

# STK581U3C2D-E

## Intelligent Power Module (IPM) 600 V, 30 A



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### Overview

This “Inverter IPM” is highly integrated device containing all High Voltage (HV) control from HV-DC to 3-phase outputs in a single SIP module (Single-In line Package). Output stage uses IGBT/FRD technology and implements Under Voltage Protection (UVP) and Over Current Protection (OCP) with a Fault Detection output flag. Internal Boost diodes are provided for high side gate boost drive.

### Function

- Single control power supply due to Internal bootstrap circuit for high side pre-driver circuit
- All control input and status output are at low voltage levels directly compatible with microcontrollers
- Built-in cross conduction prevention
- Externally accessible embedded thermistor for substrate temperature measurement
- The level of the over-current protection current is adjustable with the external resistor, “RSD”

### Certification

- UL1557 (File Number : E339285)

### Specifications

#### Absolute Maximum Ratings at Tc = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V <sub>CC</sub>	P to N, surge < 500 V *1	450	V
Collector-emitter voltage	V <sub>CE</sub>	P to U,V,W or U,V,W to N	600	V
Output current	I <sub>o</sub>	P, N, U,V,W terminal current	±30	A
		P, N, U,V,W terminal current at Tc = 100°C	±15	A
Output peak current	I <sub>op</sub>	P, N, U,V,W terminal current for a Pulse width of 1 ms.	±45	A
Pre-driver voltage	VD1,2,3,4	VB1 to U, VB2 to V, VB3 to W, V <sub>DD</sub> to V <sub>SS</sub> *2	20	V
Input signal voltage	V <sub>IN</sub>	HIN1, 2, 3, LIN1, 2, 3	-0.3 to V <sub>DD</sub>	V
FAULT terminal voltage	V <sub>FAULT</sub>	FAULT terminal	-0.3 to V <sub>DD</sub>	V
Maximum power dissipation	P <sub>d</sub>	IGBT per channel	49	W
Junction temperature	T <sub>j</sub>	IGBT, FRD	150	°C
Storage temperature	T <sub>stg</sub>		-40 to +125	°C
Operating case temperature	T <sub>c</sub>	IPM case temperature	-40 to +100	°C
Tightening torque		Case mounting screws *3	1.17	Nm
Withstand voltage	V <sub>is</sub>	50 Hz sine wave AC 1 minute *4	2000	VRMS

Reference voltage is “V<sub>SS</sub>” terminal voltage unless otherwise specified.

\*1 : Surge voltage developed by the switching operation due to the wiring inductance between “P” and “N” terminal.

\*2 : Terminal voltage : VD1 = VB1 to U, VD2 = VB2 to V, VD3 = VB3 to W, VD4 = V<sub>DD</sub> to V<sub>SS</sub>

\*3 : Flatness of the heat-sink should be 0.15 mm and below.

\*4 : Test conditions : AC 2500 V, 1 s.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

### ORDERING INFORMATION

See detailed ordering and shipping information on page 15 of this data sheet.

# STK581U3C2D-E

## Electrical Characteristics at Tc = 25°C, VD1, VD2, VD3, VD4 = 15 V

Parameter	Symbol	Conditions	Test circuit	min	typ	max	Unit	
<b>Power output section</b>								
Collector-emitter cut-off current	I <sub>CE</sub>	V <sub>CE</sub> = 600 V	Fig.1	-	-	0.1	mA	
Bootstrap diode reverse current	I <sub>R(BD)</sub>	V <sub>R(BD)</sub>		-	-	0.1	mA	
Collector to emitter saturation voltage	V <sub>CE(SAT)</sub>	I <sub>c</sub> = 30 A T <sub>J</sub> = 25°C	Upper side	Fig.2	-	1.8	2.7	V
			Lower side *1		-	2.1	3.0	
		I <sub>c</sub> = 15 A T <sub>J</sub> = 100°C	Upper side		-	1.5	-	
			Lower side *1		-	1.7	-	
Diode forward voltage	V <sub>F</sub>	I <sub>F</sub> = 30 A T <sub>J</sub> = 25°C	Upper side	Fig.3	-	2.0	2.9	V
			Lower side *1		-	2.3	3.2	
		I <sub>F</sub> = 15 A T <sub>J</sub> = 100°C	Upper side		-	1.5	-	
			Lower side *1		-	1.7	-	
Junction to case thermal resistance	θ <sub>j-c(T)</sub>	IGBT		-	-	2.5	°C/W	
	θ <sub>j-c(D)</sub>	FRD		-	-	3		
<b>Control (Pre-driver) section</b>								
Pre-driver power dissipation	I <sub>D</sub>	VD1, 2, 3 = 15 V	Fig.4	-	0.08	0.4	mA	
		VD4 = 15 V		-	1.6	4		
High level Input voltage	V <sub>in H</sub>	HIN1, HIN2, HIN3, LIN1, LIN2, LIN3 to V <sub>SS</sub>		2.5	-	-	V	
Low level Input voltage	V <sub>in L</sub>			-	-	0.8	V	
Input threshold voltage hysteresis*1	V <sub>inth(hys)</sub>			0.5	0.8	-	V	
Logic 1 input leakage current	I <sub>IN+</sub>	V <sub>IN</sub> = +3.3 V		-	100	143	μA	
Logic 0 input leakage current	I <sub>IN-</sub>	V <sub>IN</sub> = 0 V		-	-	2	μA	
FAULT terminal input electric current	I <sub>oSD</sub>	FAULT : ON / V <sub>FAULT</sub> = 0.1 V		-	2	-	mA	
FAULT clear time	FLTCLR	Fault output latch time.		18	-	80	ms	
V <sub>CC</sub> and V <sub>S</sub> undervoltage positive going threshold.	V <sub>CCUV+</sub> V <sub>SUV+</sub>			10.5	11.1	11.7	V	
V <sub>CC</sub> and V <sub>S</sub> undervoltage negative going threshold.	V <sub>CCUV-</sub> V <sub>SUV-</sub>			10.3	10.9	11.5	V	
V <sub>CC</sub> and V <sub>S</sub> undervoltage hysteresis	V <sub>CCUVH</sub> V <sub>SUVH-</sub>			0.14	0.2	-	V	
Over current protection level	ISD	PW = 100 μs, RSD = 0 Ω	Fig.5	38.5	-	48.2	A	
Output level for current monitor	ISO	I <sub>o</sub> = 30 A		0.32	0.34	0.36	V	

