1 bit to 1, respectively OL\_SEL\_HS2 bit to 1

<sup>3)</sup> Not subject to production test, specified by design

<sup>4)</sup> Measured for 20% - 80% of  $V_{\rm S}$ .

<sup>5)</sup> Not applicable in daisy chain configuration

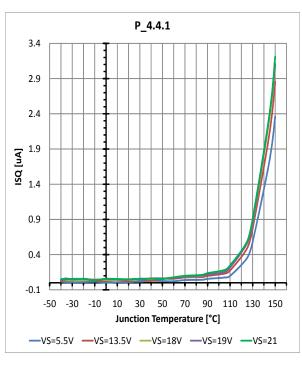


#### 4 Characterization results

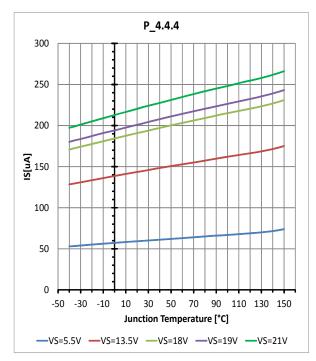
Performed on 7 devices from 2 lots, over operating temperature and nominal/extended supply range.

#### **Typical performance characteristics**

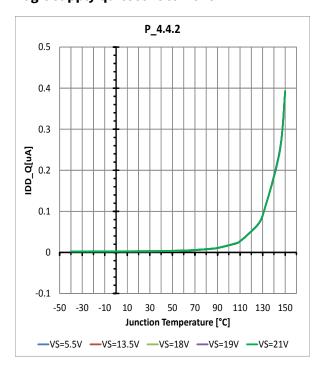
#### Supply quiescent current



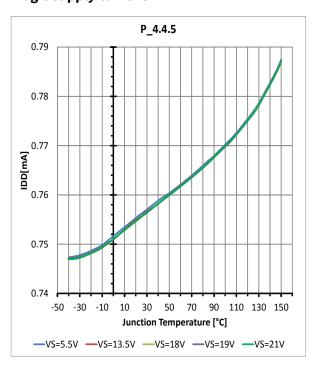
#### **Supply current**



#### Logic supply quiescent current

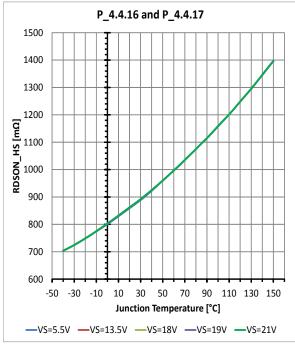


#### Logic supply current

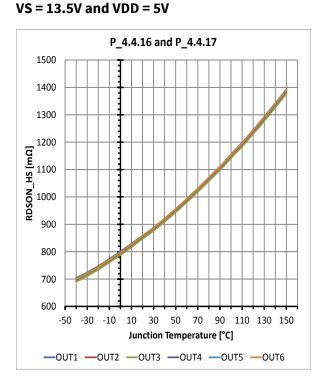




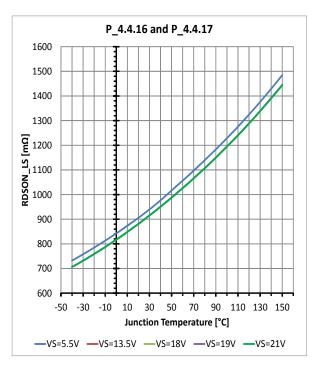
#### **HS static Drain-source ON-resistance**



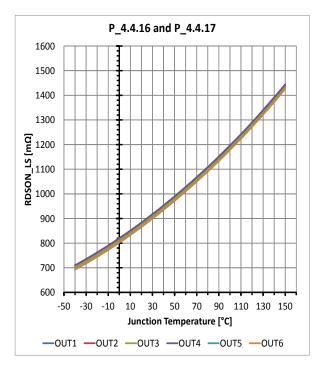
# HS static drain-source ON-resistance



#### LS static Drain-source ON-resistance



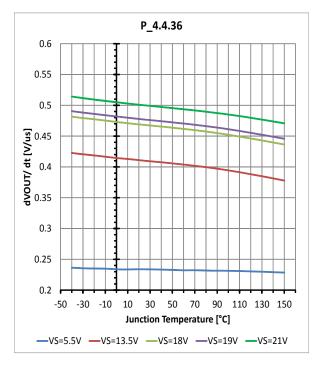
# LS static drain-source ON-resistance VS = 13.5V and VDD = 5V



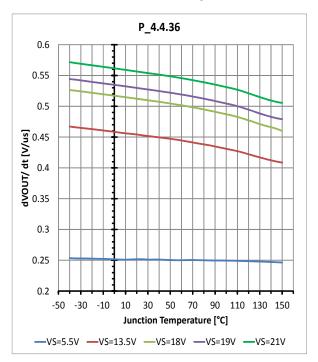
1.0



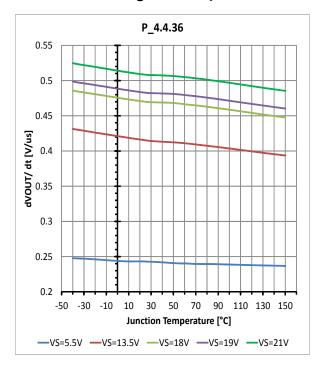
#### Slew rate ON of high-side outputs



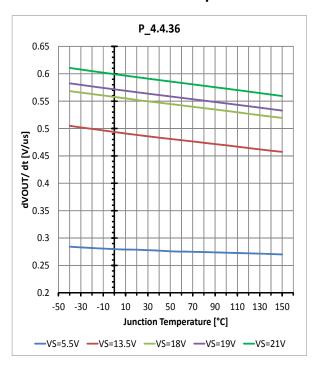
#### Slew rate ON of low-side outputs



#### Slew rate OFF of high-side outputs

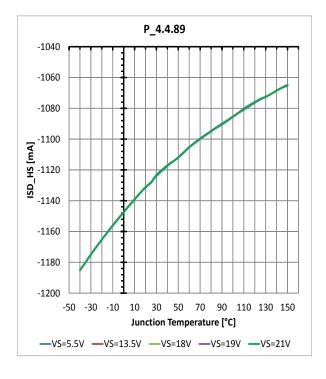


#### Slew rate OFF of low-side outputs

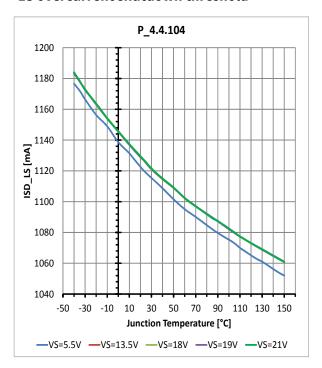




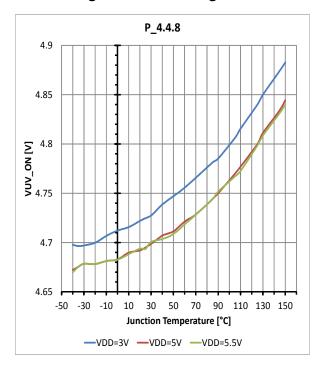
#### **HS** overcurrent shutdown threshold



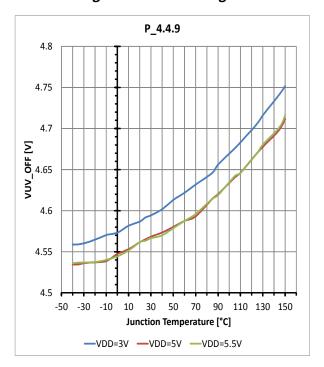
#### LS overcurrent shutdown threshold



#### **Undervoltage switch ON voltage threshold**

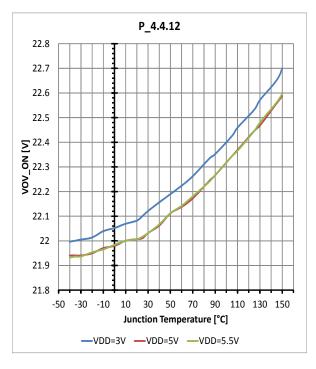


# Undervoltage switch OFF voltage threshold

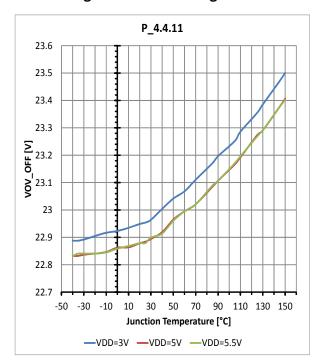




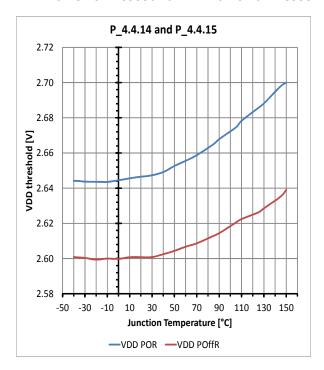
#### Overvoltage switch ON voltage threshold



#### Overvoltage switch OFF voltage threshold



#### **VDD Power-on-reset and VDD Power-off-reset**



# infineon

#### **General Description**

# **5** General Description

#### 5.1 Power Supply

The TLE94106EL has two power supply inputs,  $V_{\rm S}$  and  $V_{\rm DD}$ . The half bridge outputs are supplied by  $V_{\rm S}$ , which is connected to the 12V automotive supply rail.  $V_{\rm DD}$  is used to supply the I/O buffers and internal voltage regulator of the device.

 $V_{\rm S}$  and  $V_{\rm DD}$  supplies are separated so that information stored in the logic block remains intact in the event of voltage drop outs or disturbances on  $V_{\rm S}$ . The system can therefore continue to operate once  $V_{\rm S}$  has recovered, without having to resend commands to the device.

A rising edge on  $V_{\rm DD}$  crossing  $V_{\rm DD\,POR}$  triggers an internal Power-On Reset (POR) to initialize the IC at power-on. All data stored internally is deleted, and the outputs are switched off (high impedance).

An electrolytic and 100nF ceramic capacitors are recommended to be placed as close as possible to the  $V_S$  supply pin of the device for improved EMC performance in the high and low frequency band. The electrolytic capacitor must be dimensioned to prevent the VS voltage from exceeding the absolute maximum rating. In addition, decoupling capacitors are recommended on the  $V_{DD}$  supply pin.

#### 5.2 Operation modes

#### 5.2.1 Normal mode

The TLE94106EL enters normal mode by setting the EN input High. In normal mode, the charge pump is active and all output transistors can be configured via SPI.

#### 5.2.2 Sleep mode

The TLE94106EL enters sleep mode by setting the EN input Low. The EN input has an internal pull-down resistor.

In sleep mode, all output transistors are turned off and the SPI register banks are reset. The current consumption is reduced to  $I_{SQ} + I_{DD\_Q}$ .

#### 5.3 Reset Behaviour

The following reset triggers have been implemented in the TLE94106EL:

#### $V_{DD}$ Undervoltage Reset:

The SPI Interface shall not function if  $V_{\rm DD}$  is below the undervoltage threshold,  $V_{\rm DD\ POffR}$ . The digital block will be deactivated, the logic contents cleared and the output stages are switched off. The digital block is initialized once  $V_{\rm DD}$  voltage levels is above the undervoltage threshold,  $V_{\rm DD\ POR}$ . Then the NPOR bit is reset (NPOR = 0 in **SYS\_DIAG1** and Global Status Register).

#### Reset on EN pin:

If the EN pin is pulled Low, the logic content is reset and the device enters sleep mode.

The reset event is reported by the NPOR bit (NPOR = 0) once the TLE94106EL is in normal mode (EN = High; VDD >  $V_{DDPOR}$ ).



#### **General Description**

# 5.4 Reverse Polarity Protection

The TLE94106EL requires an external reverse polarity protection. During reverse polarity, the free-wheeling diodes across the half bridge output will begin to conduct, causing an undesired current flow ( $I_{RB}$ ) from ground potential to battery and excessive power dissipation across the diodes. As such, a reverse polarity protection diode is recommended (see **Figure 4**).

