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# Intelligent Power Module (IPM) 600 V, 10 A

### Overview

This Inverter IPM includes the output stage of a 3-phase inverter, pre-drive circuits, bootstrap circuits, and protection circuits in one package.

### Function

- SIP (single in-line package) of the transfer full mold structure.
- The emitter line of the each lower phase outputs to an external terminal with the option of control using 3-phase current detection with external resistors.
- Direct input of CMOS level control signals without an insulating circuit is possible.
- Protective circuits including over current and pre-drive low voltage protection are built in.
- A single power supply drive is enabled through the use of bootstrap circuits for upper IGBT gate drives.
- Built-in dead-time for shoot-thru protection.
- Internal substrate temperature is measured with an internal pulled up thermistor.

### Certification

• UL1557 (File Number : E339285)

# Specifications

Absolute Maximum Ratings at Tc = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V <sub>CC</sub>	V+ to VRU(VRV,VRW), surge < 500 V *1	450	V
Collector-emitter voltage	V <sub>CE</sub>	V+ to U(V,W) or U(V,W) to VRU(VRV,VRW)	600	V
Output ourroat		V+, VRU,VRV,VRW, U,V,W terminal current	±10	Α
Output current	lo	V+, VRU,VRV,VRW, U,V,W terminal current at Tc = 100°C	±6	Α
Output peak current	Іор	V+, VRU, VRV, VRW, U,V,W terminal current for a pulse width of 1 ms	±20	Α
Pre-driver voltage	VD1,2,3,4	VB1 to U, VB2 to V, VB3 to W, V <sub>DD</sub> to V <sub>SS</sub> *2	20	V
Input signal voltage	VIN	HIN1, 2, 3, LIN1, 2, 3 terminals	-0.3 to 7	V
ITRIP terminal voltage	VITRIP	ITRIP terminal	V <sub>SS</sub> +5	V
Maximum power dissipation	Pd	IGBT per 1 channel	22	W
Junction temperature	Tj	IGBT,FRD	150	°C
Storage temperature	Tstg		-40 to +125	°C
Operating case temperature	Тс	IPM case temperature	-40 to +100	°C
Tightening torque		Case mounting screws *3	0.9	Nm
Withstand voltage	Vis	50 Hz sine wave AC 1 minute *4	2000	VRMS

Reference voltage is "VSS" terminal voltage unless otherwise specified.

\*1 : Surge voltage developed by the switching operation due to the wiring inductance between V+ and VRU(VRV,VRW) terminals.

- \*2 : VD1 = VB1 to U, VD2 = VB2 to V, VD3 = VB3 to W, VD4 = V<sub>DD</sub> to  $V_{SS}$  terminal voltage.
- \*3 : Flatness of the heat-sink should be less than –50  $\mu$ m to +100  $\mu$ m.
- \*4 : Test conditions : AC 2500 V, 1 second.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

# **ORDERING INFORMATION**

See detailed ordering and shipping information on page 15 of this data sheet.



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ectrical Characteristics at Tc = 25°C, VD1, VD2, VD3, VD4 = 15 V
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Deveryortex	Oursels al		Test	Ratings			
Parameter	Symbol Conditions ci		circuit	min	typ	max	Unit
Power output section							
Collector-emitter cut-off current	ICE	V <sub>CE</sub> = 600 V		-	-	0.1	mA
Bootstrap diode reverse current	IR(BD)	Fig.		-	-	0.1	mA
Collector to emitter saturation		lo = 10 A, Tj = 25°C		-	1.4	2.3	
voltage	V <sub>CE</sub> (SAT)	lo = 5 A, Tj = 100°C	Fig.2	-	1.3	-	V
Diada famorado alta na	VF	lo = 10 A, Tj = 25°C		-	1.3	2.2	v
Diode forward voltage	VF	lo = 5 A, Tj = 100°C	Fig.3	-	1.2	-	
Bootstrap diode forward voltage	VF(BD)	IF = 0.1 A	-	-	2.0	-	V
	R <sub>BC</sub>	Resistor value for common boot charge line	-	-	2	-	
Bootstrap circuit resistance	R <sub>BS</sub>	Resister values for separate boot charge lines	-	-	10	-	Ω
lunction to ence thermal registeres	θj-c(T)	IGBT		-	4.5	5.5	
Junction to case thermal resistance	θj-c(D)	FRD	-	-	5.5	6.5	°C/W
Thermal resistance case to sink Rth(c-s)		1W/mK thermal conductivity *1		-	0.1	-	
Protection section							
FAULT clearance delay time	FLTCLR	Form time fault condition clears	-	6	9	12	ms
Quitabing time	t ON	lo = 10 A		-	0.48	-	
Switching time	t OFF	Inductive load	Fig.5	-	0.54	-	μs