Reference voltage is "V<sub>SS</sub>" terminal voltage unless otherwise specified.

\*1 : The lower side's V<sub>CE(SAT)</sub> and V<sub>F</sub> include a loss by the shunt resistance

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

## STK581U3C2D-E

**Electrical Characteristics** at  $T_c = 25^\circ\text{C}$ ,  $V_{D1}, V_{D2}, V_{D3}, V_{D4} = 15\text{ V}$ ,  $V_{CC} = 300\text{ V}$ ,  $L = 3.5\text{ mH}$

Parameter	Symbol	Conditions	Test circuit	min	typ	max	Unit
<b>Switching Character</b>							
Switching time	tON	$I_o = 30\text{ A}$	Fig.6	0.3	0.6	1.3	$\mu\text{s}$
	tOFF			-	0.9	1.6	
Turn-on switching loss	Eon	$I_o = 30\text{ A}$		-	800	-	$\mu\text{J}$
Turn-off switching loss	Eoff			-	550	-	$\mu\text{J}$
Total switching loss	Etot			-	1350	-	$\mu\text{J}$
Turn-on switching loss	Eon	$I_o = 15\text{ A}, T_c = 100^\circ\text{C}$		-	530	-	$\mu\text{J}$
Turn-off switching loss	Eoff			-	450	-	$\mu\text{J}$
Total switching loss	Etot			-	980	-	$\mu\text{J}$
Diode reverse recovery energy	Erec	$I_F = 15\text{ A}, P = 400\text{ V}, T_c = 100^\circ\text{C}$		-	24	-	$\mu\text{J}$
Diode reverse recovery time	trr		-	58	-	ns	
Reverse bias safe operating area	RBSOA	$I_o = 45\text{ A}, V_{CE} = 450\text{ V}$		Full square			
Short circuit safe operating area	SCSOA	$V_{CE} = 400\text{ V}, T_c = 100^\circ\text{C}$		4	-	-	$\mu\text{s}$
Allowable offset voltage slew rate	dv/dt	Between U, V, W to N		-50	-	50	V/ns

Reference voltage is " $V_{SS}$ " terminal voltage unless otherwise specified.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

### Notes :

- When the internal protection circuit operates, a Fault signal is turned ON (When the Fault terminal is low level, Fault signal is ON state : output form is open DRAIN) but the Fault signal does not latch. After protection operation ends, it returns automatically within about 18 ms to 80 ms and resumes operation beginning condition. So, after Fault signal detection, set all input signals to OFF (Low) at once. However, the operation of pre-drive power supply low voltage protection (UVLO : with hysteresis about 0.2 V) is as follows.

#### Upper side :

The gate is turned off and will return to regular operation when recovering to the normal voltage, but the latch will continue till the input signal will turn 'low'.

#### Lower side :

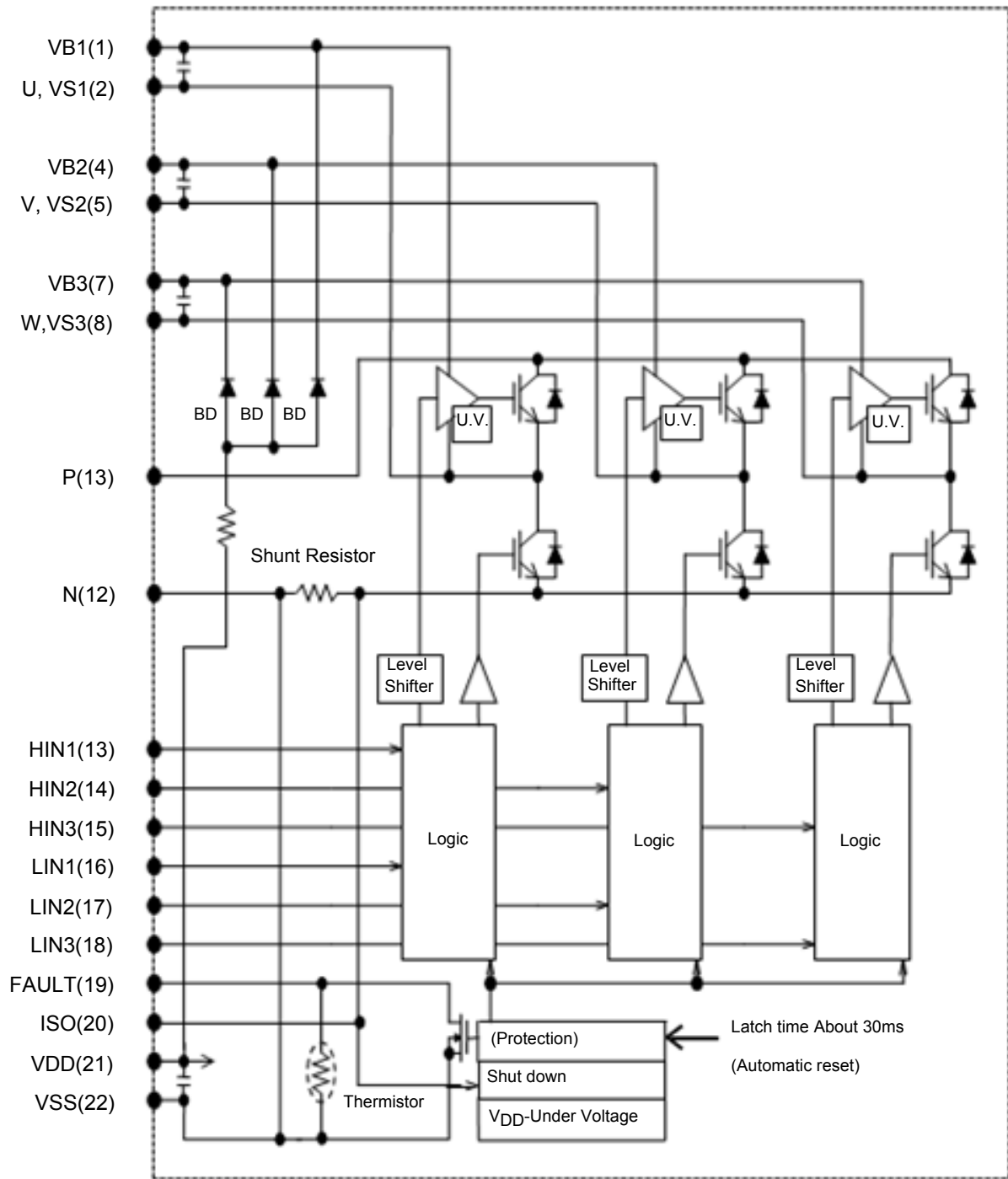
The gate is turned off and will automatically reset when recovering to normal voltage. It does not depend on input signal voltage.