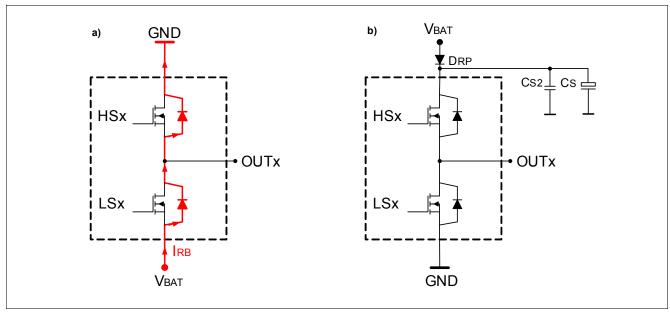


Figure 4 Reverse Polarity Protection



# 6 Half-Bridge Outputs

#### 6.1 Functional Description

The half-bridge outputs of the TLE94106EL are intended to drive motor loads. These outputs can either be driven continuously or PWM enabled via SPI.

If the outputs are driven continuously via SPI, for example HS1 and LS2 used to drive a motor, then the following suggested SPI commands shall be sent:

- Activate HS1: Bit HB1\_HS\_EN in HB\_ACT\_1\_CTRL register
- Activate LS2: Bit HB2\_LS\_EN in HB\_ACT\_1\_CTRL register

#### 6.1.1 Half-bridge operation with PWM enabled

All half-bridge outputs of the TLE94106EL are capable of PWM operation. They can either be used to drive an inductive load (e.g. DC brush motor) or optionally a resistive load (e.g. LED). Each half-bridge output has been allocated a maximum of three PWM channels with individual duty cycle settings with 8-bit resolution. Each channel is further mapped to a maximum of three PWM frequency options, i.e. 80Hz,100Hz and 200Hz. This feature enables a highly flexible PWM operation while driving loads with varying control profiles.

PWM frequency and duty cycle can be changed on demand during PWM operation of the desired half-bridge output. Glitches on the PWM output waveform, which may arise as a result of on-demand changes in PWM operation, will be prevented by the internal logic circuitry.

When operating with motor loads, active or passive free-wheeling configuration is available via SPI to select the speed at which the inductive current can decay over the full-bridge circuit. The default setting is passive free-wheeling.

Note:

Active free-wheeling is effectively applied if the selected duty cycle corresponds to turn-on times of the HS and the LS, which are longer than the sum of the cross conduction times tDHL + tDLH.

Table 6 PWM capability and frequency selection per half-bridge output

Control Register: HBx_MODEn (n=0,1)	PWM Frequency 80Hz (Control Register: PWM_CH_FREQ_CTRL)	PWM Frequency 100Hz (Control Register: PWM_CH_FREQ_CTRL)	PWM Frequency 200Hz (Control Register: PWM_CH_FREQ_CTRL)
PWM Channel 1	PWM_CH1_FREQ_n (n=0,1)	PWM_CH1_FREQ_n (n=0,1)	PWM_CH1_FREQ_n (n=0,1)
	Bit '01 <sub>B</sub> '	Bit '10 <sub>B</sub> '	Bit '11 <sub>B</sub> '
PWM Channel 2	PWM_CH2_FREQ_n (n=0,1)	PWM_CH2_FREQ_n (n=0,1)	PWM_CH2_FREQ_n (n=0,1)
	Bit '01 <sub>B</sub> '	Bit '10 <sub>B</sub> '	Bit '11 <sub>B</sub> '
PWM Channel 3	PWM_CH3_FREQ_n (n=0,1)	PWM_CH3_FREQ_n (n=0,1)	PWM_CH3_FREQ_n (n=0,1)
	Bit '01 <sub>B</sub> '	Bit '10 <sub>B</sub> '	Bit '11 <sub>B</sub> '



#### 6.1.1.1 Inductive Load

An illustration is shown in **Figure 5** with OUT1 and OUT2 driving a DC brush motor. With this configuration, HS1 is permanently driven while LS2 is driven in PWM operation. HS2 serves to actively free-wheel (FW) the motor current load, reducing the power dissipation of the device.

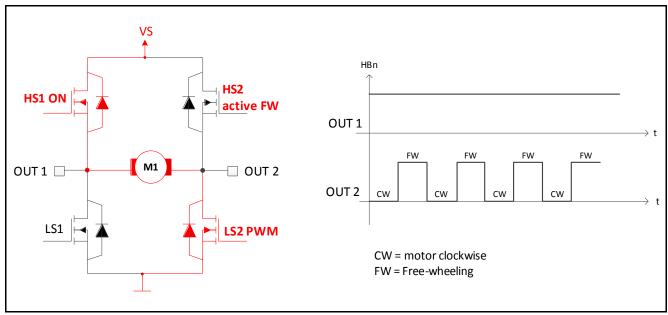


Figure 5 PWM operation on OUT 2

Assuming HBx Mode = 00 and both HSx and LSx are considered off (tri-state). The suggested SPI control commands for proper PWM operation are:

#### Option 1: The considered output is not put in parallel with another one

- Configure the frequency to 00 (PWM is stopped and off) for selected PWM channel
- Configure active or passive free-wheeling of the inductive decay current in FW\_CTRL register
- Assign an appropriate PWM channel for selected half-bridge output in HB\_MODE\_CTRL register
- Configure the duty cycle of the selected half-bridge output in PWM\_DC\_CTRL register
- Select the PWM frequency in PWM\_CH\_FREQ\_CTRL register to begin the PWM period
- Activate the channel to be driven in PWM operation: HSn or LSn in the HB\_ACT\_CTRL register

# Option 2: Outputs controlled by different control registers are put paralleled. This sequence ensures that corresponding HS or LS are activated simultaneously

- Configure the frequency 00 (PWM is stopped and off) for selected PWM channel
- Configure active or passive free-wheeling of the inductive decay current in FW\_CTRL register
- · Assign an appropriate PWM channel for selected half-bridge output in HB\_MODE\_CTRL register
- Configure the duty cycle of the selected half-bridge output in PWM\_DC\_CTRL register
- Activate the channel to be driven in PWM operation: HSn or LSn in the HB\_ACT\_CTRL register
- Select the PWM frequency in PWM\_CH\_FREQ\_CTRL register to begin the PWM period

Careful attention should be paid to the free-wheeling configuration of the half-bridge required to be driven in PWM operation. For example, in the event a high-side channel is activated and assigned a PWM channel, and active free-wheeling is selected, but a frequency mode of '00' (PWM is stopped and off) is configured in the

# **TLE94106EL**



#### **Half-Bridge Outputs**

PWM\_CH\_FREQ\_CTRL register, then the respective high-side channel will be configured low and the adjacent low-side channel within the half-bridge will be enabled. This is a result of enabling active free-wheeling.



#### 6.1.1.2 LED mode (optional)

Outputs, OUT1 and OUT2, are designed to optionally drive low current loads such as LEDs. The high-side channels, HS1 and HS2 are equipped with a lower open load threshold detection current and shorter filter time, specifically for low current loads such as LEDs. See OL\_SEL\_HS1 and OL\_SEL\_HS2 bits in FW\_OL\_CTRL register. Setting HS1 or HS2 in LED mode increases the R<sub>DSON</sub> and decreases the open load detection threshold. An illustration is shown in **Figure 6** with OUT1 driving an LED. With this configuration, HS1 is driven in PWM operation while LS1 is deactivated.

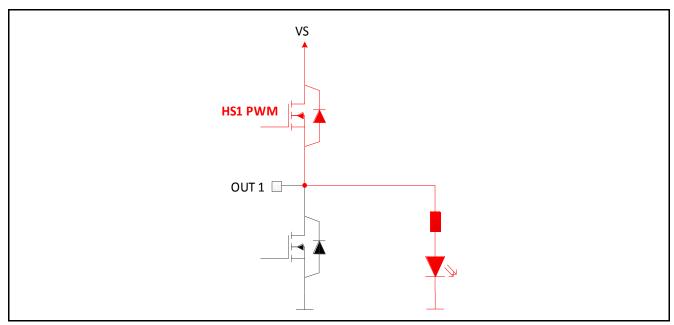


Figure 6 PWM operation on OUT 1

Assuming HBx Mode = 00 and both HSx and LSx are considered off (tri-state). The suggested SPI control commands are:

- Configure frequency 00 (PWM is stopped and off) for selected channel to ensure PWM is off.
- Assign an appropriate PWM channel for selected HS1 or HS2 output in HB\_MODE\_CTRL register
- Configure duty cycle of selected HS1 or HS2 output in PWM\_DC\_CTRL register
- Activate channel to be driven in PWM operation: HS1 or HS2 in the HB\_ACT\_CTRL register
- Select low current open load detection threshold for HS1 or HS2 in FW\_OL\_CTRL register
- Select PWM frequency in PWM\_CH\_FREQ\_CTRL register to begin the PWM period.



# 6.2 Protection & Diagnosis

The TLE94106EL is equipped with an SPI interface to control and diagnose the state of the half-bridge drivers.

This device has embedded protective functions which are designed to prevent IC destruction under fault conditions described in the following sections. Fault conditions are treated as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

The following table provides a summary of fault conditions, protection mechanisms and recovery states embedded in the TLE94106EL device.

Table 7 Summary of diagnosis and monitoring of outputs

Fault condition	Error Flag (EF) behaviour	Error bit: Status Register	Output Protection mechanism	Output error state	Output and error flag (EF) recovery
Overcurrent	Latch	1. Load Error bit, LE (bit 6) in SYS_DIAG 1: Global Status 1 Register 2. Localized error for each HS and LS channel of half-bridge, HBn_HS_OC and HBn_LS_OC bits in SYS_DIAG_2, SYS_DIAG_3 status registers.	Error output shutdown and latched	High-Z	Half-bridge control bits remain set despite error, however the output stage is shutdown. Clear EF to reactivate output stage.
Open load	Latch	1. Load Error bit, LE (bit 6) in SYS_DIAG 1: Global Status 1 Register 2. Localized error for each HS and LS channel of half-bridge, HBn_HS_OL and HBn_LS_OL bits in SYS_DIAG_5, SYS_DIAG_6 status registers.	None	No state change	An open load detection does not change the state of the output. EF to be cleared.
Temperature pre-warning	Latch	Global error bit 1, TPW in SYS_DIAG_1: Global Status 1 register	None	No state change	Not applicable
Temperature shutdown	Latch	Global error bit 2, TSD in SYS_DIAG_1: Global Status 1 register	All outputs shutdown and latched.	High-Z	Half-bridge control bits remain set despite error, however the output stage is shutdown. Clear EF to reactivate output stage.

# **TLE94106EL**



# **Half-Bridge Outputs**

**Table 7** Summary of diagnosis and monitoring of outputs (cont'd)

Fault condition	Error Flag (EF) behaviour	Error bit: Status Register	Output Protection mechanism	Output error state	Output and error flag (EF) recovery
Power supply failure due to undervoltage	Latch	Global error bit 5, VS_UV in SYS_DIAG_1: Global Status 1 register	All outputs shutdown and automatically recovers.	High-Z	Half-bridge control bits remain set despite error, however the output stage is shutdown. They will automatically be reactivated once the power supply recovers. EF to be cleared.
Power supply failure due to overvoltage	Latch	Global error bit 4, VS_OV in SYS_DIAG_1: Global Status 1 register	All outputs shutdown and automatically recover.	High-Z	Half-bridge control bits remain set despite error, however the output stage is shutdown. They will automatically be reactivated once the power supply recovers. EF to be cleared.



#### 6.2.1 Short Circuit of Output to Supply or Ground

The high-side switches are protected against short to ground whereas the low-side switches are protected against short to supply.

The high-side and low-side power switches will enter into an over-current condition if the current within the switch exceeds the overcurrent shutdown detection threshold,  $I_{\rm SD}$ . Upon detection of the  $I_{\rm SD}$  threshold, an overcurrent shutdown filter,  $t_{\rm dSD}$  is begun. As the current rises beyond the threshold  $I_{\rm SD}$ , it will be limited by the current limit threshold,  $I_{\rm LIM}$ . Upon expiry of the overcurrent shutdown filter time, the affected power switch is latched off and the corresponding error bit, HBn\_HS\_OC or HBn\_LS\_OC is set and latched. See **Figure 7** and **Figure 8** for more detail. A global load error bit, LE, contained in the global status register, SYS\_DIAG\_1, is also set for ease of error scanning by the application software. The power switch remains deactivated as long as the error bit is set.

