

# LMV7271/LMV7275/LMV7272 Single & Dual, 1.8V Low Power Comparators with Rail-to-Rail Input

Check for Samples: [LMV7271](#)

## FEATURES

- ( $V_S = 1.8V$ ,  $T_A = 25^\circ C$ , Typical values unless specified).
- Single or Dual Supplies
- Ultra low supply current  $9\mu A$  per channel
- Low input bias current  $10nA$
- Low input offset current  $200pA$
- Low guaranteed  $V_{OS}$   $4mV$
- Propagation delay  $880ns$  ( $20mV$  overdrive)
- Input common mode voltage range  $0.1V$  beyond rails
- LMV7272 is available in DSBGA package

## APPLICATIONS

- Mobile communications
- Laptops and PDA's
- Battery powered electronics
- General purpose low voltage applications

## DESCRIPTION

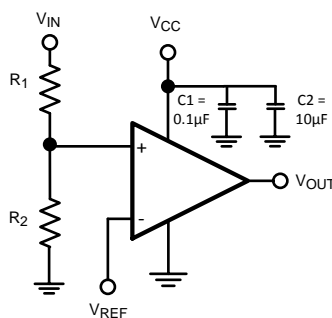
The LMV727X are rail-to-rail input low power comparators, which are characterized at supply voltage  $1.8V$ ,  $2.7V$  and  $5.0V$ . They consume only  $9\mu A$  supply current per channel while achieving a  $800ns$  propagation delay.

The LMV7271/LMV7275 (single) are available in SC70 and SOT-23 packages. The LMV7272 (dual) is available in DSBGA package. With these tiny packages, the PC board area can be significantly reduced. They are ideal for low voltage, low power and space critical designs.

The LMV7271/LMV7272 both feature a push-pull output stage which allows operation with minimum power consumption when driving a load. The LMV7275 features an open drain output stage that allows for wired-OR configurations. The open drain output also offers the advantage of allowing the output to be pulled to any voltage up to  $5.5V$ , regardless of the supply voltage of the LMV7275.

The LMV727X are built with Texas Instruments' advance submicron silicon-gate BiCMOS process. They all have bipolar inputs for improved noise performance and CMOS outputs for rail-to-rail output swing.

## Typical Circuit


**Figure 1. Threshold Detector**

Part Number	Single/Dual	Package	Output
LMV7271	Single	SC70, SOT-23	Push/Pull
LMV7272	Dual	DSBGA	Push/Pull
LMV7275	Single	SC70, SOT-23	Open Drain



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### Absolute Maximum Ratings <sup>(1)(2)</sup>

ESD Tolerance	2KV <sup>(3)</sup>
	200V <sup>(4)</sup>
V <sub>IN</sub> Differential	±Supply Voltage
Supply Voltage (V <sup>+</sup> - V <sup>-</sup> )	6V
Voltage at Input/Output pins	V <sup>+</sup> +0.1V, V <sup>-</sup> -0.1V
Soldering Information	
Infrared or Convection (20 sec.)	235°C
Wave Soldering (10 sec.)	260°C
Storage Temperature Range	-65°C to +150°C
Junction Temperature <sup>(5)</sup>	+150°C

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. *Operating Ratings* indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.
- (2) **If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office / Distributors for availability and specifications.**
- (3) Human body model, 1.5kΩ in series with 100pF.
- (4) Machine Model, 0Ω in series with 200pF.
- (5) Typical values represent the most likely parametric norm.

### Operating Ratings <sup>(1)</sup>

Supply Voltage Range	1.8V to 5.5V
Temperature Range <sup>(2)</sup>	-40°C to +85°C
Package Thermal Resistance <sup>(2)</sup>	
SOT-23	325°C/W
SC70	265°C/W
8-Bump DSBGA	220°C/W

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. *Operating Ratings* indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.
- (2) The maximum power dissipation is a function of T<sub>J(MAX)</sub>, θ<sub>JA</sub>, and T<sub>A</sub>. The maximum allowable power dissipation at any ambient temperature is P<sub>D</sub> = (T<sub>J(MAX)</sub> - T<sub>A</sub>)/θ<sub>JA</sub>. All numbers apply for packages soldered directly into a PC board.

### 1.8V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for T<sub>J</sub> = 25°C, V<sup>+</sup> = 1.8V, V<sup>-</sup> = 0V. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min <sup>(1)</sup>	Typ <sup>(2)</sup>	Max <sup>(1)</sup>	Units
V <sub>OS</sub>	Input Offset Voltage			0.3	<b>4</b> <b>6</b>	mV
TC V <sub>OS</sub>	Input Offset Temperature Drift	V <sub>CM</sub> = 0.9V <sup>(3)</sup>		20		μV/°C
I <sub>B</sub>	Input Bias Current			10		nA
I <sub>OS</sub>	Input Offset Current			200		pA
I <sub>S</sub>	Supply Current	LMV7271/LMV7275		9	<b>12</b> <b>14</b>	μA
		LMV7272		18	<b>25</b> <b>28</b>	μA

- (1) All limits are guaranteed by testing or statistical analysis.
- (2) Typical values represent the most likely parametric norm.
- (3) Offset Voltage average drift determined by dividing the change in V<sub>OS</sub> at temperature extremes into the total temperature change.