- When assembling the IPM on the heat sink with M3 type screw, tightening torque range is 0.79 Nm to 1.17 Nm.
- The pre-drive low voltage protection is the feature to protect devices when the pre-driver supply voltage falls due to an operating malfunction.

**Module Pin-Out Description**

<b>Pin</b>	<b>Name</b>	<b>Description</b>
1	VB1	High Side Floating Supply Voltage 1
2	U, VS1	Output 1 - High Side Floating Supply Offset Voltage
3	–	Without Pin
4	VB2	High Side Floating Supply voltage 2
5	V, VS2	Output 2 - High Side Floating Supply Offset Voltage
6	–	Without Pin
7	VB3	High Side Floating Supply voltage 1
8	W, VS3	Output 1 - High Side Floating Supply Offset Voltage
9	–	Without Pin
10	P	Positive Bus Input Voltage
11	–	Without Pin
12	N	Positive Bus Input Voltage
13	–	Without Pin
14	HIN1	Logic Input High Side Gate Driver - Phase U
15	HIN2	Logic Input High Side Gate Driver - Phase V
16	HIN3	Logic Input High Side Gate Driver - Phase W
17	LIN1	Logic Input Low Side Gate Driver - Phase U
18	LIN2	Logic Input Low Side Gate Driver - Phase V
19	LIN3	Logic Input Low Side Gate Driver - Phase W
20	ISO	Current monitor output
21	VDD	+15V Main Supply
22	VSS	Negative Main Supply

Equivalent Block Diagram



# STK581U3C2D-E

## Test Circuit

(The tested phase : U+ shows the upper side of the U phase and U- shows the lower side of the U phase.)

### ■ ICE / IR(BD)

	U+	V+	W+	U-	V-	W-
M	10	10	10	2	5	8
N	2	5	8	12	12	12

	U(BD)	V(BD)	W(BD)
M	1	4	7
N	22	22	22

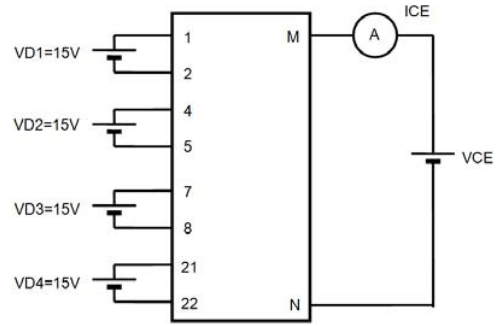


Fig. 1

### ■ VCE(SAT) (Test by pulse)

	U+	V+	W+	U-	V-	W-
M	10	10	10	2	6	8
N	2	5	8	12	12	12
m	13	14	15	16	17	18

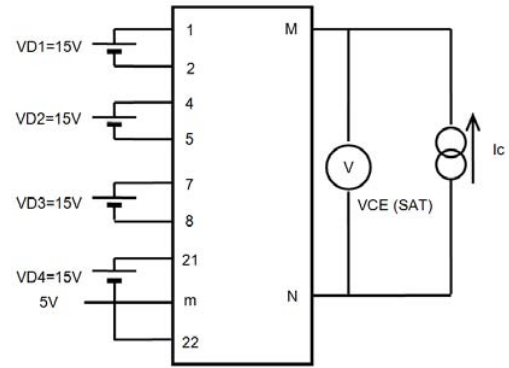


Fig. 2

### ■ VF (Test by pulse)

	U+	V+	W+	U-	V-	W-
M	10	10	10	2	5	8
N	2	5	8	12	12	12

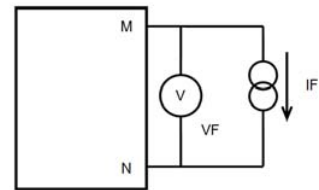


Fig. 3

### ■ ID

	VD1	VD2	VD3	VD4
M	1	4	7	21
N	2	5	8	22

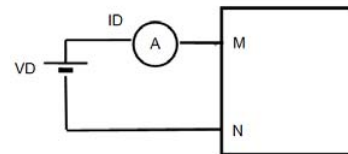


Fig. 4

■ ISD

Input signal

(0 to 5 V)