To resume normal functionality of the power switch (in the event the overcurrent condition disappears or to verify if the failure still exists) the microcontroller shall clear the error bit in the respective status register to reactivate the desired power switch.

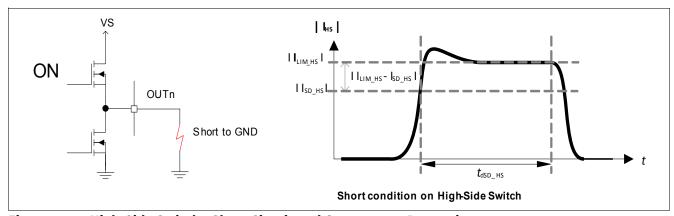


Figure 7 High-Side Switch - Short Circuit and Overcurrent Protection

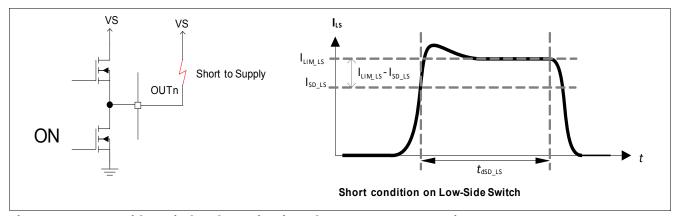


Figure 8 Low-Side Switch - Short Circuit and Overcurrent Protection

# **TLE94106EL**



#### **Half-Bridge Outputs**

Table 8 Control and Status register bit state in the event of an overcurrent condition for an activated power switch

REGISTER TYPE	REGISTER NAME	Bit	BEFORE OVERCURRENT	DURING OVERCURRENT	AFTER OVERCURRENT	
			Bit State	Bit State	Bit State	
Control	HB_ACT_CTRL_n	HBn_HS_EN HBn_LS_EN	1	1	1 (corresponding half-bridge deactivated)	
Status	SYS_DIAG_1: Global Status 1	LE	0	0	1	
Status	SYS_DIAG_x where x=2,3	HBn_HS_OC HBn_LS_OC	0	0	1	



#### 6.2.2 Cross-Current

In bridge configurations the high-side and low-side power transistors are ensured never to be simultaneously "ON" to avoid cross currents. This is achieved by integrating delays in the driver stage of the power outputs to create a dead-time between switching off of one power transistor and switching on of the adjacent power transistor within the half-bridge. The dead times,  $t_{\rm DHL}$  and  $t_{\rm DLH}$ , as shown in **Figure 9** case 3 and **Figure 10** case 3, have been specified to ensure that the switching slopes do not overlap with each other. This prevents a cross conduction event.

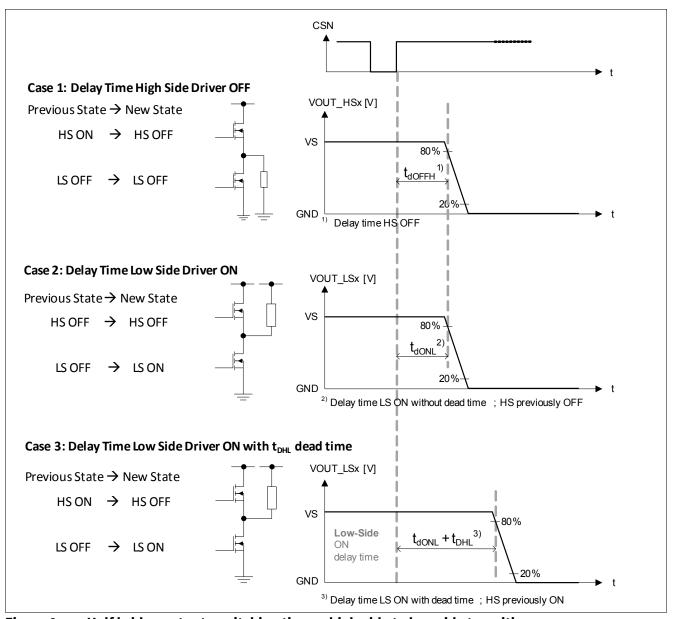


Figure 9 Half bridge outputs switching times - high-side to low-side transition



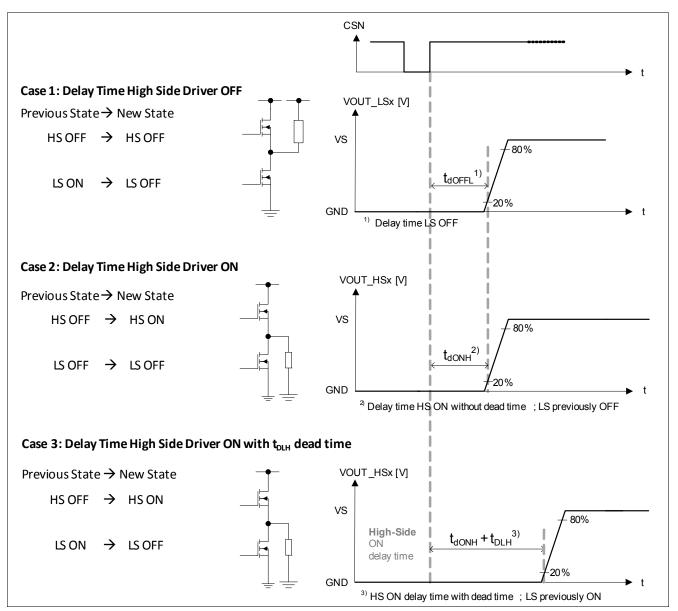


Figure 10 Half bridge outputs switching times-low-side to high-side transition



#### 6.2.3 Temperature Monitoring

Temperature sensors are integrated in the power stages. The temperature monitoring circuit compares the measured temperature to the warning and shutdown thresholds. If one or more temperature sensors reach the warning temperature, the temperature pre-warning bit, TPW is set. This bit is latched and can only be cleared via SPI. The outputs stages however remain activated.

If one or more temperature sensors reach the shut-down temperature threshold, **all outputs are latched off**. The TSD bit in SYS\_DIAG\_1: Global Status 1 is set. All outputs remain deactivated until the TSD bit is cleared. See **Figure 11**.

To resume normal functionality of the power switch (in the event the overtemperature condition disappears, or to verify if the failure still exists) the microcontroller shall clear the TSD error bit in the status register to reactivate the respective power switch.

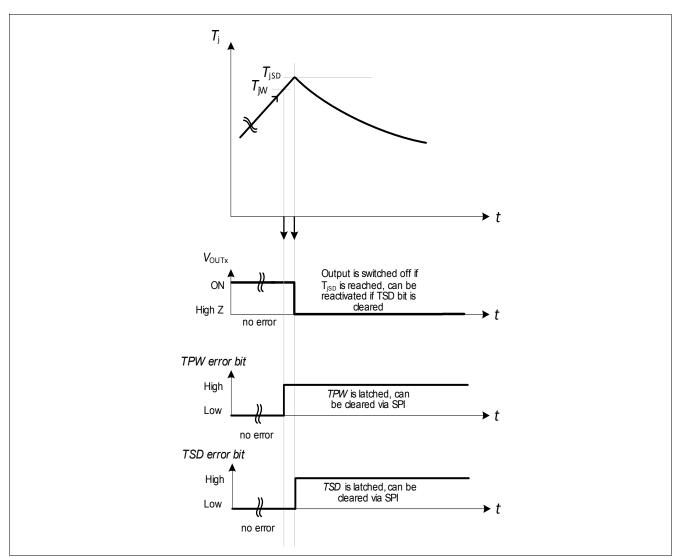


Figure 11 Overtemperature Behavior



Table 9 Control and Status register bit state in the event of an overtemperature condition for an activated power switch

REGISTER TYPE	REGISTER NAME	Bit	$T_{\rm j} < T_{\rm jw}$ Bit State	$T_j > T_{jW}$ Bit State	$T_{\rm j} > T_{\rm jSD}$ Bit State	$T_{j} < T_{jSD} - T_{jHYS}$ Bit State
Control	HB_ACT_CTRL_n	HBn_HS_EN HBn_LS_EN	1	1	1 (all outputs are latched off)	'1' (outputs are latched off unless error is cleared)
Status	SYS_DIAG_1: Global status 1	TPW	0	1 (latched)	1 (latched)	'0' if error is cleared and $T_i < T_{jW}$ , else '1'
Status	SYS_DIAG_1: Global status 1	TSD	0	0	1 (latched)	'0' if error is cleared, else '1'

#### 6.2.4 Overvoltage and undervoltage shutdown

The power supply rails  $V_S$  and  $V_{DD}$  are monitored for supply fluctuations. The  $V_S$  supply is monitored for underand over-voltage conditions where as the  $V_{DD}$  supply is monitored for under-voltage conditions.

#### **6.2.4.1** $V_s$ Undervoltage

In the event the supply voltage  $V_S$  drops below the switch off voltage  $V_{UV\,OFF}$ , all output stages are switched off, however, the logic information remains intact and uncorrupted. The  $V_S$  under-voltage error bit, VS\_UV, located in SYS\_DIAG\_1: Global Status 1 status register, will be set and latched. If  $V_S$  rises again and reaches the switch on voltage  $V_{UV\,ON}$  threshold, the power stages will automatically be activated. The VS\_UV error bit should be cleared to verify if the supply disruption is still present. See **Figure 12**.

#### 6.2.4.2 $V_s$ Overvoltage

In the event the supply voltage  $V_S$  rises above the switch off voltage  $V_{OV\,OFF}$ , all output stages are switched off. The  $V_S$  over-voltage error bit, VS\_OV, located in SYS\_DIAG\_1: Global Status 1 status register, will be set and latched. If  $V_S$  falls again and reaches the switch on voltage  $V_{OV\,ON}$  threshold, the power stages will automatically be activated. The VS\_OV error bit should be cleared to verify if the overvoltage condition is still present. See **Figure 12**.

#### 6.2.4.3 $V_{\rm DD}$ Undervoltage

In the event the VDD logic supply decreases below the undervoltage threshold,  $V_{\rm DD\ POffR}$ , the SPI interface shall no longer be functional and the TLE94106EL will enter reset.

The digital block will be initialized and the output stages are switched off to High impedance. The undervoltage reset is released once  $V_{\rm DD}$  voltage levels are above the undervoltage threshold,  $V_{\rm DD\,POR}$ .

The reset event is reported in SYS\_DIAG1 by the NPOR bit (NPOR = 0) once the TLE94106EL is in normal mode (EN = High; VDD >  $V_{DD\ POR}$ ).



#### **Half-Bridge Outputs**

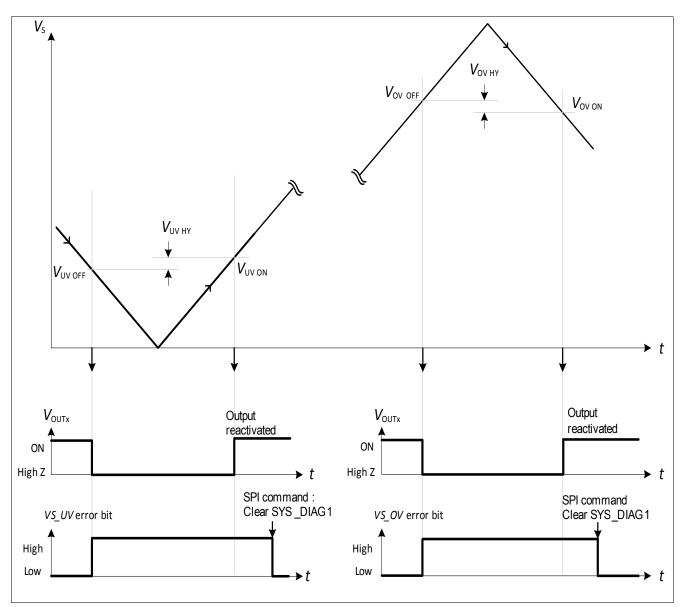


Figure 12 Output behavior during under- and overvoltage V<sub>s</sub> condition

#### 6.2.5 Open Load

Both high-side and low-side switches of the half-bridge power outputs are capable of detecting an open load in their activated state. If a load current lower than the open load detection threshold,  $I_{\rm OLD}$  for at least  $t_{\rm dOLD}$  is detected at the activated switch, the corresponding error bit, HBn\_HS\_OL or HBn\_LS\_OL is set and latched. A global load error bit, LE, in the global status register, SYS\_DIAG\_1: Global Status 1, is also set for ease of error scanning by the application software. The half-bridge output however, remains activated.