## 1.8V Electrical Characteristics (continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 1.8\text{V}$ ,  $V^- = 0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min (1)	Typ (2)	Max (1)	Units
$I_{SC}$	Output Short Circuit Current	Sourcing, $V_O = 0.9\text{V}$ (LMV7271/LMV7272 only)	3.5	6		mA
		Sinking, $V_O = 0.9\text{V}$	4	6		
$V_{OH}$	Output Voltage High (LMV7271/LMV7272 only)	$I_O = 0.5\text{mA}$	1.7	1.74		V
		$I_O = 1.5\text{mA}$	1.47	1.63		
$V_{OL}$	Output Voltage Low	$I_O = -0.5\text{mA}$		52	100	mV
		$I_O = -1.5\text{mA}$		166	220	
$V_{CM}$	Input Common Mode Voltage Range	CMRR > 45 dB			1.9	V
			-0.1			V
CMRR	Common Mode Rejection Ratio	$0 < V_{CM} < 1.8\text{V}$	46	78		dB
PSRR	Power Supply Rejection Ratio	$V^+ = 1.8\text{V}$ to 5V	55	80		dB
$I_{LEAKAGE}$	Output Leakage Current	$V_O = 1.8\text{V}$ (LMV7275 only)		2		pA

## 1.8V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 1.8\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{CM} = 0.5\text{V}$ ,  $V_O = V^+/2$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min (1)	Typ (2)	Max (1)	Units
$t_{PHL}$	Propagation Delay (High to Low)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		880		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		570		ns
$t_{PLH}$	Propagation Delay (Low to High)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		1100		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		800		ns

(1) Machine Model, 0 $\Omega$  in series with 200pF.

(2) All limits are guaranteed by testing or statistical analysis.

## 2.7V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min (1)	Typ (2)	Max (1)	Units
$V_{OS}$	Input Offset Voltage			0.3	4 <b>6</b>	mV
TC $V_{OS}$	Input Offset Temperature Drift	$V_{CM} = 1.35\text{V}$ <sup>(3)</sup>		20		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current			10		nA
$I_{OS}$	Input offset Current			200		pA
$I_S$	Supply Current	LMV7271/LMV7275		9	13 <b>15</b>	$\mu\text{A}$
		LMV7272		18	25 <b>28</b>	$\mu\text{A}$
$I_{SC}$	Output Short Circuit Current	Sourcing, $V_O = 1.35\text{V}$ (LMV7271/LMV7272 only)	10	15		mA
		Sinking, $V_O = 1.35\text{V}$	10	15		

(1) Machine Model, 0 $\Omega$  in series with 200pF.

(2) All limits are guaranteed by testing or statistical analysis.

(3) Offset Voltage average drift determined by dividing the change in  $V_{OS}$  at temperature extremes into the total temperature change.

## 2.7V Electrical Characteristics (continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min (1)	Typ (2)	Max (1)	Units
$V_{OH}$	Output Voltage High (LMV7271/LMV7272 only)	$I_O = 0.5\text{mA}$	2.63	2.66		V
		$I_O = 2.0\text{mA}$	2.48	2.55		
$V_{OL}$	Output Voltage Low	$I_O = -0.5\text{mA}$		50	70	mV
		$I_O = -2\text{mA}$		155	220	
$V_{CM}$	Input Common Voltage Range	CMRR > 45dB			2.8	V
			-0.1			V
CMRR	Common Mode Rejection Ratio	$0 < V_{CM} < 2.7\text{V}$	46	78		dB
PSRR	Power Supply Rejection Ratio	$V^+ = 1.8\text{V}$ to $5\text{V}$	55	80		dB
$I_{LEAKAGE}$	Output Leakage Current	$V_O = 2.7\text{V}$ (LMV7275 only)		2		pA

## 2.7V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 2.7\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{CM} = 0.5\text{V}$ ,  $V_O = V^+/2$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min (1)	Typ (2)	Max (1)	Units
$t_{PHL}$	Propagation Delay (High to Low)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		1200		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		810		ns
$t_{PLH}$	Propagation Delay (Low to High)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		1300		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		860		ns

(1) Machine Model, 0 $\Omega$  in series with 200pF.

(2) All limits are guaranteed by testing or statistical analysis.