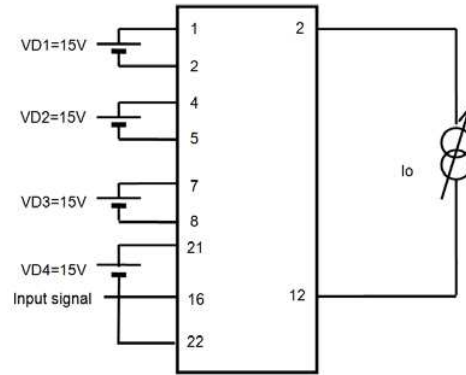
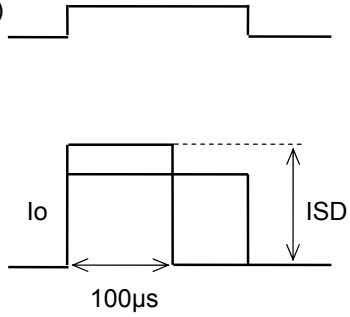


Fig. 5

■ Switching time (The circuit is a representative example of the lower side U phase.)

Input signal  
(0 to 5 V)

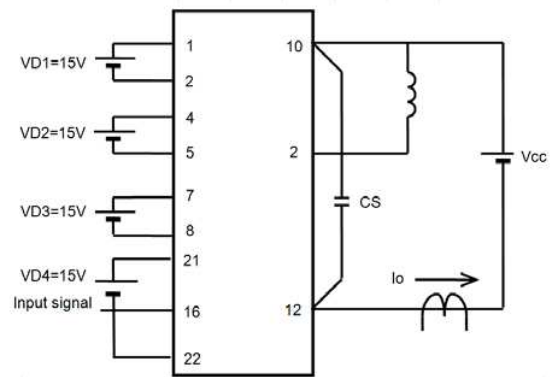
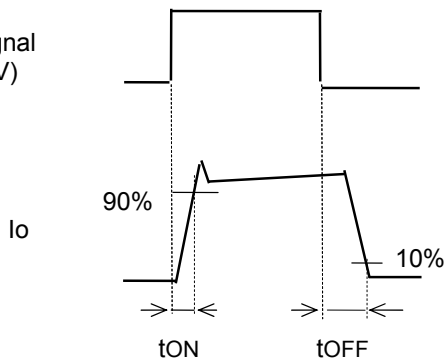