Reference voltage is "VSS" terminal voltage unless otherwise specified.

\*1 : At 100  $\mu m$  thickness of the thermal grease.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

# **Electrical Characteristics Driver Function** at Tc = 25°C

Deremeter	Cumbal	Test	Ratings			Unit
Parameter	Symbol	circuit	min	typ	max	Onit
Supply section						
	V <sub>DDUV+</sub>		10.5	11.1	11.7	v
$V_{DD}$ and $V_{BS}$ supply undervoltage protection reset	V <sub>BSUV+</sub>	-	10.5	11.1	11.7	v
	V <sub>DDUV-</sub>		10.2	10.0	11 5	v
$V_{DD}$ and $V_{BS}$ supply undervoltage protection set	V <sub>BSUV-</sub>	-	10.3	10.9	11.5	v
	V <sub>DDUVH</sub>		0.14	0.2	-	v
$V_{DD}$ and $V_{BS}$ supply undervoltage hysteresis	V <sub>BSUVH</sub>	-	0.14			v
Quiescent V <sub>DD</sub> supply current		Fig. 4	-	2.0	4.0	mA
Quiescent V <sub>BS</sub> supply current	I <sub>QBS</sub>	Fig.4	-	0.08	0.4	mA
Input section						
Logic low input voltage	V <sub>INL</sub>	-	-	-	0.8	V
Logic high input voltage	V <sub>INH</sub>	-	2.5		-	V
Logic 0 input leakage current	I <sub>IN+</sub>	-	76	118	160	μA
Logic 1 input leakage current	I <sub>IN-</sub>	-	97	150	203	μA
ITRIP threshold voltage (OUT = LO or OUT = HI)	VITRIP	-	3.67	4.17	4.67	V
Dynamic section						
Dead time (Internal dead time injected by driver)	DT	-	220	300	380	ns
ITRIP to shutdown propagation delay	t <sub>ITRIP</sub>	-	1.0	1.2	1.4	μs
ITRIP blanking time	t <sub>ITRPBL</sub>	-	-	0.9	-	μs

Reference voltage is "VSS" terminal voltage unless otherwise specified.

# Switching Characteristics at $Tc = 25^{\circ}C$ , VD1, VD2, VD3, VD4 = 15 V

Demonster	Querra ha a l	O and difference		Ratings		1.1 34	
Parameter	Symbol	Symbol Conditions		typ	max	Unit	
Turn-on switching loss	Eon	I <sub>C</sub> = 5 A, V+ = 400 V	-	195	-	μJ	
Turn-off switching loss	Eoff	V <sub>DD</sub> = 15 V, L = 3.9 mH,	-	122	-	μJ	
Total switching loss	Etot	Tc = 25°C	-	317	-	μJ	
Turn-on switching loss	Eon	I <sub>C</sub> = 5 A, V+ = 400 V,	-	224	-	μJ	
Turn-off switching loss	Eoff	V <sub>DD</sub> = 15 V, L = 3.9 mH,	-	186	-	μJ	
Total switching loss	Etot	Tc = 100°C	-	410	-	μJ	
Diode reverse recovery time	trr	I <sub>F</sub> = 5A, V+ = 400 V, V <sub>DD</sub> = 15 V, L = 3.9 mH, Tc = 100°C	-	70	-	ns	
Reverse bias safe operating area	RBSOA	lo = 20 A, V <sub>CE</sub> = 450 V	FULL SQUARE		-		
Short circuit safe operating area	SCSOA	V <sub>CE</sub> = 400 V	4	-	-	μs	

 $V_{DD} = V_{B1} = V_{B2} = V_{B3} = 15 \text{ V}, V_{SS} = V_{S1} = V_{S2} = V_{S3} = 0 \text{ V}$ , outputs loaded with 1 nF, all voltage are referenced to  $V_{SS}$ ; unless otherwise noted.