## 5.0V Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min (1)	Typ (2)	Max (1)	Units
$V_{OS}$	Input Offset Voltage			0.3	<b>4</b> <b>6</b>	mV
TC $V_{OS}$	Input Offset Temperature Drift	$V_{CM} = 2.5\text{V}$ (3)		20		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current			10		nA
$I_{OS}$	Input Offset Current			200		pA
$I_S$	Supply Current	LMV7271/LMV7275		10	<b>14</b> <b>16</b>	$\mu\text{A}$
		LMV7272		20	<b>27</b> <b>30</b>	$\mu\text{A}$
$I_{SC}$	Output Short Circuit Current	Sourcing, $V_O = 2.5\text{V}$ (LMV7271/LMV7272 only)	18	34		mA
		Sinking, $V_O = 2.5\text{V}$	18	34		
$V_{OH}$	Output Voltage High (LMV7271/LMV7272 only)	$I_O = 0.5\text{mA}$	4.93	4.96		V
		$I_O = 4.0\text{mA}$	4.675	4.77		

(1) Machine Model, 0 $\Omega$  in series with 200pF.

(2) All limits are guaranteed by testing or statistical analysis.

(3) Offset Voltage average drift determined by dividing the change in  $V_{OS}$  at temperature extremes into the total temperature change.

### 5.0V Electrical Characteristics (continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Min (1)	Typ (2)	Max (1)	Units
$V_{OL}$	Output Voltage Low	$I_O = -0.5\text{mA}$		27	70	mV
		$I_O = -4.0\text{mA}$		225	315	
$V_{CM}$	Input Common Voltage Range	CMRR > 45dB			5.1	V
			-0.1			
CMRR	Common Mode Rejection Ratio	$0 < V_{CM} < 5.0\text{V}$	46	78		dB
PRSS	Power Supply Rejection Ratio	$V^+ = 1.8\text{V to } 5\text{V}$	55	80		dB
$I_{LEAKAGE}$	Output Leakage Current	$V_O = 5\text{V}$ (LMV7275 only)		2		pA

### 5.0V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ ,  $V^+ = 5.0\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{CM} = 0.5\text{V}$ ,  $V_O = V^+/2$  and  $R_L > 1\text{M}\Omega$  to  $V^-$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Min (1)	Typ (2)	Max (1)	Units
$t_{PHL}$	Propagation Delay (High to Low)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		2100		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		1380		ns
$t_{PLH}$	Propagation Delay (Low to High)	Input Overdrive = 20mV Load = 50pF//5k $\Omega$		1800		ns
		Input Overdrive = 50mV Load = 50pF//5k $\Omega$		1100		ns

- (1) Machine Model, 0 $\Omega$  in series with 200pF.
- (2) All limits are guaranteed by testing or statistical analysis.

### CONNECTION DIAGRAMS

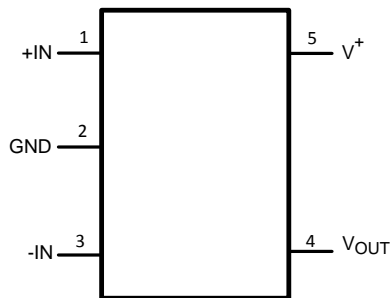


Figure 2. 5-Pin SOT-23/SC70 (LMV7271/LMV7275) (Top View)

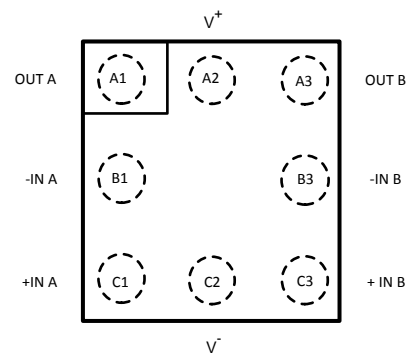


Figure 3. 8-Bump DSBGA (LMV7272) (bump side down)

**TYPICAL PERFORMANCE CHARACTERISTICS**

( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).

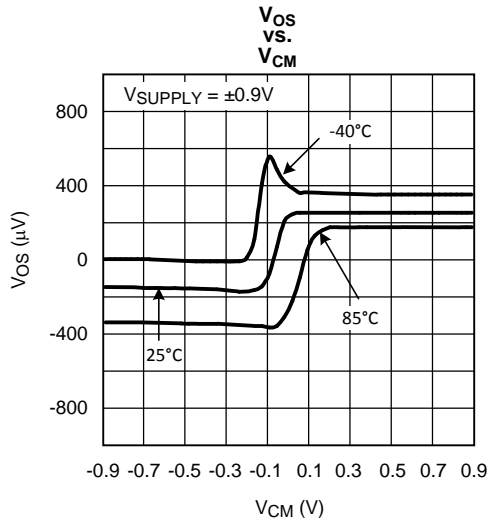


Figure 4.