Fig. 6

Logic Timing Chart

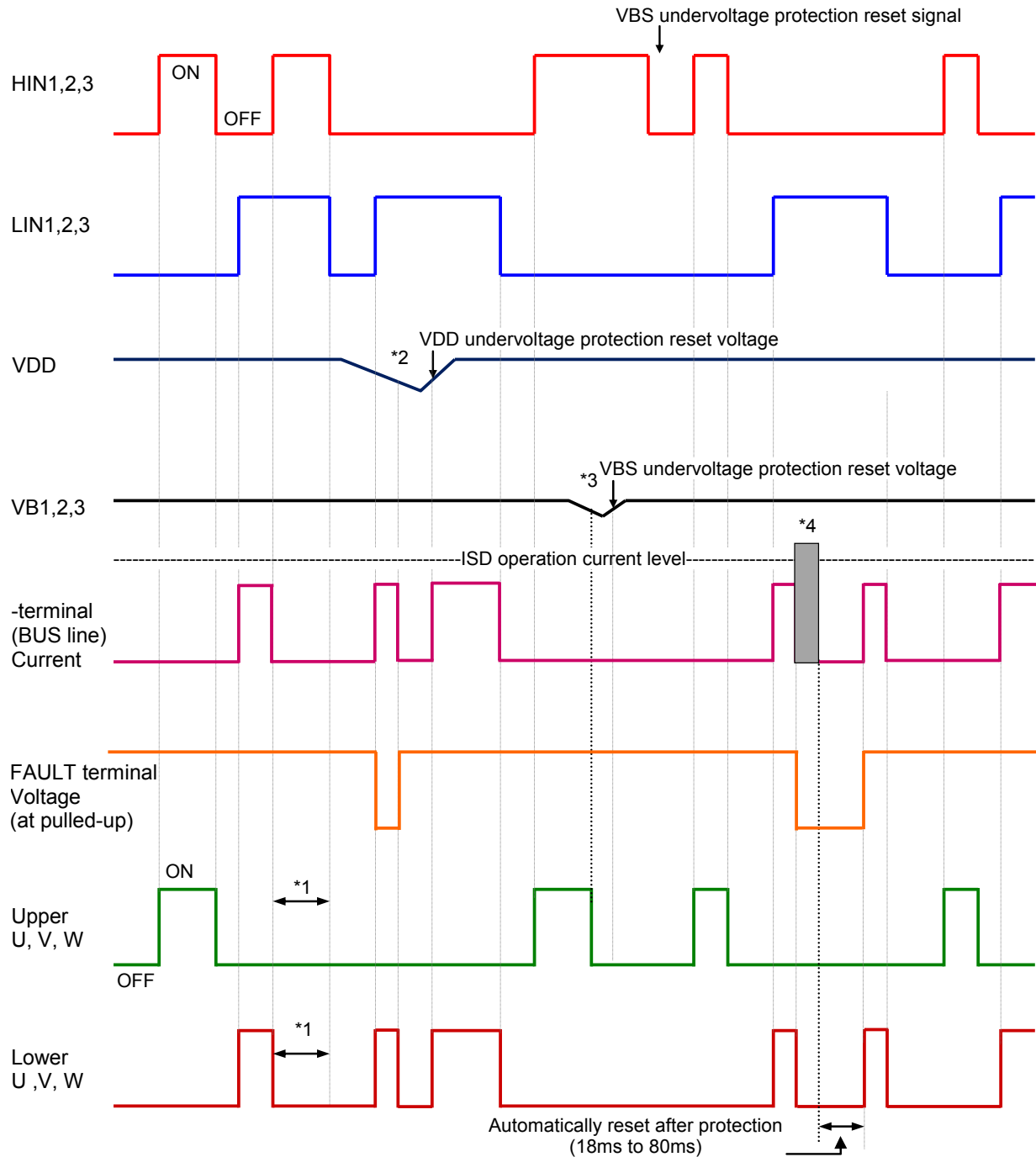


Fig. 7

Notes

- \*1 : Diagram shows the prevention of shoot-through via control logic. More dead time to account for switching delay needs to be added externally.
- \*2 : When  $V_{DD}$  decreases all gate output signals will go low and cut off all of 6 IGBT outputs. When  $V_{DD}$  rises the operation will resume immediately.
- \*3 : When the upper side gate voltage at VB1, VB2 and VB3 drops only, the corresponding upper side output is turned off. The outputs return to normal operation immediately after the upper side gate voltage rises.
- \*4 : In case of over current detection, all IGBT's are turned off and the FAULT output is asserted. Normal operation resumes in 18 to 80 ms after the over current condition is removed.