The microcontroller must clear the error bit in the respective status register to determine if the open load is still present or disappeared.

High-side outputs, HS1 and HS2, are specifically designed to detect open load thresholds for LED loads. Both HS1 and HS2 have a unique and lower open load current threshold and filter time which are configurable via SPI in control register, FW\_OL\_CTRL.

During PWM operation, the open load detection is blanked and will not be visible in the status register for power stages used in active free-wheeling



# **7** Serial Peripheral Interface (SPI)

The TLE94106EL has a 16-bit SPI interface for output control and diagnostics. This section describes the SPI protocol, the control and status registers.

#### 7.1 SPI Description

The 16-bit wide Control Input Word is read via the data input SDI, which is synchronized with the clock input SCLK provided by the microcontroller. SCLK must be Low during CSN falling edge (Clock Polarity = 0). The SPI incorporates an in-frame response: the content of the addressed register is shifted out at SDO within the same SPI frame (see **Figure 19** and **Figure 21**). The transmission cycle begins when the chip is selected by the input CSN (Chip Select Not), Low active. After the CSN input returns from Low to High, the word that has been read is interpreted according to the content. The SDO output switches to tri-state status (High impedance) at this point, thereby releasing the SDO bus for other use. The state of SDI is shifted into the input register with every falling edge on SCLK. The state of SDO is shifted out of the output register at every rising edge on SCLK (Clock Phase = 1). The SPI protocol of the TLE94106EL is compatible with independent slave configuration and with daisy chain. Daisy chaining is applicable to SPI devices with the same protocol.

Writing, clearing and reading is done byte wise. The SPI configuration and status bits are not cleared automatically by the device and therefore must be cleared by the microcontroller, e.g. if the TSD bit was set due to over temperature (refer to the respective register description for detailed information).

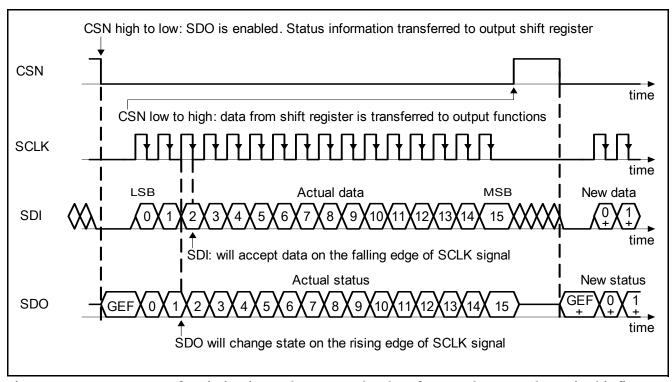


Figure 13 SPI Data Transfer Timing (note the reversed order of LSB and MSB as shown in this figure compared to the register description)

SPI messages are only recognized if a minimum set time, tSET, is observed upon rising edge of the EN pin (**Figure 14**).



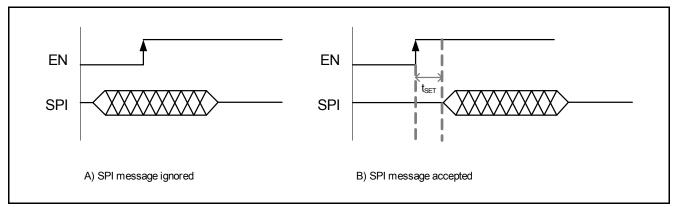


Figure 14 Setup time from EN rising edge to first SPI communication

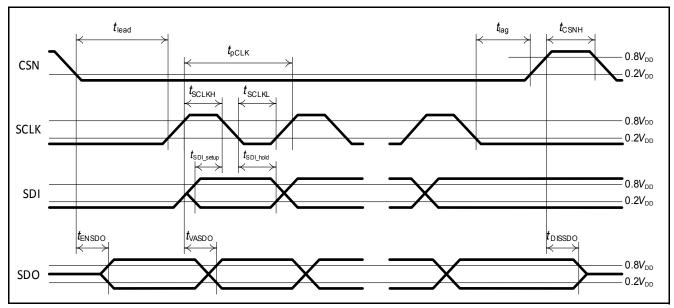


Figure 15 SPI Data Timing

#### 7.1.1 Global Error Flag

A logic OR combination between Global Error Flag (GEF) and the signal present on SDI is reported on SDO between a CSN falling edge and the first SCLK rising edge (**Figure 13**). GEF is set if a fault condition is detected or if the device comes from a Power On Reset (POR).

Note: The SDI pin of all devices in daisy chain or non daisy chain mode must be Low at the beginning of the SPI frame (between the CSN falling edge and the first SCLK rising edge).

It is possible to check if the TLE94106EL has detected a fault by reading the GEF without SPI clock pulse (**Figure 16**).



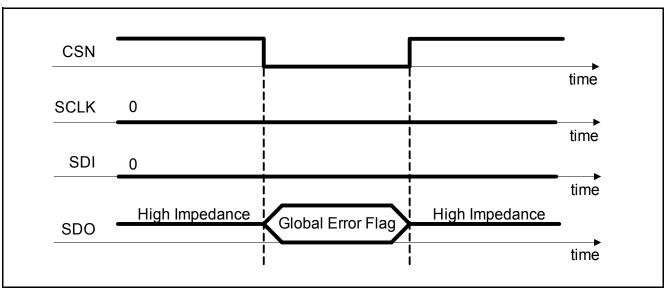


Figure 16 SDO behaviour with 0-clock cycle

#### 7.1.2 Global Status Register

The SDO shifts out during the first eight SCLK cycles the Global Status Register. This register provides an overview of the device status. All failures conditions are reported in this byte:

- SPI protocol error (SPI\_ERR)
- Load Error (LE bit): logical OR between Open Load (OL) and Overcurrent (OC) failures
- VS Undervoltage (VS\_UV bit)
- VS Overvoltage (VS\_OV bit)
- Negated Power ON Reset (NPOR bit)
- Temperature Shutdown (TSD bit)
- · Temperature Pre-Warning (TPW bit)

See Chapter 7.7.1 for details.

Note:

The Global Error Flag is a logic OR combination of every bit of the Global Status Register with the exception of NPOR:  $GEF = (SPI\_ERR) OR (LE) OR (VS\_UV) OR (VS\_OV) OR (NOT(NPOR)) OR (TSD) OR (TPW).$ 

The following table shows how failures are reported in the Global Status Register and by the Global Error Flag.

Table 10 Failure reported in the Global Status Register and Global Error Flag

Type of Error	Failure reported in the Global Status Register	Global Error Flag
SPI protocol error	SPI_ERR = 1	1
Open load or Overcurrent	LE = 1	1
VS Undervoltage	VS_UV = 1	1
VS Overvoltage	VS_OV = 1	1
Power ON Reset	NPOR = 0	1
Thermal Shutdown	TSD = 1	1



Table 10 Failure reported in the Global Status Register and Global Error Flag

Type of Error	Failure reported in the Global Status Register	Global Error Flag
Thermal Warning	TPW = 1	1
No Error and no Power ON Reset	SPI_ERR = 0 LE = 0 VS_UV = 0 VS_OV = 0 NPOR = 1 TSD = 0 TPW = 0	0

Note: The default value (after Power ON Reset) of NPOR is 0, therefore the default value of GEF is 1.

#### 7.1.3 SPI protocol error detection

The SPI incorporates an error flag in the Global Status Register (SPI\_ERR, Bit7) to supervise and preserve the data integrity. If an SPI protocol error is detected during a given frame, the SPI\_ERR bit is set in the next SPI communication.

The SPI\_ERR bit is set in the following error conditions:

- the number of SCLK clock pulses received when CSN is Low is not 0, or is not a multiple of 8 and at least 16
- the microcontroller sends an SPI command to an unused address. In particular, SDI stuck to High is reported in the SPI\_ERR bit
- the LSB of an address byte is not set to 1. In particular, SDI stuck to Low is reported in the SPI\_ERR bit
- the Last Address Bit Token (LABT, bit 1 of the address byte, see Chapter 7.2) in independent slave configuration is not set to 1
- the LABT bit of the last address byte in daisy chain configuration is not set to 1 (see Chapter 7.3)
- a clock polarity error is detected (see Figure 17 Case 2 and Case 3): the incoming clock signal was High
  during CSN rising or falling edges.

For a correct SPI communication:

- SCLK must be Low for a minimum t<sub>BFF</sub> before CSN falling edge and t<sub>lead</sub> after CSN falling edge
- SCLK must be Low for a minimum  $t_{lag}$  before CSN rising edge and  $t_{BEH}$  after CSN rising edge



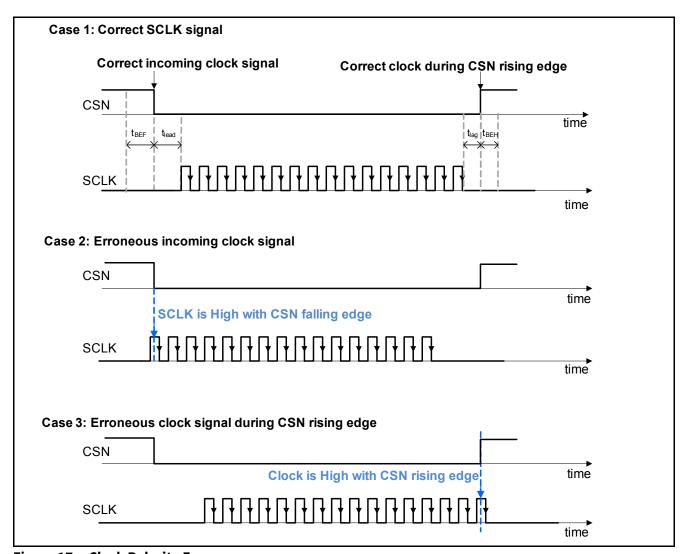


Figure 17 Clock Polarity Error



### 7.2 SPI with independent slave configuration

In an independent slave configuration, the microcontroller controls the CSN of each slave individually (Figure 18).

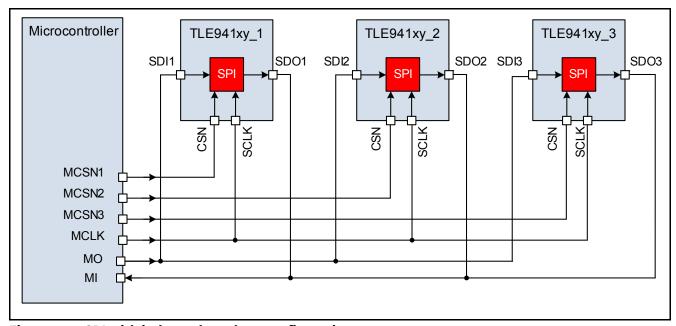


Figure 18 SPI with independent slave configuration

Each SPI communication starts with one address byte followed by one data byte (**Figure 19**). The LSB of the data byte must be set to '1'. The address bytes specifies:

- the type of operation: READ ONLY (OP bit =0) or READ/ WRITE (OP bit = 1) of the configuration bits, and READ ONLY (OP bit =0)or READ & CLEAR (OP bit = 1) of the status bits.
- The target register address (A[6:2])

The Last Address Byte Token bit (LABT, Bit1 of the address byte) must be set to 1, as no daisy chain configuration is used.

While the microcontroller sends the address byte on SDI, SDO shifts out GEF and the Global Status Register.

A further data byte (Bit15...8) is allocated to either configure the half-bridges or retrieve status information of the TLE94106EL.