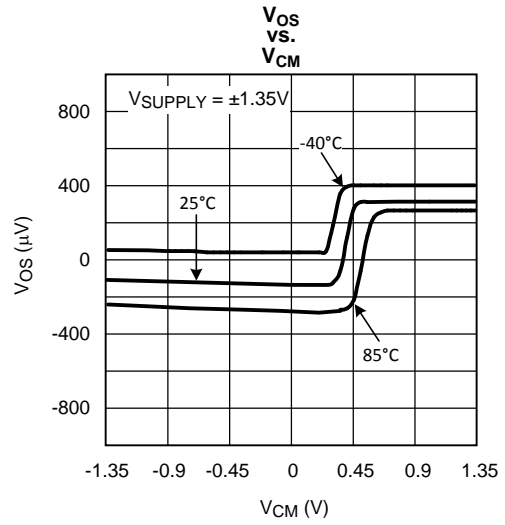


Figure 5.

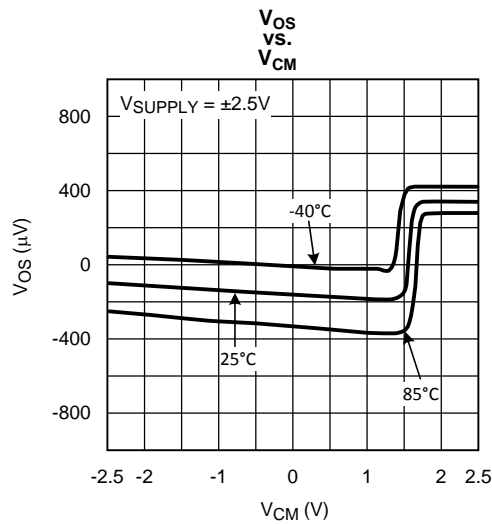


Figure 6.

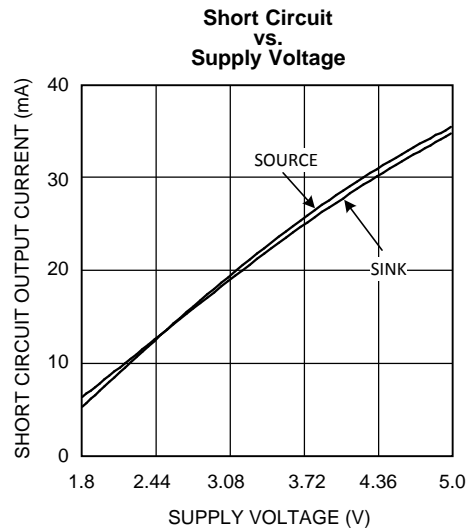


Figure 7.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).

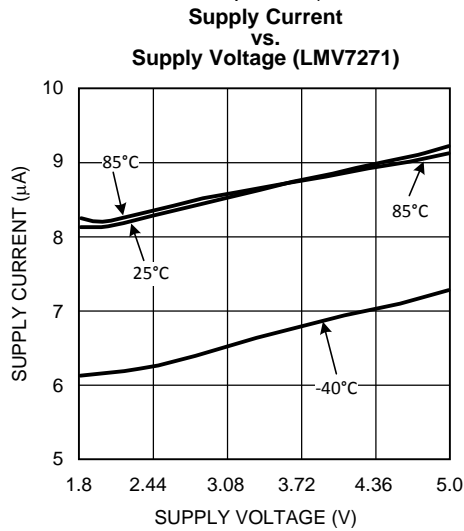


Figure 8.

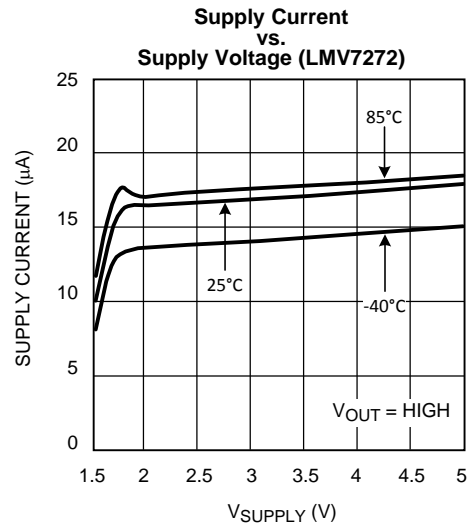


Figure 9.

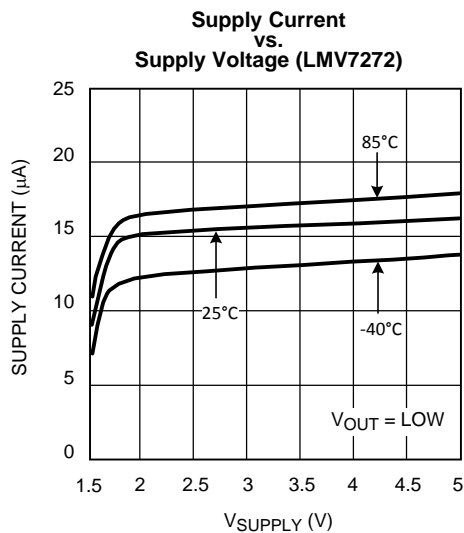


Figure 10.