# STK581U3C2D-E

## Logic level table

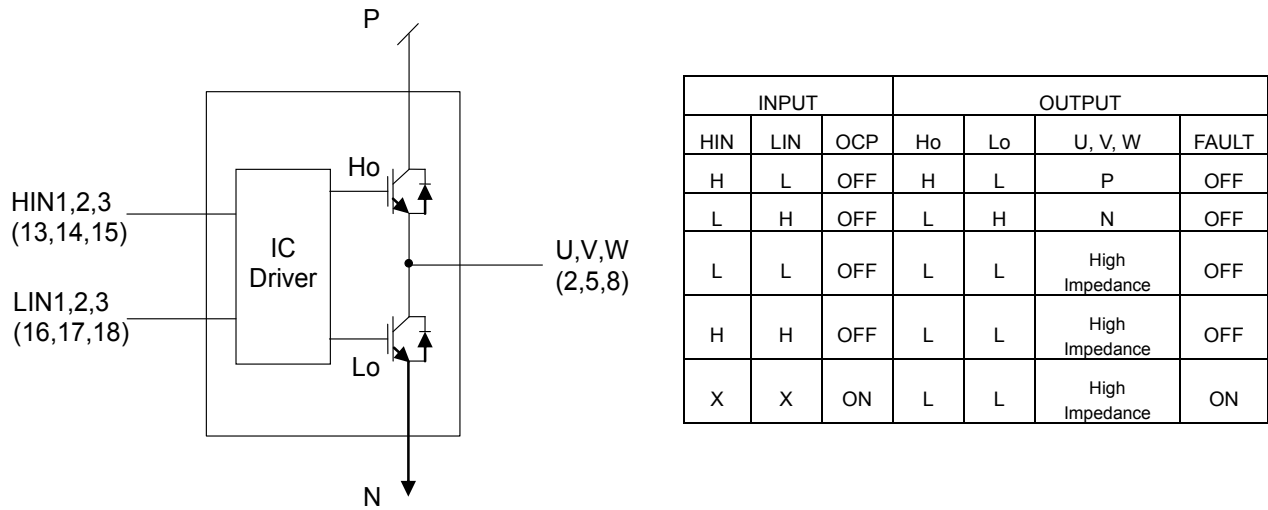


Fig. 8

## Sample Application Circuit

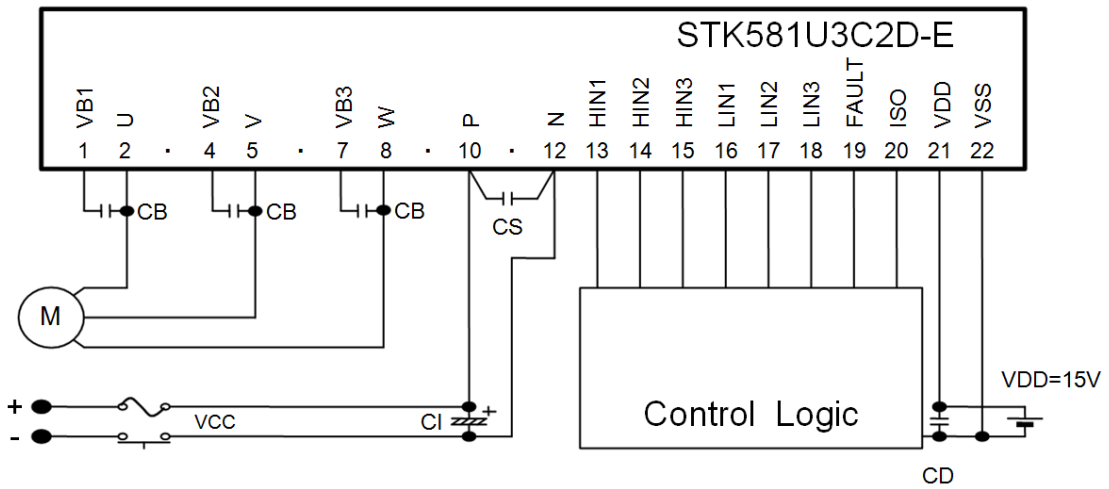


Fig. 9

**Recommended Operating Conditions** at Tc = 25°C

Item	Symbol	Conditions	min	typ	max	Unit
Supply voltage	V <sub>CC</sub>	P to N	0	280	450	V
Pre-driver supply voltage	VD1, 2, 3	VB1 to U, VB2 to V, VB3 to W	12.5	15	17.5	V
	VD4	V <sub>DD</sub> to V <sub>SS</sub> *1	13.5	15	16.5	
ON-state input voltage	V <sub>IN(ON)</sub>	HIN1, HIN2, HIN3,	3.0	-	5.0	V
OFF-state input voltage	V <sub>IN(OFF)</sub>	LIN1, LIN2, LIN3	0	-	0.3	
PWM frequency	f <sub>PWM</sub>		1	-	20	kHz
Dead time	DT	Turn-off to turn-on	2	-	-	μs
Allowable input pulse width	PWIN	ON and OFF	1	-	-	μs
Tightening torque		'M4' type screw	0.79	-	1.17	Nm

\*1 Pre-drive power supply (VD4 = 15 ±1.5 V) must have the capacity of I<sub>o</sub> = 20 mA (DC), 0.5 A (Peak).

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

**Usage Precautions**

1. This IPM includes bootstrap diode and resistors. Therefore, by adding a capacitor “CB”, a high side drive voltage is generated; each phase requires an individual bootstrap capacitor. The recommended value of CB is in the range of 1 to 47 μF, however this value needs to be verified prior to production. If selecting the capacitance more than 47 μF (±20%), connect a resistor (about 20 Ω) in series between each 3-phase upper side power supply terminals (VB1, 2, 3) and each bootstrap capacitor.  
When not using the bootstrap circuit, each upper side pre-drive power supply requires an external independent power supply.
2. It is essential that wiring length between terminals in the snubber circuit be kept as short as possible to reduce the effect of surge voltages. Recommended value of “CS” is in the range of 0.1 to 10 μF.
3. “ISO” (pin20) is terminal for current monitor. When the pull-down resistor is used, please select it more than 5.6 kΩ
4. “FAULT” (pin19) is open DRAIN output terminal (Active Low). Pull up resistor is recommended more than 5.6 kΩ.
5. Inside the IPM, a thermistor used as the temperature monitor for internal substrate is connected between V<sub>SS</sub> terminal and TH terminal, therefore, an external pull up resistor connected between the TH terminal and an external power supply should be used. The temperature monitor example application is as follows, please refer the Fig.10, and Fig.11 below.
6. Pull down resistor of 33 kΩ is provided internally at the signal input terminals. An external resistor of 2.2 k to 3.3 kΩ should be added to reduce the influence of external wiring noise.
7. The over-current protection feature is not intended to protect in exceptional fault condition. An external fuse is recommended for safety.
8. When input pulse width is less than 1.0 μs, an output may not react to the pulse. (Both ON signal and OFF signal)

This data shows the example of the application circuit, does not guarantee a design as the mass production set.

**The characteristic of thermistor**

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Resistance	R <sub>25</sub>	T <sub>c</sub> = 25°C	99	100	101	kΩ
Resistance	R <sub>100</sub>	T <sub>c</sub> = 100°C	5.12	5.38	5.66	kΩ
B-Constant (25 to 50°C)	B		4165	4250	4335	K
Temperature Range			-40	-	+125	°C