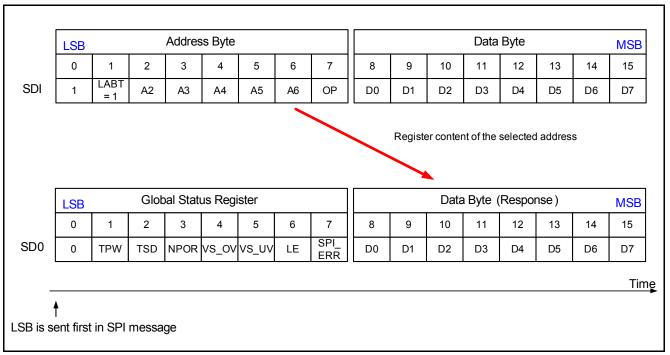


Figure 19 SPI Operation Mode with independent slave configuration

The in-frame response characteristic enables the microcontroller to read the contents of the addressed register within the SPI command. See **Figure 19**.



#### 7.3 Daisy chain operation

The TLE94106EL supports daisy chain operation with devices with the same SPI protocol. This section describes the daisy chain hardware configuration with three devices from the TLE941xy family (See Figure 20).

The master output (noted MO) is connected to a slave SDI and the first slave SDO is connected to the next slave SDI to form a chain. The SDO of the final slave in the chain will be connected to the master input (MI) to close the loop of the SPI communication frame. In daisy chain configuration, a single chip select, CSN, and clock signal, SCLK, connected in parallel to each slave device, are used by the microcontroller to control or access the SPI devices.

In this configuration, the Master Output must send the address bytes and data bytes in the following order:

- All address bytes must be sent first:
  - Address Byte 1 (for TLE941xy\_1) is sent first, followed by Address Byte 2 (for TLE941xy\_2) etc,...
  - The LABT bit of the last address byte must be 1, while the LABT bit of all the other address bytes must be 0
- The data bytes are sent all together once all address bytes have been transmitted: Data Byte 1 (for TLE941xy\_1) is sent first, followed by Data Byte 2 (for TLE941xy\_2) etc,...

Note:

The signal on the SDI pin of the first IC in daisy chain (and in non-daisy chain mode), must be Low at the beginning of the SPI frame (between CSN falling edge and the first SCLK rising edge). This is because each Global Error Flaq in daisy chain operation is implemented in OR logic.

The Master Input (MI), which is connected to the SDO of the last device in the daisy chain receives:

- A logic OR combination of all Global Error Flags (GEF), at the beginning of the SPI frame, between CSN falling edge and the first SCLK rising edge
- The logic OR combination of the GEFs is followed by the Global Status Registers in reverse order. In other words MI receives first the Global Status Register of the last device of the daisy chain
- Once all Global Status Registers are received, MI receives the response bytes corresponding to the
  respective address and data bytes in reverse order. For example, if the daisy chain consists of three devices
  with SDO or TLE941xy\_3 connected to MI, the master receives first the Response Byte 3 of TLE941xy\_3
  (corresponding to Address Byte 3 and Data Byte 3) followed by the Response Byte 2 of TLE941xy\_2 and
  finally the Response Byte 1 of TLE941xy\_1.

An example of an SPI frame with three devices from the TLE941xy family is shown in Figure 21.



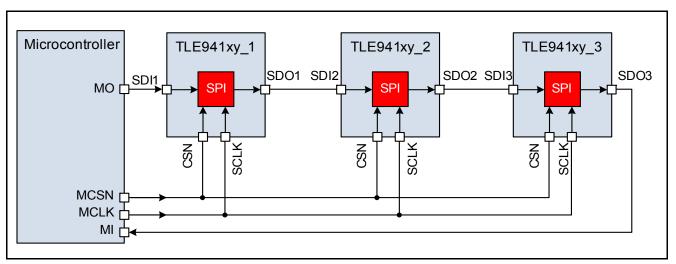


Figure 20 Example of daisy chain hardware configuration with devices from the TLE941xy family

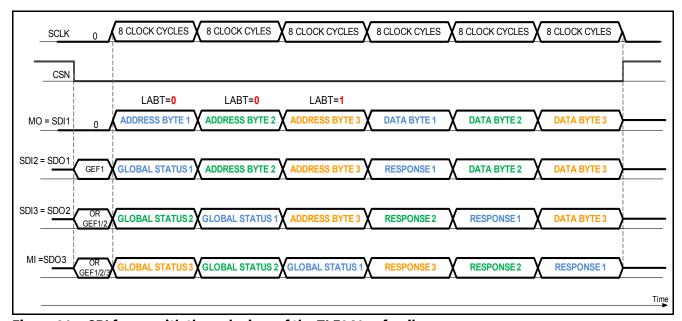


Figure 21 SPI frame with three devices of the TLE941xy family

Like in the individual slave configuration, it is possible to check if one or several TLE941xy have detected a fault condition by reading the logic OR combination of all the Global Error Flags when CSN goes Low without any clock cycle (**Figure 22**).



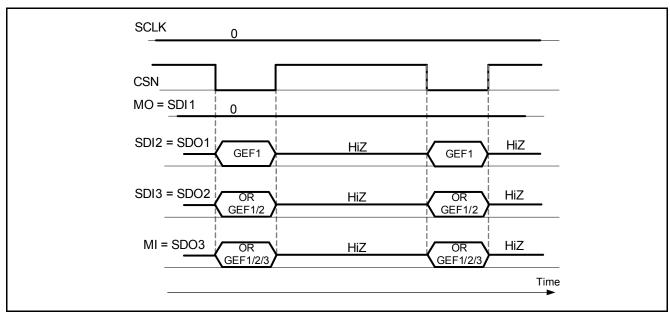


Figure 22 Global Error Flag with zero SCLK clock cycle in daisy chain consisting only of TLE941xy devices

Note:

Some SPI protocol errors such as the LSB of an address byte is wrongly equal to 0, may be reported in the SPI\_ERR bit of another device in the daisy chain (refer to Chapter 7.1.3 and Chapter 7.7 for more details on SPI\_ERR). In this case some devices might accept wrong data during the corrupted SPI frame. Therefore if one of the devices in the daisy chain reports an SPI error, it is recommended to verify the content of the registers of all devices.

#### 7.4 Status register change during SPI communication

If a new failure occurs after the transfer of the data byte(s), i.e. between the end of the last address byte and the CSN rising edge, this failure will be reported in the next SPI frame (see example in **Figure 23**).

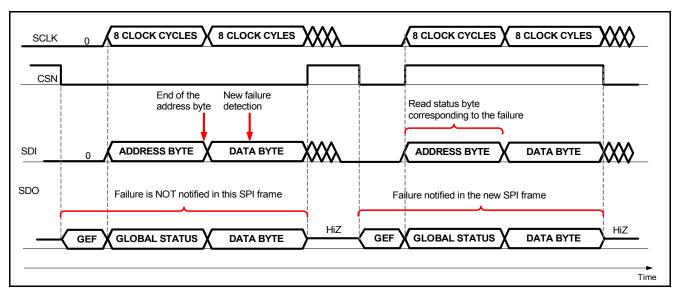


Figure 23 Status register change during transfer of data byte - Example in independent slave configuration

#### **TLE94106EL**



#### **Serial Peripheral Interface (SPI)**

No information is lost, even if a status register is changed during a SPI frame, in particular during a Read and Clear command. For example:

- the microcontroller sends a Read and Clear command to a status register
- the TLE94106EL detects during the transfer the data byte(s) a new fault condition, which is normally reported in the target status register

The incoming Clear command will be ignored, so that the microcontroller can read the new failure in the subsequent SPI frames.

Data inconsistency between the Global Status Register (see **Chapter 7.7**) and the data byte (status register) within the same SPI frame is possible if:

- an open load or overcurrent error is detected during the transfer of the data byte
- · the target status register corresponds to the new detected failure

In this case the new failure:

- is not reported in the Global Status Register of the current SPI frame but in the next one
- is reported in the data byte of the current SPI frame

Refer to Figure 23.



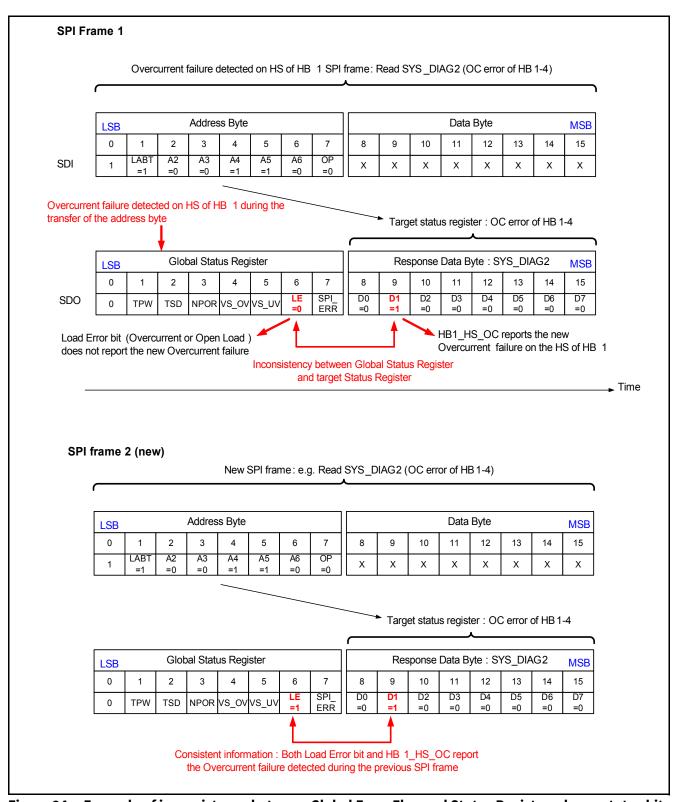


Figure 24 Example of inconsistency between Global Error Flag and Status Register when a status bit is changed during the transfer of an address byte



#### 7.5 SPI Bit Mapping

The SPI Registers have been mapped as shown in Figure 25 and Figure 26 respectively.

The control registers are READ/ WRITE registers. To set the control register to READ, bit 7 of the address byte (OP bit) must be programmed to '0', otherwise '1' for WRITE.

The status registers are READ/CLEAR registers. To CLEAR any Status Register, bit 7 of the address byte must be set to '1', otherwise '0' for READ.

	15	14	13	12	11	10	9	8	7	6 5 4 3 2 1
				8 Data Bits	s [D7D0]				8 Addi Access	ress Bits [A70]
			for Confi	guration &	Status Inf	ormation			type	
				HB_ACT	_1_CTRL				read/write	0 0 0 0 0 LABT
				HB_ACT	_2_CTRL				read/write	1 0 0 0 0 LABT
S					read/write	1 1 0 0 0 LAB				
O L E R					read/write	0 0 1 0 0 LAB				
~ =			F		read/write	0 1 1 0 0 LAB				
N T S I S		PWM1_DC_CTRL								1 1 1 0 0 LAB
C O E G				PWM2_E	C_CTRL				read/write	0 0 0 1 0 LAB
~ ~				PWM3_E	C_CTRL				read/write	1 0 0 1 0 LAB
				FW_OL	_CTRL				read/write	0 1 0 1 0 LAB
				CONFIG	G_CTRL				read	1 1 0 0 1 LAB
R S			SYS	S_DIAG_1 :	Global stat	us 1			read/clear	0 0 1 1 0 LAB
rus Ter			SYS_D	IAG_2 : OF	PERROR_1	_STAT			read/clear	1 0 1 1 0 LAB
A .	SYS_DIAG_3 : OP ERROR_2_STAT read/clear 0 1									0 1 1 1 0 LAB
S T E G	SYS_DIAG_5 : OP ERROR_4_STAT read/clear 0 0 0 0 1 LABT									
œ			SYS_D	IAG_6 : OF	ERROR_5	STAT			read/clear	1 0 0 0 1 LAB

Figure 25 TLE94106EL SPI Register mapping

Note: LABT: Last Address Bit Token, refer to Chapter 7.2 and Chapter 7.3.