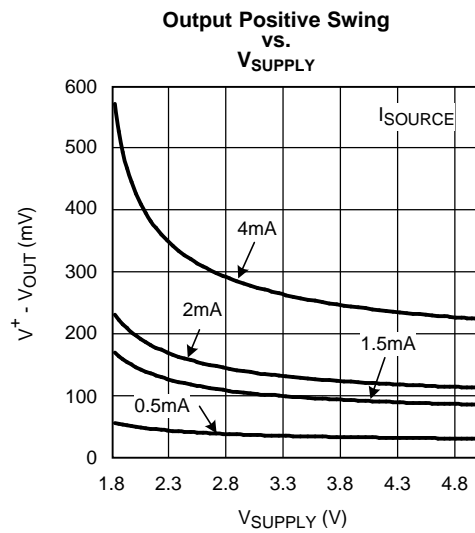


Figure 11.

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).

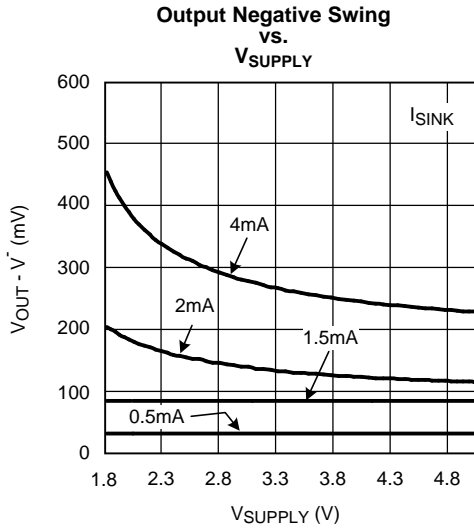


Figure 12.

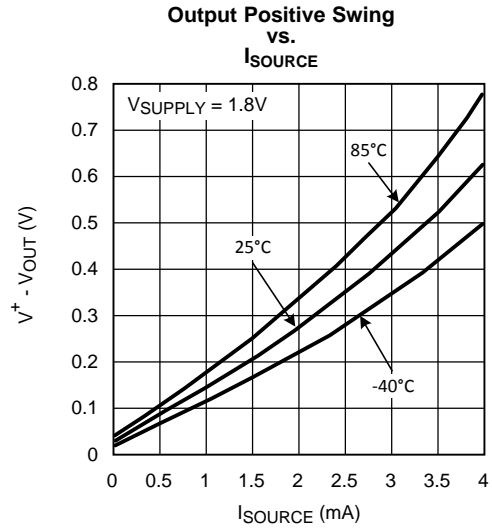


Figure 13.

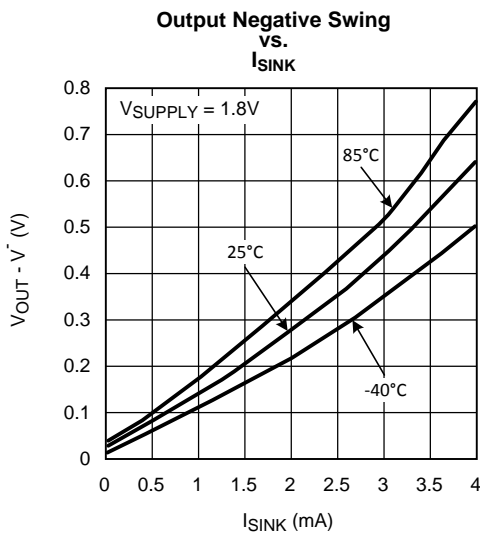


Figure 14.

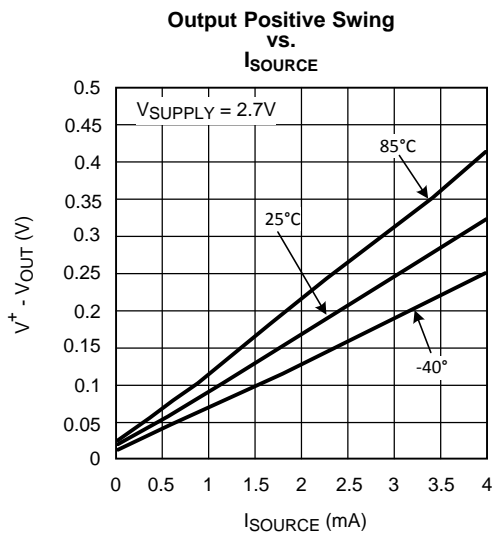


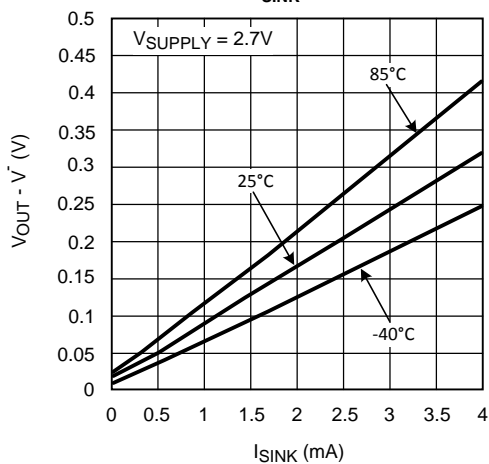
Figure 15.