Internal NTC-Thermistor	Characteristics
-------------------------	-----------------

Parameter		Conditions	Тур.	Unit
R25	Resistance	Tc = 25°C	100 ±3%	kΩ
R125	Resistance	Tc = 125°C	2.522 ±3%	kΩ
B B-Constant (25 to 50°C)		$R_2 = R_1 e^{[B(1/T2-1/T1)]}$	4250 ±1%	к
Temperature range		-	-40 to +125	°C
Typ. Dissipation constant		Tc=25°C	1	mW/°C

### Notes

1. The pre-drive power supply low voltage protection has approximately 200 mV of hysteresis and operates as follows.

Upper side : The gate is turned off and will return to regular operation when recovering to the normal voltage, but the latch will continue till the input signal will turn 'high'.

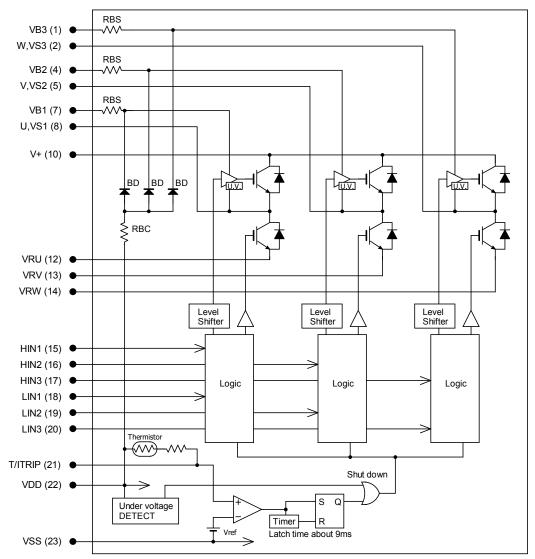
Lower side : The gate is turned off and will automatically reset when recovering to normal voltage. It does not depend on input signal voltage.

2. When assembling the IPM on the heat sink the tightening torque range is 0.6 Nm to 0.9 Nm.

3. The pre-drive low voltage protection protects the device when the pre-drive supply voltage falls due to an operating malfunction.

4. When use the over-current protection with external shunt resistor, please set the current protection level to be equal to or less than the rating of output peak current (lop).

# **Equivalent Block Diagram**



# **Module Pin-Out Description**

Pin	Name	Description
1	VB3	High Side Floating Supply Voltage 3
2	W, VS3	Output 3, High Side Floating Supply Offset Voltage 3
3	-	Without pin
4	VB2	High Side Floating Supply Voltage 2
5	V, VS2	Output 2, High Side Floating Supply Offset Voltage 2
6	-	Without pin
7	VB1	High Side Floating Supply Voltage 1
8	U, VS1	Output 1, High Side Floating Supply Offset Voltage 1
9	-	Without pin
10	V+	Positive Bus Input Voltage
11	-	Without pin
-	-	-

Pin	Name	Description
12	VRU	Low Side Emitter Connection – Phase 1
13	VRV	Low Side Emitter Connection – Phase 2
14	VRW	Low Side Emitter Connection – Phase 3
15	HIN1	Logic Input High Side Gate Driver – Phase 1
16	HIN2	Logic Input High Side Gate Driver – Phase 2
17	HIN3	Logic Input High Side Gate Driver – Phase 3
18	LIN1	Logic Input Low Side Gate Driver – Phase 1
19	LIN2	Logic Input Low Side Gate Driver – Phase 2
20	LIN3	Logic Input Low Side Gate Driver – Phase 3
21	T/Itrip	Temperature Monitor and Shut-down Pin
22	VDD	+15 V Main Supply
23	VSS	Negative Main Supply

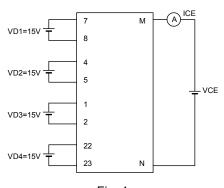
# **Test Circuit**

The tested phase U+ shows the upper side of the U phase and U- shows the lower side of the U phase.