_			-	-	_		_									. 1	
1			LABT 1	LABT 1	LABT 1	LABT 1	LABT 1	LABT	LABT	LABT	LABT 1	LABT	Tav I	LABT	Tav	LABT 1	1 ART 1
3 2	۷7A0		0 0	0 0	0 0	0 0	0		0 1	0	0	0 1	-	0			1
5 4	s Bits /		0 0	0	1 0	1	1			0		1	0	- +		- 0	0
9	Addres	be	te 0	1 at	te 1	te 0	0 9	-	0 ·	_	0 ·	_	_	_	+	+	+
7 6 5 4 3 2		Access ty	read/write	read/write	read/write	_	-+	-+	-+	+	read/write	read	roodoloor	read/clear	road/ologi	read/clear	read/clear
8	ć		HB1_LS_EN	HB5_LS_EN	HB1_MODE0	HB5_MODE0	PWM_CH1_FREQ_0	PWM1_DC_CTRL_0	PWM2_DC_CTRL_0	PWM3_DC_CTRL_0	OL_SEL_HS1	DEV_ID0	c	HB1 IS OC	HBF IS OC	HB1 LS OL	HB5_LS_OL
6	ă	PJ	HB1_HS_EN	HB5_HS_EN	HB1_MODE1	HB5_MODE1	PWM_CH1_FREQ_1	PWM1_DC_CTRL_1	PWM2_DC_CTRL_1	PWM3_DC_CTRL_1	OL_SEL_HS2	DEV_ID1	WIGH	HB1 HS OC	HB5 HS OC	HB1 HS OL	HB5_HS_OL
10	ć	D2	HB2_LS_EN	HB6_LS_EN	HB2_MODE0	HB6_MODE0	PWM_CH2_FREQ_0	PWM1_DC_CTRL_2	PWM2_DC_CTRL_2	PWM3_DC_CTRL_2	FW_HB1	DEV_ID2	TST	HB2 I S OC	HB6 I S OC	HB2 LS OL	HB6_LS_OL
11	D7D0	D3 GISTERS	HB2_HS_EN	HB6_HS_EN	HB2_MODE1	HB6_MODE1	PWM_CH2_FREQ_1	PWM1_DC_CTRL_3	PWM2_DC_CTRL_3	PWM3_DC_CTRL_3	FW_HB2	reserved	BOOM	HB2 HS OC	HB6 HS OC	HB2 HS OL	HB6_HS_OL
12	Data Bits D7	CONTROL REGISTERS	HB3_LS_EN	reserved	HB3_MODE0	reserved	PWM_CH3_FREQ_0	PWM1_DC_CTRL_4	PWM2_DC_CTRL_4	PWM3_DC_CTRL_4	FW_HB3	reserved reserved	//O 9/	HB3 I S OC	DO POLICIONAL DE LA COMPANSA DE LA C	HB3 LS OL	reserved
13	ú	DS	HB3_HS_EN	reserved	HB3_MODE1	reserved	PWM_CH3_FREQ_1	PWM1_DC_CTRL_5	PWM2_DC_CTRL_5	PWM3_DC_CTRL_5	FW_HB4	reserved	WI SV	HB3 HS OC	DO D	HB3 HS OF	reserved
14	Ġ.	D6	HB4_LS_EN	reserved	HB4_MODE0	reserved	FM_CLK_MOD0	PWM1_DC_CTRL_6	PWM2_DC_CTRL_6	PWM3_DC_CTRL_6	FW_HB5	reserved	2	HB4 I S OC	respond	HB4 LS OL	reserved
15	6	D7	HB4_HS_EN	reserved	HB4_MODE1	reserved	FM_CLK_MOD1	PWM1_DC_CTRL_7	PWM2_DC_CTRL_7	PWM3_DC_CTRL_7	FW_HB6	reserved	993	HB4 HS OC	00 00 00 00	HB4 HS OL	reserved
	Register Name		HB_ACT_1_CTRL	HB_ACT_2_CTRL	HB_MODE_1_CTRL	HB_MODE_2_CTRL	PWM_CH_FREQ_CTRL	PWM1_DC_CTRL	PWM2_DC_CTRL	PWM3_DC_CTRL	FW_OL_CTRL	CONFIG_CTRL	SVS DIAG 1. Global status 1	SYS DIAG 3 OD EDDOR 1 STAT	SVS DIAG 3 : OB EDBOD 2 STAT	SYS DIAG 5:0P ERROR 4 STAT	SYS_DIAG_6: OP ERROR_5_STAT
					В:	э Т	SI	ອ : 0 I	3 8				В	эт	SI	S T E G	

Figure 26 TLE94106EL Bit Mapping

Note: LABT: Last Address Bit Token, refer to Chapter 7.2 and Chapter 7.3.



#### 7.6 SPI Control Registers

The Control Registers have a READ/WRITE access (see **Chapter 7.5**):

- The 'POR' value is defined by the register content after a POR or device Reset
  - The default value of all control registers is 0000 0000<sub>B</sub> with the exception of CONFIG\_CTRL
  - The default value of the CONFIG\_CTRL register is 0000  $0011_{\rm B}$
- One 16-bit SPI command consists of two bytes (see Figure 25 and Figure 26), i.e.
  - an address byte
  - followed by a data byte
- The control bits are not cleared or changed automatically by the device. This must be done by the microcontroller via SPI programming.
- Reading a register is done byte wise by setting the SPI bit 7 to "0" (= READ ONLY).
- Writing to a register is done byte wise by setting the SPI bit 7 to "1".



# 7.6.1 Control register definition

# HB\_ACT\_1\_CTRL Half-bridge output control 1 (Address Byte [OP] 000 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
HB4_HS_EN	HB4_LS_EN	HB3_HS_EN	HB3_LS_EN	HB2_HS_EN	HB2_LS_EN	HB1_HS_EN	HB1_LS_EN
						ř	
rw							

Field	Bits	Туре	Description
HB4_HS_EN	D7	rw	Half-bridge output 4 high side switch enable  0 <sub>B</sub> HS4 OFF/ High-Z (default value)  1 <sub>B</sub> HS4 ON
HB4_LS_EN	D6	rw	Half-bridge output 4 low side switch enable  0 <sub>B</sub> LS4 OFF/ High-Z (default value)  1 <sub>B</sub> LS4 ON
HB3_HS_EN	D5	rw	Half-bridge output 3 high side switch enable  0 <sub>B</sub> HS3 OFF/ High-Z (default value)  1 <sub>B</sub> HS3 ON
HB3_LS_EN	D4	rw	Half-bridge output 3 low side switch enable  0 <sub>B</sub> LS3 OFF/ High-Z (default value)  1 <sub>B</sub> LS3 ON
HB2_HS_EN	D3	rw	Half-bridge output 2 high side switch enable  0 <sub>B</sub> HS2 OFF/ High-Z (default value)  1 <sub>B</sub> HS2 ON
HB2_LS_EN	D2	rw	Half-bridge output 2 low side switch enable  0 <sub>B</sub> LS2 OFF/ High-Z (default value)  1 <sub>B</sub> LS2 ON
HB1_HS_EN	D1	rw	Half-bridge output 1 high side switch enable  0 <sub>B</sub> HS1 OFF/ High-Z (default value)  1 <sub>B</sub> HS1 ON
HB1_LS_EN	D0	rw	Half-bridge output 1 low side switch enable  0 <sub>B</sub> LS1 OFF/ High-Z (default value)  1 <sub>B</sub> LS1 ON

Note:

The simultaneous activation of both HS and LS switch within a half-bridge is prevented by the digital block to avoid cross current. If both LS\_EN and HS\_EN bits of a given half-bridge are set, the logic turns off this half-bridge.



# HB\_ACT\_2\_CTRL Half-bridge output control 2 (Address Byte [OP]100 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
reserved	reserved	reserved	reserved	HB6_HS_EN	HB6_LS_EN	HB5_HS_EN	HB5_LS_EN
rw	rw	rw	rw	rw	rw	rw	rw

Field	Bits	Туре	Description
reserved	D7	rw	Reserved. Always reads as '0'
reserved	D6	rw	Reserved. Always reads as '0'
reserved	D5	rw	Reserved. Always reads as '0'
reserved	D4	rw	Reserved. Always reads as '0'
HB6_HS_EN	D3	rw	Half-bridge output 6 high side switch enable  0 <sub>B</sub> HS6 OFF/ High-Z (default value)  1 <sub>B</sub> HS6 ON
HB6_LS_EN	D2	rw	Half-bridge output 6 low side switch enable  0 <sub>B</sub> LS6 OFF/ High-Z (default value)  1 <sub>B</sub> LS6 ON
HB5_HS_EN	D1	rw	Half-bridge output 5 high side switch enable  0 <sub>B</sub> HS5 OFF/ High-Z (default value)  1 <sub>B</sub> HS5 ON
HB5_LS_EN	D0	rw	Half-bridge output 5 low side switch enable  0 <sub>B</sub> LS5 OFF/ High-Z (default value)  1 <sub>B</sub> LS5 ON

Note:

The simultaneous activation of both HS and LS switch within a half-bridge is prevented by the digital block to avoid cross current. If both LS\_EN and HS\_EN bits of a given half-bridge are set, the logic turns off this half-bridge.



# HB\_MODE\_1\_CTRL Half-bridge output mode control 1 (Address Byte [OP]110 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
HB4_MODE1	HB4_MODE0	HB3_MODE1	HB3_MODE0	HB2_MODE1	HB2_MODE0	HB1_MODE1	HB1_MODE0
rw							

Field	Bits	Туре	Description
HB4_MODEn	D7:D6	rw	Half-bridge output 4 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3
HB3_MODEn	D5:D4	rw	Half-bridge output 3 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3
HB2_MODEn	D3:D2	rw	Half-bridge output 2 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3
HB1_MODEn	D1:D0	rw	Half-bridge output 1 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3



# HB\_MODE\_2\_CTRL Half-bridge output mode control 2 (Address Byte [OP]001 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
reserved	reserved	reserved	reserved	HB6_MODE1	HB6_MODE0	HB5_MODE1	HB5_MODE0
rw	rw	rw	rw	rw	rw	rw	rw

Field	Bits	Туре	Description
reserved	D7:D6	rw	Reserved. Always reads as '0'.
reserved	D5:D4	rw	Reserved. Always reads as '0'.
HB6_MODEn	D3:D2	rw	Half-bridge output 6 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3
HB5_MODEn	D1:D0	rw	Half-bridge output 5 mode select
(n = 0,1)			00 <sub>B</sub> No PWM (default value)
			01 <sub>B</sub> PWM control with PWM Channel 1
			10 <sub>B</sub> PWM control with PWM Channel 2
			11 <sub>B</sub> PWM control with PWM Channel 3



# PWM\_CH\_FREQ\_CTRL PWM channel frequency select (Address Byte [OP]011 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
FM_CLK_ MOD1	FM_CLK_ MOD0	PWM_CH3_F REQ_1	PWM_CH3_F REQ_0	PWM_CH2_F REQ_1	PWM_CH2_F REQ_0	PWM_CH1_F REQ_1	PWM_CH1_F REQ_0
rw	rw	rw	rw	rw	rw	rw	rw

Field	Bits	Type	Description
FM_MOD_EN	D7:D6	rw	FM Modulation Enable <sup>1)</sup>
			00 <sub>B</sub> No modulation (default)
			01 <sub>B</sub> Modulation frequency 15.625kHz
			10 <sub>B</sub> Modulation frequency 31.25kHz
			11 <sub>B</sub> Modulation frequency 62.5kHz
PWM_CH3_FREQ_	D5:D4	rw	PWM Channel 3 frequency select
n (n=0,1)			00 <sub>B</sub> PWM is stopped and off (default value)
			01 <sub>B</sub> PWM frequency 1:80Hz
			10 <sub>B</sub> PWM frequency 2: 100Hz
			11 <sub>B</sub> PWM frequency 3: 200Hz
PWM_CH2_FREQ_	D3:D2	rw	PWM Channel 2 frequency select
n (n=0,1)			00 <sub>B</sub> PWM is stopped and off (default value)
			01 <sub>B</sub> PWM frequency 1:80Hz
			10 <sub>B</sub> PWM frequency 2: 100Hz
			11 <sub>B</sub> PWM frequency 3: 200Hz
PWM_CH1_FREQ_	D1:D0	rw	PWM Channel 1 frequency select
n (n=0,1)			00 <sub>B</sub> PWM is stopped and off (default value)
			01 <sub>B</sub> PWM frequency 1:80Hz
			10 <sub>B</sub> PWM frequency 2: 100Hz
			11 <sub>B</sub> PWM frequency 3: 200Hz