**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

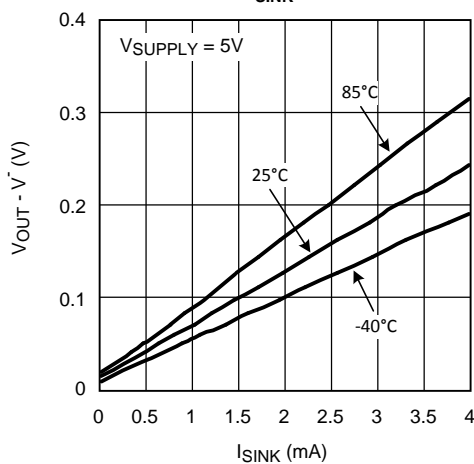
( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).

**Output Negative Swing vs.  $I_{\text{SINK}}$**



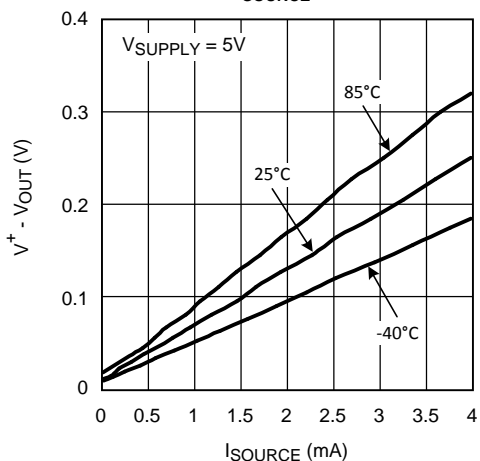
**Figure 16.**

**Output Negative Swing vs.  $I_{\text{SINK}}$**



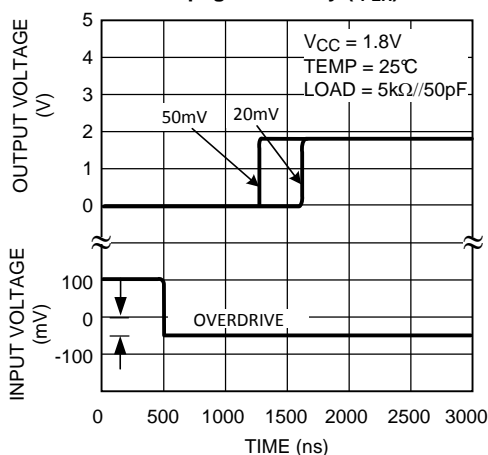
**Figure 17.**

**Output Positive Swing vs.  $I_{\text{SOURCE}}$**



**Figure 18.**

**Propagation Delay ( $t_{\text{PLH}}$ )**

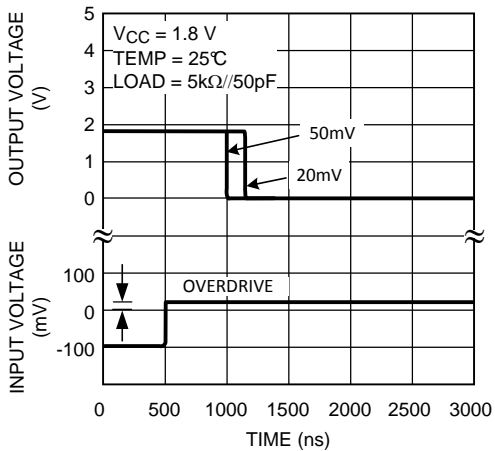


**Figure 19.**

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

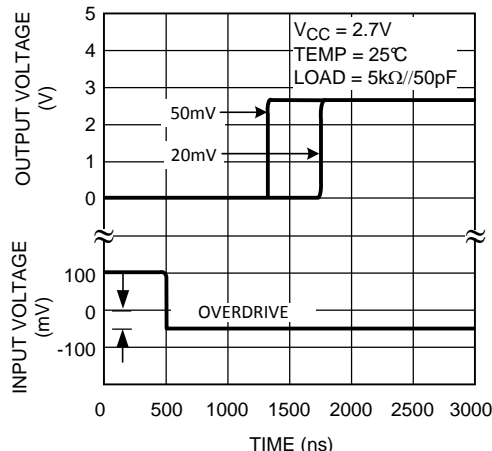
( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).