■ I<sub>CE</sub> / IR(BD)

	U+	V+	W+	U-	V-	W-
М	10	10	10	8	5	2
Ν	8	5	2	12	13	14

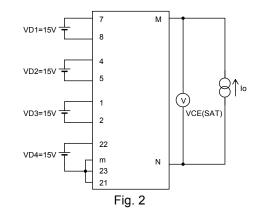
	U(BD)	V(BD)	W(BD)
М	7	4	1
Ν	23	23	23





■ V<sub>CE</sub>(SAT) (test by pulse)

	U+	V+	W+	U-	V-	W-
М	10	10	10	8	5	2
Ν	8	5	2	12	13	14
m	15	16	17	18	19	20

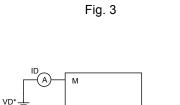


V<sub>F</sub> (test by pulse)

	U+	V+	W+	U-	V-	W-
М	10	10	10	8	5	2
Ν	8	5	2	12	13	14

∎ ID

		VD1	VD2	VD3	VD4
ĺ	М	7	4	1	22
	Ν	8	5	2	23



М

Ν

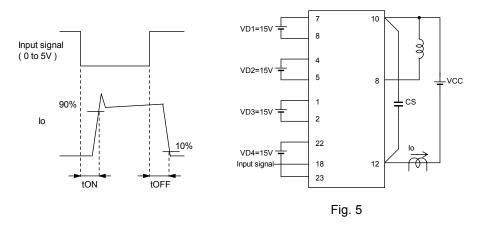
V VF

lo

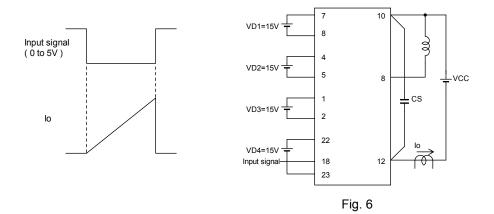
Fig. 4

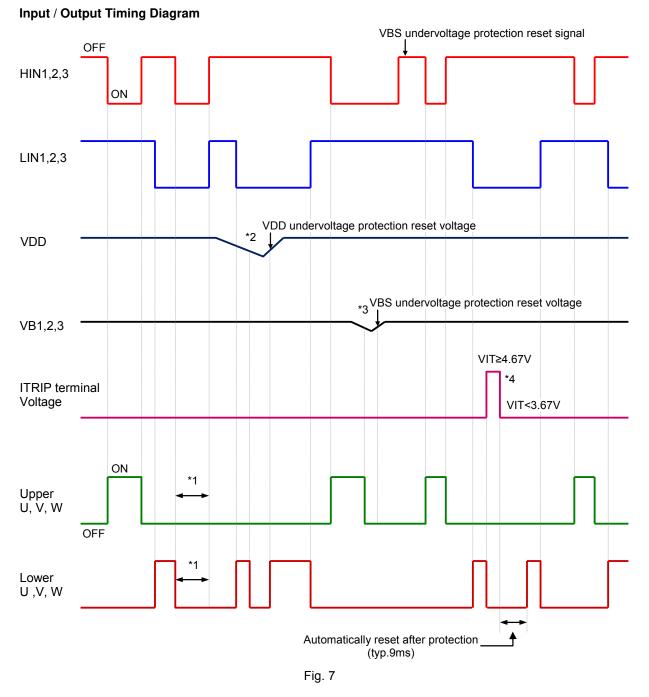
Ν

Switching time (The circuit is a representative example of the lower side U phase.)



RB-SOA (The circuit is a representative example of the lower side U phase.)