<sup>1)</sup> Not subject to production test, guaranteed by design. Frequency may deviate by ±10%



#### PWM1\_DC\_CTRL

### PWM channel 1 duty cycle configuration (Address Byte [OP]111 00[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
PWM1_DC_ CTRL_7	PWM1_DC_ CTRL_6	PWM1_DC_ CTRL_5	PWM1_DC_ CTRL_4	PWM1_DC_ CTRL_3	PWM1_DC_ CTRL_2	PWM1_DC_ CTRL_1	PWM1_DC_ CTRL_0
rw							

Field	Bits	Туре	Description
PWM1_DC_CTRLn	D7:D0	rw	PWM Channel 1 Duty Cycle configuration (bit7=MSB; bit0)  0000 0000 <sub>R</sub> 100% OFF (default value)
			xxxx xxxx <sub>B</sub> parts of 255 ON 1111 1111 <sub>B</sub> 100% ON

Note: Refer to Chapter 6.1.1 for more information on PWM operation

### PWM2\_DC\_CTRL

#### PWM channel 2 duty cycle configuration (Address [OP]000 10[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
PWM2_DC_ CTRL_7	PWM2_DC_ CTRL_6	PWM2_DC_ CTRL_5	PWM2_DC_ CTRL_4	PWM2_DC_ CTRL_3	PWM2_DC_ CTRL_2	PWM2_DC_ CTRL_1	PWM2_DC_ CTRL_0
rw							

Field	Bits	Туре	Description
PWM2_DC_CTRLn	D7:D0	rw	PWM Channel 2 Duty Cycle configuration (bit7=MSB; bit0)
			0000 0000 <sub>B</sub> 100% OFF (default value) xxxx xxxx <sub>B</sub> parts of 255 ON 1111 1111 <sub>B</sub> 100% ON

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# PWM3\_DC\_CTRL PWM channel 3 duty cycle configuration (Address Byte [OP]100 10[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
PWM3_DC_ CTRL_7	PWM3_DC_ CTRL_6	PWM3_DC_ CTRL_5	PWM3_DC_ CTRL_4	PWM3_DC_ CTRL_3	PWM3_DC_ CTRL_2	PWM3_DC_ CTRL_1	PWM3_DC_ CTRL_0
rw							

Field	Bits	Туре	Description
PWM3_DC_CTRLn	D7:D0	rw	PWM Channel 3 Duty Cycle configuration (bit7=MSB; bit0)
			0000 0000 <sub>B</sub> 100% OFF (default value) xxxx xxxx <sub>B</sub> parts of 255 ON 1111 1111 <sub>B</sub> 100% ON



 $FW_OL\_CTRL\\ Free-wheeling configuration and Open load detection setting of HS1 and HS2 (Address Byte [OP]010\\ 10[LABT]1_B)$ 

D7	D6	D5	D4	D3	D2	D1	D0
FW_HB6	FW_HB5	FW_HB4	FW_HB3	FW_HB2	FW_HB1	OL_SEL_HS2	OL_SEL_HS1
rw	rw						

Field	Bits	Туре	Description
FW_HB6	D7	rw	HB6 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
FW_HB5	D6	rw	HB5 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
FW_HB4	D5	rw	HB4 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
FW_HB3	D4	rw	HB3 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
FW_HB2	D3	rw	HB2 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
FW_HB1	D2	rw	HB1 free-wheeling configuration  0 <sub>B</sub> Passive free-wheeling (default value)  1 <sub>B</sub> Active free-wheeling
OL_SEL_HS2	D1	rw	HS2 open load detection current and filter time select  0 <sub>B</sub> High-current mode (default value)  1 <sub>B</sub> LED Mode (Low current mode)
OL_SEL_HS1	D0	rw	HS1 open load detection current and filter time select  0 <sub>B</sub> High current mode (default value)  1 <sub>B</sub> LED Mode (Low current mode)

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### **Serial Peripheral Interface (SPI)**

# CONFIG\_CTRL Device Configuration control (Address Byte [OP]110 01[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
reserved	reserved	reserved	reserved	reserved	DEV_ID2	DEV_ID1	DEV_ID0
r	r	r	r	r	r	r	r

Field	Bits	Type	Description		
reserved	D7:D3	r	Always reads as '0'		
DEV_IDn	D2:D0	r	Device/ derivative identifier		
			Note: These bits can be used to verify the silicon content of the device		
			000 <sub>B</sub> TLE94112EL chip		
			001 <sub>B</sub> TLE94110EL chip		
			010 <sub>B</sub> TLE94108EL chip		
			011 <sub>B</sub> TLE94106EL chip		
			100 <sub>B</sub> TLE94104EP chip		
			101 <sub>B</sub> TLE94103EP chip		
			110 <sub>B</sub> reserved		
			111 <sub>B</sub> reserved		



#### 7.7 SPI Status Registers

The Control Registers have a READ/CLEAR access (see also **Chapter 7.5**):

- The 'POR Value' of the Status registers (content after a POR or device Reset) and is 0000 0000<sub>B</sub>.
- One 16-bit SPI command consists of two bytes (see Figure 25 and Figure 26), i.e.
  - an address byte
  - followed by a data byte
- Reading a register is done byte wise by setting the SPI bit 7 of the address byte to "0" (= Read Only).
- Clearing a register is done byte wise by setting the SPI bit 7 of the address byte to "1".
- SPI status registers are not cleared automatically by the device. This must be done by the microcontroller via SPI command.



## 7.7.1 Status register definition

# SYS\_DIAG1 Global status 1 (Address Byte [OP]001 10[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
SPI_ERR	LE	VS_UV	vs_ov	NPOR	TSD	TPW	reserved
rc	r	rc	rc	rc	rc	rc	r

Field	Bits	Type	Description
SPI_ERR	D7	rc	SPI error detection  0 <sub>B</sub> No SPI protocol error is detected (default value).  1 <sub>B</sub> An SPI protocol error is detected.
LE	D6	r	Load error detection (logic OR combination of Open Load and Overcurrent)  0 <sub>B</sub> No Open Load and no Overcurrent detected (default value)  1 <sub>B</sub> Open Load or Overcurrent detected in at least one of the power outputs. Error latched. Faulty output is latched off in case of Overcurrent
VS_UV	D5	rc	$ \begin{array}{ll} \textbf{VS Undervoltage error detection} \\ \textbf{0}_{\text{B}} & \text{No undervoltage on } \textit{V}_{\text{S}} \text{ detected (default value)} \\ \textbf{1}_{\text{B}} & \text{Undervoltage on } \textit{V}_{\text{S}} \text{ detected. Error latched and all outputs} \\ & \text{disabled.} \end{array} $
VS_OV	D4	rc	$\begin{array}{ll} \textbf{VS Overvoltage error detection} \\ \textbf{0}_{\text{B}} & \text{No overvoltage on } V_{\text{S}} \text{ detected (default value)} \\ \textbf{1}_{\text{B}} & \text{Overvoltage on } V_{\text{S}} \text{ detected. Error latched and all outputs} \\ & \text{disabled.} \end{array}$
NPOR	D3	rc	Not Power On Reset (NPOR) detection  0 <sub>B</sub> POR on EN or VDD supply rail (default value)  1 <sub>B</sub> No POR
TSD	D2	rc	Temperature shutdown error detection  0 <sub>B</sub> Junction temperature below temperature shutdown threshold (default value)  1 <sub>B</sub> Junction temperature has reached temperature shutdown threshold. Error latched and all outputs disabled.
TPW	D1	rc	Temperature pre-warning error detection  0 <sub>B</sub> Junction temperature below temperature pre-warning threshold (default value)  1 <sub>B</sub> Junction temperature has reached temperature pre-warning threshold.
reserved	D0	r	Bit reserved. Always reads '0'.

Note:

The LE bit in the Global Status register is read only. It reflects an OR combination of the respective open load and overcurrent errors of the half-bridge channels. If all OC/OL bits of the respective high-side and low-side channels are cleared to '0', the LE bit will be automatically updated to '0'.



# SYS\_DIAG\_2: OP\_ERROR\_1\_STAT Overcurrent error status of half-bridge outputs 1 - 4 (Address Byte [OP]101 10[LABT]1<sub>B</sub>)

	D7	D6	D5	D4	D3	D2	D1	D0
F	HB4_HS_OC	HB4_LS_OC	HB3_HS_OC	HB3_LS_OC	HB2_HS_OC	HB2_LS_OC	HB1_HS_OC	HB1_LS_OC
	rc							

Field	Bits	Type	Description			
HB4_HS_OC	D7	rc	High-side (HS) switch of half-bridge 4 overcurrent detection  0 <sub>B</sub> No error on HS4 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS4 switch. Error latched and HS4 disabled.			
HB4_LS_OC	D6	rc	Low-side (LS) switch of half-bridge 4 overcurrent detection  0 <sub>B</sub> No error on LS4 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS4 switch. Error latched and LS4 disabled.			
HB3_HS_OC	D5	rc	High-side (HS) switch of half-bridge 3 overcurrent detection  0 <sub>B</sub> No error on HS3 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS3 switch. Error latched and HS3 disabled.			
HB3_LS_OC	D4	rc	Low-side (LS) switch of half-bridge 3 overcurrent detection  0 <sub>B</sub> No error on LS3 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS3 switch. Error latched and LS3 disabled.			
HB2_HS_OC	D3	rc	High-side (HS) switch of half-bridge 2 overcurrent detection  0 <sub>B</sub> No error on HS2 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS2 switch. Error latched and HS2 disabled.			
HB2_LS_OC	D2	rc	Low-side (LS) switch of half-bridge 2 overcurrent detection  0 <sub>B</sub> No error on LS2 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS2 switch. Error latched and LS2 disabled.			
HB1_HS_OC	D1	rc	High-side (HS) switch of half-bridge 1 overcurrent detection  0 <sub>B</sub> No error on HS1 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS1 switch. Error latched and HS1 disabled.			
HB1_LS_OC	D0	rc	Low-side (LS) switch of half-bridge 1 overcurrent detection  0 <sub>B</sub> No error on LS1 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS1 switch. Error latched and LS1 disabled.			



# SYS\_DIAG\_3: OP\_ERROR\_2\_STAT Overcurrent error status of half-bridge outputs 5 - 8 (Address Byte [OP]011 10[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
reserved	reserved	reserved	reserved	HB6_HS_OC	HB6_LS_OC	HB5_HS_OC	HB5_LS_OC
rc	rc	rc	rc	rc	rc	rc	rc

Field	Bits	Type	Description			
reserved	D7	rc	Reserved. Always reads as '0'.			
reserved	D6	rc	Reserved. Always reads as '0'.			
reserved	D5	rc	Reserved. Always reads as '0'.			
reserved	D4	rc	Reserved. Always reads as '0'.			
HB6_HS_OC	D3	rc	High-side (HS) switch of half-bridge 6 overcurrent detection  0 <sub>B</sub> No error on HS6 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS6 switch. Error latched and HS6 disabled.			
HB6_LS_OC	D2	rc	Low-side (LS) switch of half-bridge 6 overcurrent detection  0 <sub>B</sub> No error on LS6 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS6 switch. Error latched and LS6 disabled.			
HB5_HS_OC	D1	rc	High-side (HS) switch of half-bridge 5 overcurrent detection  0 <sub>B</sub> No error on HS5 switch (default value)  1 <sub>B</sub> Overcurrent detected on HS5 switch. Error latched and HS5 disabled.			
HB5_LS_OC	D0	rc	Low-side (LS) switch of half-bridge 5 overcurrent detection  0 <sub>B</sub> No error on LS5 switch (default value)  1 <sub>B</sub> Overcurrent detected on LS5 switch. Error latched and LS5 disabled.			