**Propagation Delay ( $t_{PHL}$ )**



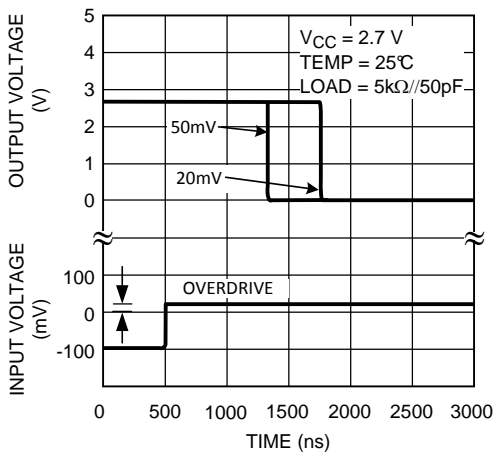
**Figure 20.**

**Propagation Delay ( $t_{PLH}$ )**



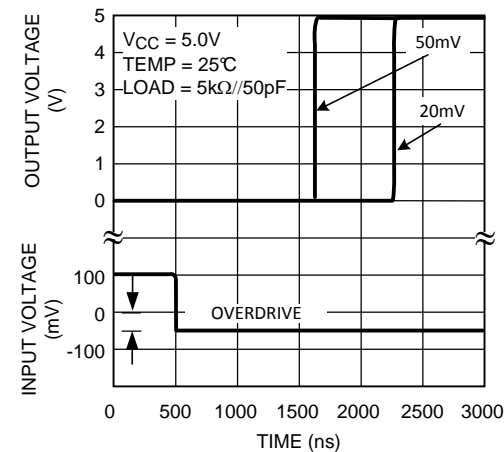
**Figure 21.**

**Propagation Delay ( $t_{PHL}$ )**



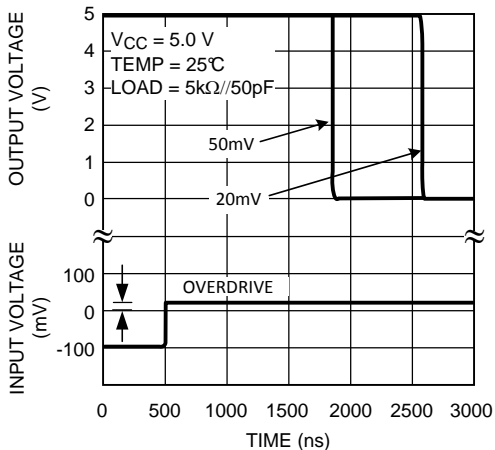
**Figure 22.**

**Propagation Delay ( $t_{PLH}$ )**



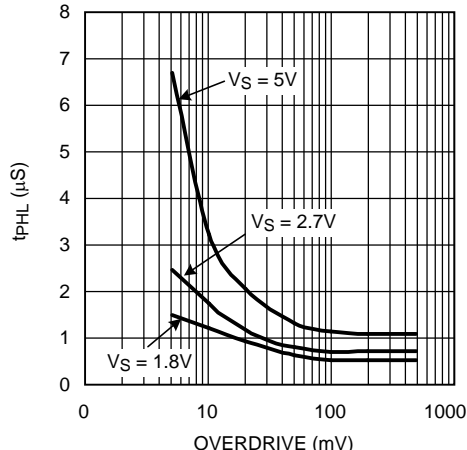
**Figure 23.**

**Propagation Delay ( $t_{PHL}$ )**



**Figure 24.**

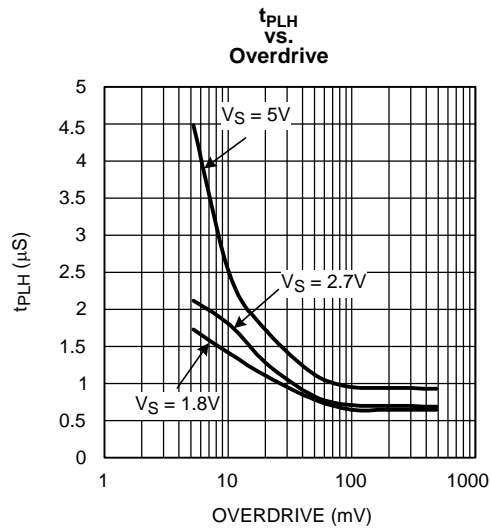
**$t_{PHL}$  vs. Overdrive**



**Figure 25.**

**TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

( $T_A = 25^\circ\text{C}$ , Unless otherwise specified).