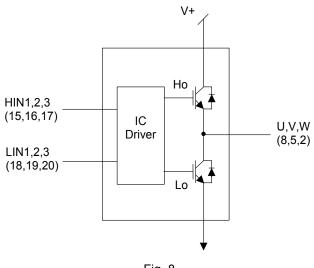




### Notes

- \*1 : Shows the prevention of shoot-thru via control logic, however, more dead time must be added to account for switching delay externally.
- \*2 : When V<sub>DD</sub> decreases all gate output signals will go low and cut off all 6 IGBT outputs. When V<sub>DD</sub> rises the operation will resume immediately.
- \*3 : When the upper side voltage at VB1, VB2 and VB3 drops only the corresponding upper side output is turned off. The outputs return to normal operation immediately after the upper side gate voltage rises.
- \*4 : When VITRIP exceeds threshold all IGBT's are turned off and normal operation resumes 9 ms (typ) after over current condition is removed.

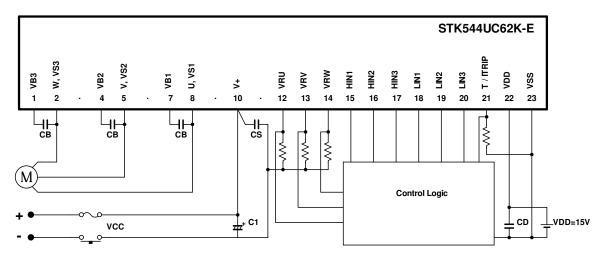
Logic level table



Itrip	HIN1,2,3	LIN1,2,3	U,V,W
0	0	1	V+
0	1	0	0
0	1	1	х
1	Х	х	х

Fig. 8

# Sample Application Circuit



### Recommended Operating Conditions at Tc = 25°C

Item	Symbol	Conditions	Ratings			Unit
nem	Symbol	Conditions	min	typ	max	Unit
Supply voltage	V <sub>CC</sub>	Between V+ to VRU(VRV,VRW)	0	280	450	V
Des deiver surgebung berge	VD1,2,3	Between VB1 to U,VB2 to V,VB3 to W	12.5	15	17.5	
Pre-driver supply voltage	VD4	Between V <sub>DD</sub> to V <sub>SS</sub> *1	13.5	15	16.5	V
ON-state input voltage	VIN(ON)	HIN1,HIN2,HIN3,LIN1,LIN2,LIN3	0	-	0.3	V
OFF-state input voltage	VIN(OFF)	terminal	3.0	-	5.0	v
PWM frequency	fPWM	-	1	-	20	kHz
Dead time	DT	Turn-off to turn-on (External)	0.5	-	-	μs
Mounting torque	-	'M3' type screw	0.6	-	0.9	Nm

\*1 : Pre-drive power supply (VD4 = 15 ±1.5 V) must be have the capacity of Io = 20 mA (DC), 0.5 A (Peak).

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

### Usage Precaution

- This IPM includes internal bootstrap diode and resistor. By adding a capacitor CB, a single high side drive voltage is generated; each phase requires an individual bootstrap capacitor. The recommended value of CB is in the range of 1 to 47 μF, however, this value needs to be verified prior to production. When not using the bootstrap circuit, each upper side pre-drive power supply requires an external independent power supply.
- 2. It is essential thet wirning length between terminals in the snubber circuit be kept as short as possible to reduce the effect of surge voltages. Recommended value of CS is in the range of 0.1 to 10 μF.
- 3. VRU, VRV, and VRW terminals are direct outputs of the emitter line of the each lower phase IGBT and can be used to monitor each phase's or collective current with external resistors. IF current is not monitored, each resistor should be short-circuites.
- 4. Disconnection of terminals U, V, or W during normal motor operation will cause damage to IPM, use caution with this connection.
- 5. Zner diode with 10 V is connected with the inside of the signal input terminal. When inputting voltage which exceeds 5 V, connect resistor between the side of the power and the signal input terminal, for the input current of the signal input terminal become equal to or less than 0.5 mA. This resistor is effective with the noise absorption of the signal terminal, too.
- 6. A fuse on  $V_{CC}$  is recommended.
- 7. Inside the IPM, a thermistor used as the temperature monitor is connected between V<sub>DD</sub> terminal and T/ITRIP terminal, therefore, an external pull down resistor connected between the T/ITRIP terminal and V<sub>SS</sub> terminal should be used. The temperature monitor example application is as follows, please refer the Fig.10, Fig.11, and Fig.12 below.
- 8. All data shown implements an example of the application circuit but does not guarantee a design for the mass production.