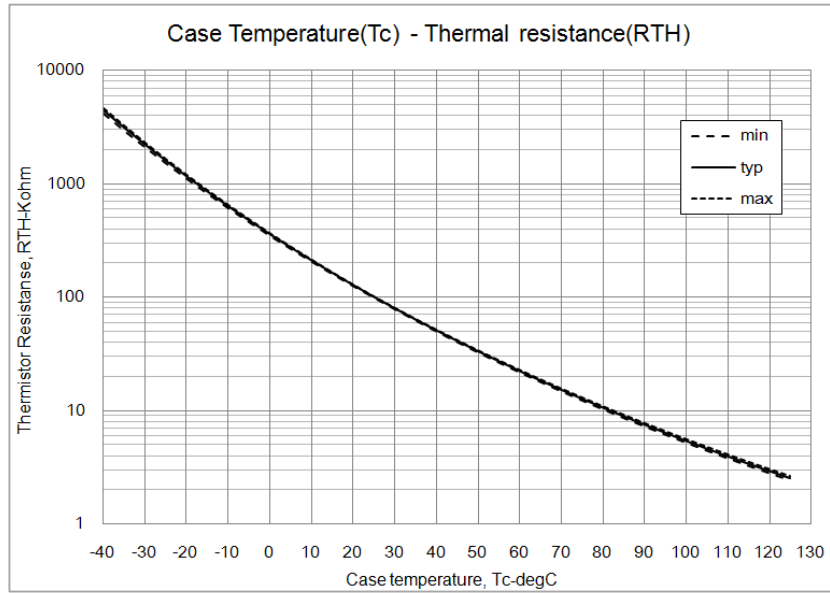


Fig. 10

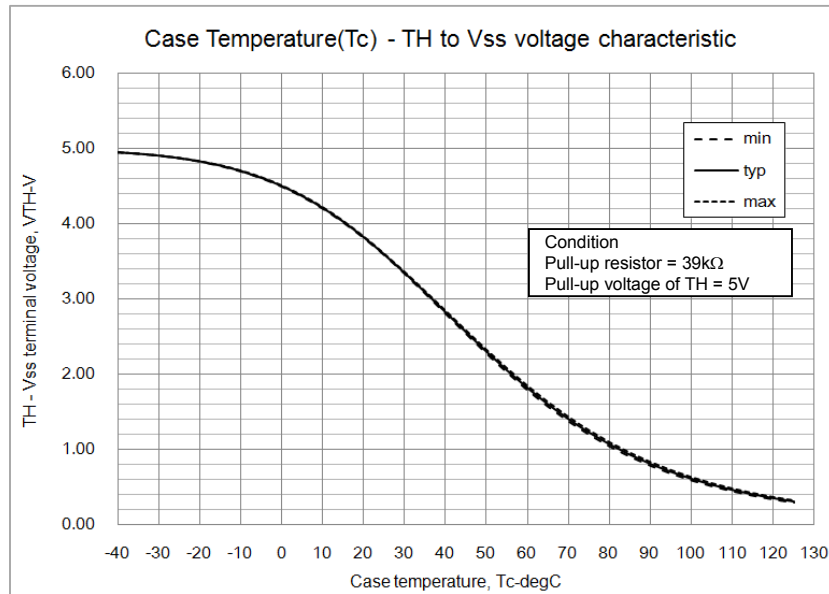


Fig. 11

**The characteristic of PWM switching frequency**

Maximum sinusoidal phase current as function of switching frequency ( $V_{BUS} = 300\text{ V}$ ,  $T_c = 100^\circ\text{C}$ )

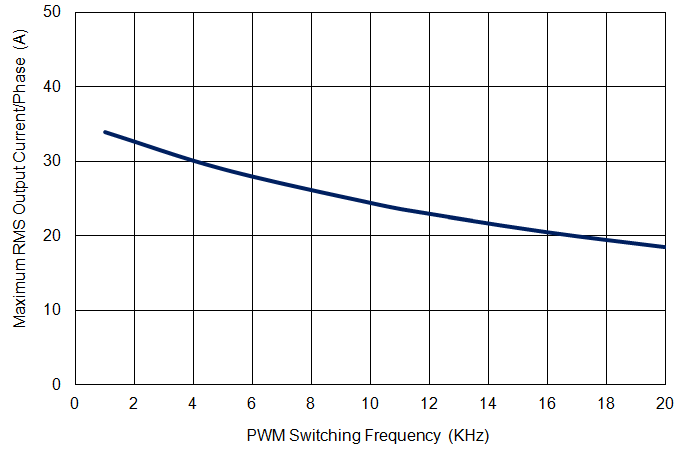


Fig.12

**Switching waveform**

IGBT Turn-on. Typical turn-on waveform @ $T_c = 100^\circ\text{C}$ ,  $V_{BUS} = 400\text{ V}$

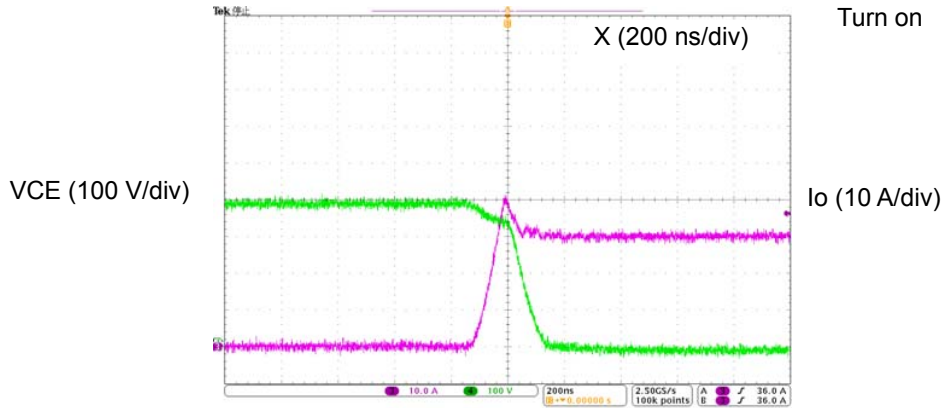


Fig. 13

IGBT Turn-off. Typical turn-off waveform @ $T_c = 100^\circ\text{C}$ ,  $V_{BUS} = 400\text{ V}$

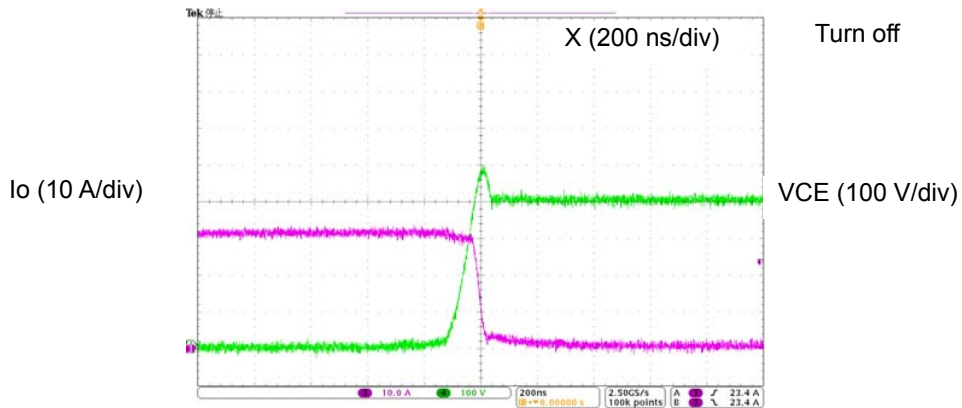


Fig. 14

**CB capacitor value calculation for bootstrap circuit**

**Calculate condition**

Item	Symbol	Value	Unit
Upper side power supply.	VBS	15	V
Total gate charge of output power IGBT at 15 V.	Qg	266	nC
Upper side power supply low voltage protection.	UVLO	12	V
Upper side power dissipation.	IDmax	400	μA
ON time required for CB voltage to fall from 15 V to UVLO	Tonmax	-	s