# SYS\_DIAG\_5: OP\_ERROR\_4\_STAT Open load error status of half-bridge outputs 1 - 4 (Address Byte [OP]000 01[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
HB4_HS_OL	HB4_LS_OL	HB3_HS_OL	HB3_LS_OL	HB2_HS_OL	HB2_LS_OL	HB1_HS_OL	HB1_LS_OL
rc							

Field	Bits	Type	Description
HB4_HS_OL	D7	rc	High-side (HS) switch of half-bridge 4 open load detection  0 <sub>B</sub> No error on HS4 switch (default value)  1 <sub>B</sub> Open load detected on HS4 switch. Error latched.
HB4_LS_OL	D6	rc	Low-side (LS) switch of half-bridge 4 open load detection  0 <sub>B</sub> No error on LS4 switch (default value)  1 <sub>B</sub> Open load detected on LS4 switch. Error latched.
HB3_HS_OL	D5	rc	High-side (HS) switch of half-bridge 3 open load detection $0_B$ No error on HS3 switch (default value) $1_B$ Open load detected on HS3 switch. Error latched.
HB3_LS_OL	D4	rc	Low-side (LS) switch of half-bridge 3 open load detection  0 <sub>B</sub> No error on LS3 switch (default value)  1 <sub>B</sub> Open load detected on LS3 switch. Error latched.
HB2_HS_OL	D3	rc	High-side (HS) switch of half-bridge 2 open load detection  0 <sub>B</sub> No error on HS2 switch (default value)  1 <sub>B</sub> Open load detected on HS2 switch. Error latched.
HB2_LS_OL	D2	rc	Low-side (LS) switch of half-bridge 2 open load detection  0 <sub>B</sub> No error on LS2 switch (default value)  1 <sub>B</sub> Open load detected on LS2 switch. Error latched.
HB1_HS_OL	D1	rc	High-side (HS) switch of half-bridge 1 open load detection  0 <sub>B</sub> No error on HS1 switch (default value)  1 <sub>B</sub> Open load detected on HS1 switch. Error latched.
HB1_LS_OL	D0	rc	Low-side (LS) switch of half-bridge 1 open load detection  0 <sub>B</sub> No error on LS1 switch (default value)  1 <sub>B</sub> Open load detected on LS1 switch. Error latched.



# SYS\_DIAG\_6: OP\_ERROR\_5\_STAT Open load error status of half-bridge outputs 5 - 8 (Address Byte [OP]100 01[LABT]1<sub>B</sub>)

D7	D6	D5	D4	D3	D2	D1	D0
reserved	reserved	reserved	reserved	HB6_HS_OL	HB6_LS_OL	HB5_HS_OL	HB5_LS_OL
rc	rc	rc	rc	rc	rc	rc	rc

Field	Bits	Type	Description				
reserved	D7	rc	Reserved. Always reads as '0'.				
reserved	D6	rc	Reserved. Always reads as '0'.				
reserved	D5	rc	Reserved. Always reads as '0'.				
reserved	D4	rc	Reserved. Always reads as '0'.				
HB6_HS_OL	D3	rc	High-side (HS) switch of half-bridge 6 open load detection $0_B$ No error on HS6 switch (default value) $1_B$ Open load detected on HS6 switch. Error latched.				
HB6_LS_OL	D2	rc	Low-side (LS) switch of half-bridge 6 open load detection  0 <sub>B</sub> No error on LS6 switch (default value)  1 <sub>B</sub> Open load detected on LS6 switch. Error latched.				
HB5_HS_OL	D1	rc	High-side (HS) switch of half-bridge 5 open load detection $0_B$ No error on HS5 switch (default value) $1_B$ Open load detected on HS5 switch. Error latched.				
HB5_LS_OL	D0	rc	Low-side (LS) switch of half-bridge 5 open load detection  0 <sub>B</sub> No error on LS5 switch (default value)  1 <sub>B</sub> Open load detected on LS5 switch. Error latched.				



#### **Application Information**

# **8** Application Information

Note:

The following simplified application examples are given as a hint for the implementation of the device only and shall not be regarded as a description or warranty of a certain functionality, condition or quality of the device. The function of the described circuits must be verified in the real application.

### 8.1 Application Diagram

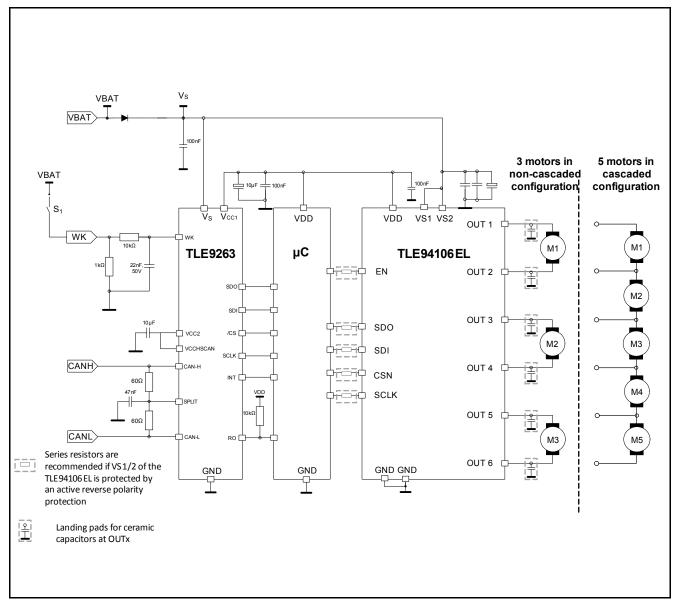


Figure 27 Application example for DC-motor loads



#### **Application Information**

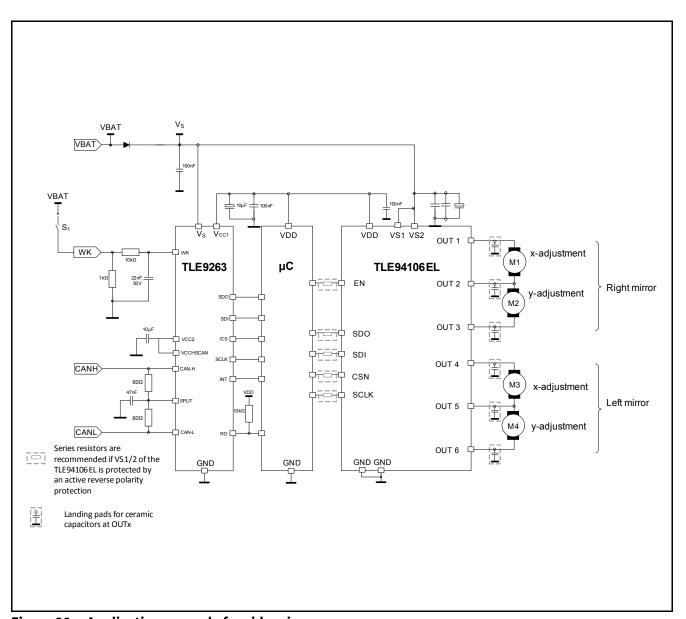


Figure 28 Application example for side mirrors

#### Notes on the application example

- Series resistors between the microcontroller and the signal pins of the TLE94106EL are recommended if an
  active reverse polarity protection (MOSFET) is used to protect VS1 and VS2 pins. These resistors limit the
  current between the microcontroller and the device during negative transients on VBAT (e.g. ISO/TR 7637
  pulse 1)
- Landing pads for ceramic capacitors at the outputs of the TLE94106EL as close as possible to the connectors
  are recommended (the ceramic capacitors are not populated if unused). These ceramic capacitors can be
  mounted if a higher performance in term of ESD capability is required.
- The electrolytic capacitor at the VSx pins should be dimensioned in order to prevent the VS voltage from
  exceeding the absolute maximum rating. PWM operation with a too low capacitance can lead to a VS voltage
  overshoot, which results in a VS overvoltage detection.
- 4. Unused outputs are recommended to be left unconnected (open) in the application. If unused output pins are routed to an external connector which leaves the PCB, then these outputs should have provision for a zero ohm jumper (depopulated if unused) or ESD protection. In other words, unused pins should be treated like used pins.

## **TLE94106EL**



#### **Application Information**

5. Place bypass ceramic capacitors as close as possible to the VSx pins, with shortest connections the GND pins and GND layer, for best EMC performance



#### **Application Information**

# 8.2 Thermal application information

Ta = -40°C, Ch1 to Ch6 are dissipating a total of 1.2W (0.2W each).

Ta = 85°C, Ch1 to Ch6 are dissipating a total of 0.81W (0.135W each).

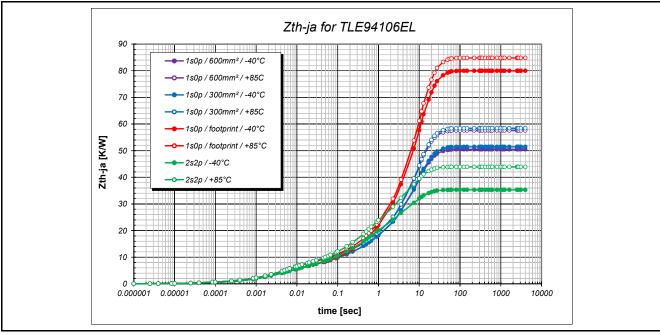


Figure 29 ZthJA Curve for different PCB setups

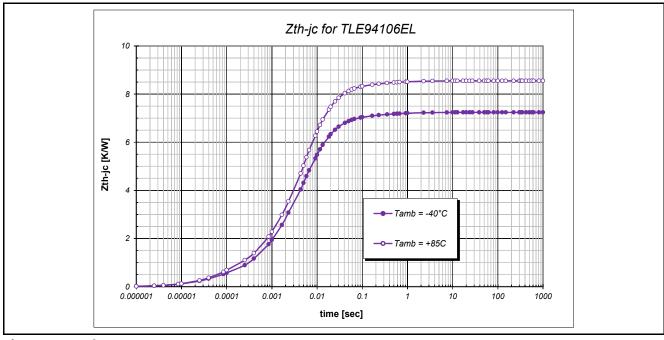


Figure 30 ZthJC Curve

#### **TLE94106EL**



#### **Application Information**

#### 8.3 EMC Enhancement

In the event the emissions of the device exceed the allowable limits, a modulation of the oscillator frequency is incorporated to reduce eventual harmonics of the 8MHz base clock. The frequencies can be selected based on the resolution bandwidth of the peak detector during EMC testing.

The selection is achieved by setting the FM\_CLK\_MODn bits in the PWM\_CH\_FREQ\_CTRL register as follows:

00<sub>B</sub>: OFF

 $01_{\rm B}$ : FM CLK=15.625 kHZ

10<sub>B</sub>: FM CLK=31.25 kHz

 $11_B$ : FM CLK=62.5 kHz



#### **Package Outlines**

# 9 Package Outlines

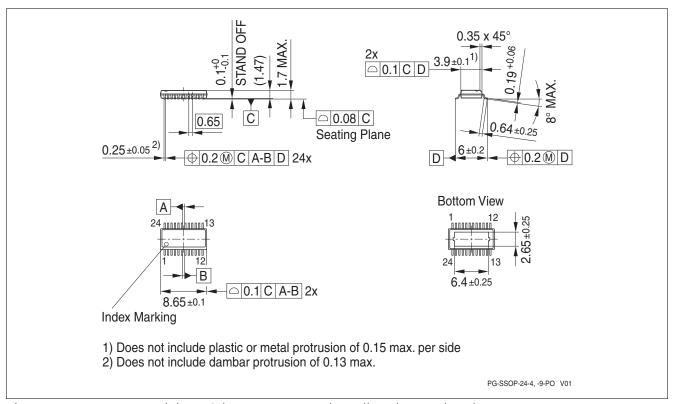


Figure 31 PG-SSOP-24 (Plastic/Plastic Green - Dual Small Outline Package)

#### **Green Product (RoHS compliant)**

To meet the world-wide customer requirements for environmentally friendly products and to be compliant with government regulations the device is available as a green product. Green products are RoHS-Compliant (i.e lead-free finish on leads and suitable for Pb-free soldering according to IPC/JEDEC J-STD-020).

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# **TLE94106EL**



### **Revision History**

# 10 Revision History

Revision	Date	Changes
1.0	2016-09-08	Initial release

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