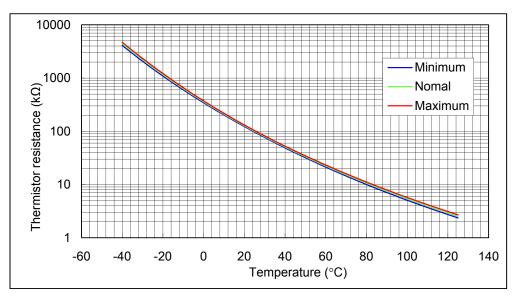


Fig. 10 Variation of thermistor resistance with temperature

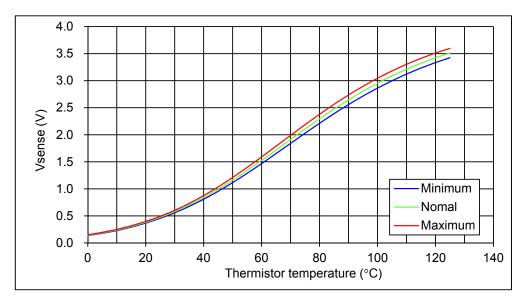


Fig. 11 Variation of temperature sense voltage with thermistor temperature by using external bias resistance of 4.3 k $\Omega \pm 1\%$  and V<sub>DD</sub> = 15 V

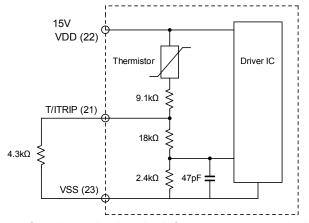


Fig. 12 Sample application circuit for temperature monitoring

# The characteristic of PWM switching frequency

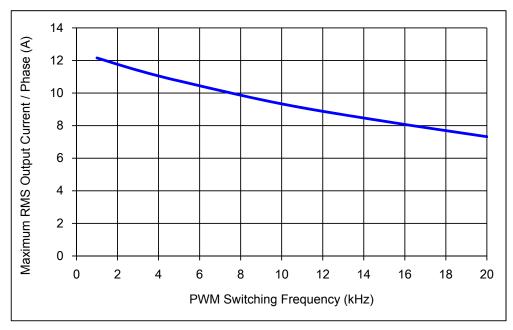


Fig. 13 Maximum sinusoidal phase current as function of switching frequency at Tc = 100°C, V\_{CC} = 400 V

# Switching waveform

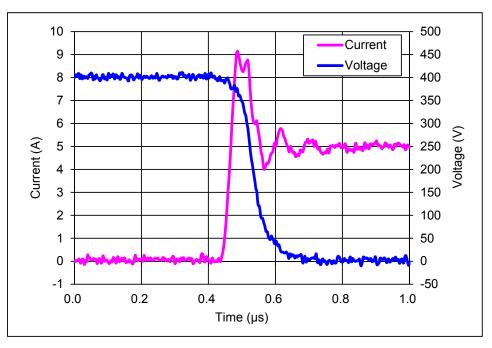


Fig. 14 IGBT Turn-on. Typical turn-on waveform at Tc = 100°C, V<sub>CC</sub> = 400 V

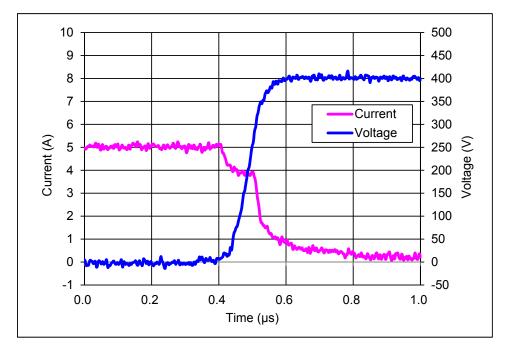


Fig. 15 IGBT Turn-off. Typical turn-off waveform at Tc = 100°C, V<sub>CC</sub> = 400 V