**Capacitance calculation formula**

CB must not be discharged below to the upper limit of the UVLO - the maximum allowable on-time (Tonmax) of the upper side is calculated as follows:

$$VBS * CB - Qg - IDmax * Tonmax = UVLO * CB$$

$$CB = (Qg + IDmax * Tonmax) / (VBS - UVLO)$$

The relationship between Tonmax and CB becomes as follows. CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μF, however, the value needs to be verified prior to production.

Tonmax-CB characteristic

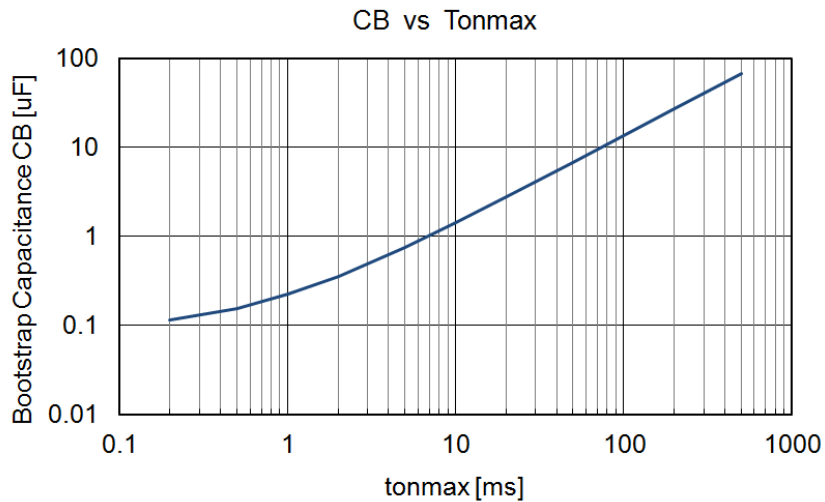


Fig 15

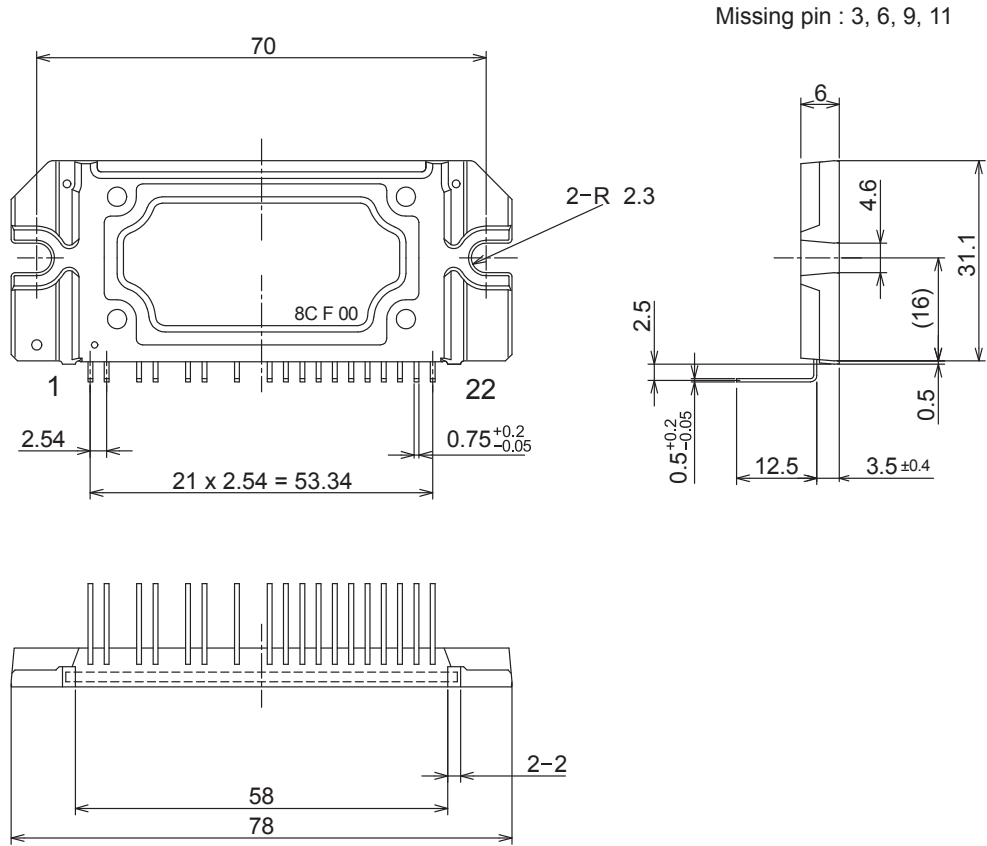
**STK581U3C2D-E**

**Package Dimensions**

unit : mm

The tolerances of length are +/- 0.5 mm unless otherwise specified.

**SIP22 70x31.1**  
CASE 127BU  
ISSUE O



**STK581U3C2D-E****ORDERING INFORMATION**

Device	Package	Shipping (Qty / Packing)
STK581U3C2D-E	SIP22 70x31.1 (Pb-Free)	7 / Tube

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