# CB capacitor value calculation for bootstrap circuit

# **Calculate conditions**

Parameter	Symbol	Value	Unit
Upper side power supply	VBS	15	V
Total gate charge of output power IGBT at 15 V	QG	89	nC
Upper limit power supply low voltage protection	UVLO	12	V
Upper side power dissipation	IDMAX	95	μA
ON time required for CB voltage to fall from 15 V to UVLO	TONMAX	-	S

# Capacitance calculation formula

Thus, the following formula are true

 $\mathsf{VBS}\times\mathsf{CB}-\mathsf{QG}-\mathsf{IDMAX}\times\mathsf{TONMAX}=\mathsf{UVLO}\times\mathsf{CB}$ 

therefore,

 $CB = (QG + IDMAX \times TONMAX) / (VBS - UVLO)$ 

The relationship between TONMAX and CB becomes as follows. CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 µF, however, this value needs to be verified prior to production.

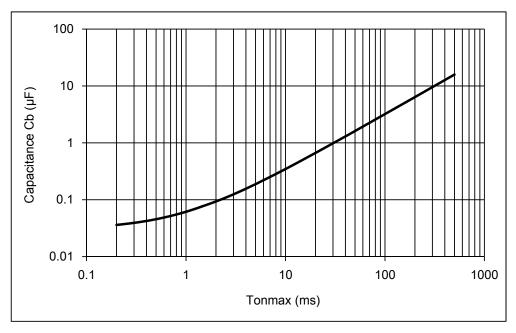


Fig. 16 TONMAX vs CB characteristic

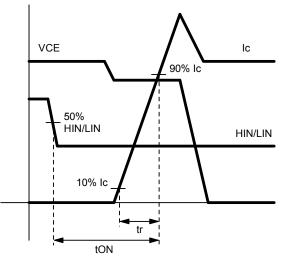


Fig. 17 Input to output propagation turn-on delay time

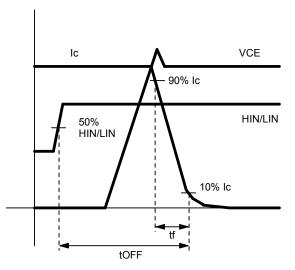


Fig. 18 Input to output propagation turn-off delay time

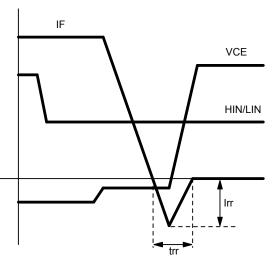
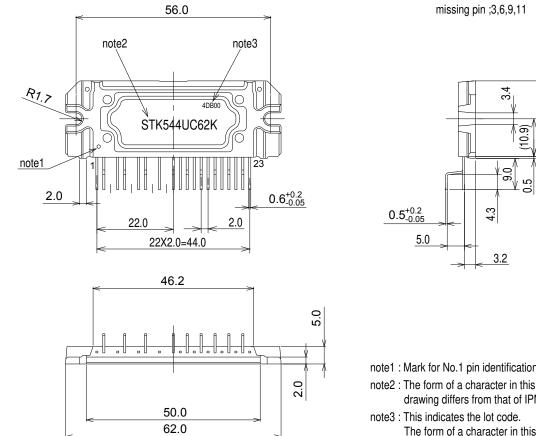
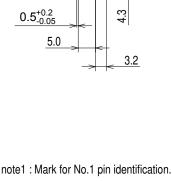


Fig. 19 Diode reverse recovery



ω

5



drawing differs from that of IPM.

The form of a character in this drawing differs from that of IPM.

The tolerances of length are +/- 0.5 mm unless otherwise specified.

# **ORDERING INFORMATION**

Device	Package	Shipping (Qty / Packing)
STK544UC62K-E	SIP23 (Pb-Free)	8 / Tube

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