**Figure 26.**

## APPLICATION NOTES

### BASIC COMPARATOR

A comparator is often used to convert an analog signal to a digital signal. As shown in [Figure 28](#), the comparator compares an input voltage ( $V_{IN}$ ) to a reference voltage ( $V_{REF}$ ). If  $V_{IN}$  is less than  $V_{REF}$ , the output ( $V_O$ ) is low. However, if  $V_{IN}$  is greater than  $V_{REF}$ , the output voltage ( $V_O$ ) is high.

Figure 27. LMV7271

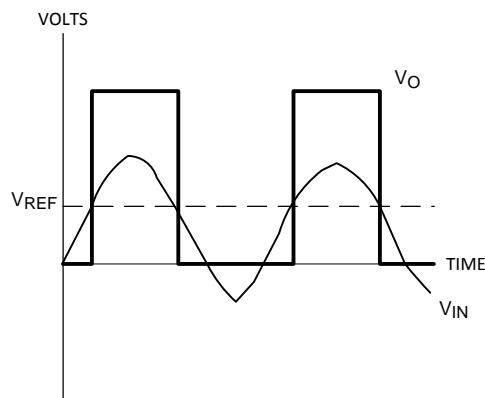
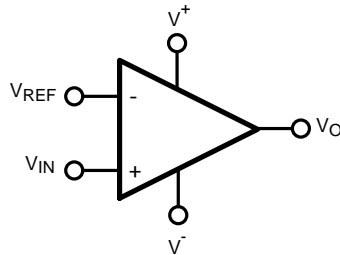


Figure 28. LMV7271 Basic Comparator

### RAIL-TO-RAIL INPUT STAGE

The LMV727X has an input common mode voltage range ( $V_{CM}$ ) of  $-0.1V$  below the  $V^-$  to  $0.1V$  above  $V^+$ . This is achieved by using paralleled PNP and NPN differential input pairs. When the  $V_{CM}$  is near  $V^+$ , the NPN pair is on and the PNP pair is off. When the  $V_{CM}$  is near  $V^-$ , the NPN pair is off and the PNP pair is on. The crossover point between the NPN and PNP input stages is around  $950mV$  from  $V^+$ . Since each input stage has its own offset voltage ( $V_{OS}$ ), the  $V_{OS}$  of the comparator becomes a function of the  $V_{CM}$ . See curves for  $V_{OS}$  vs.  $V_{CM}$  in Typical Performance Characteristics section. In application design, it is recommended to keep the  $V_{CM}$  away from the crossover point to avoid problems. The wide input voltage range makes LMV727X ideal in power supply monitoring circuits, where the comparators are used to sense signals close to ground and power supplies.

### OUTPUT STAGE

The LMV7271 and LMV7272 have a push-pull output stage. This output stage keeps the total system power consumption to the absolute minimum. The only current consumed is the low supply current and the current going directly into the load. When the output switches, both PMOS and NMOS at the output stage are on at the same time for a very short time. This allows current to flow directly between  $V^+$  and  $V^-$  through output transistors. The result is a short spike of current (shoot-through current) drawn from the supply and glitches in the supply voltages. The glitches can spread to other parts of the board as noise. To prevent the glitches in supply lines, power supply bypass capacitors must be installed. See section for supply bypassing in the Application Notes for details.

## HYSTERESIS

It is a standard procedure to use hysteresis (positive feedback) around a comparator, to prevent oscillation, and to avoid excessive noise on the output because the comparator is a good amplifier of its own noise.

### Inverting Comparator with Hysteresis

The inverting comparator with hysteresis requires a three resistor network that is referenced to the supply voltage  $V_{CC}$  of the comparator (Figure 29). When  $V_{IN}$  at the inverting input is less than  $V_{A1}$ , the voltage at the non-inverting node of the comparator ( $V_{IN} < V_{A1}$ ), the output voltage is high (for simplicity assume  $V_O$  switches as high as  $V_{CC}$ ). The three network resistors can be represented as  $R_1 || R_3$  in series with  $R_2$ . The lower input trip voltage  $V_{A1}$  is defined as

$$V_{A1} = \frac{V_{CC} R_2}{(R_1 || R_3) + R_2} \quad (1)$$

When  $V_{IN}$  is greater than  $V_A$  ( $V_{IN} > V_A$ ), the output voltage is low and very close to ground. In this case the three network resistors can be presented as  $R_2 // R_3$  in series with  $R_1$ . The upper trip voltage  $V_{A2}$  is defined as

$$V_{A2} = \frac{V_{CC} (R_2 || R_3)}{R_1 + (R_2 || R_3)} \quad (2)$$

The total hysteresis provided by the network is defined as

$$\Delta V_A = V_{A1} - V_{A2} \quad (3)$$

A good typical value of  $\Delta V_A$  would be in the range of 5 to 50mV. This is easily obtained by choosing  $R_3$  as 1000 to 100 times ( $R_1 || R_2$ ) for 5V operation, or as 300 to 30 times ( $R_1 || R_2$ ) for 1.8V operation.

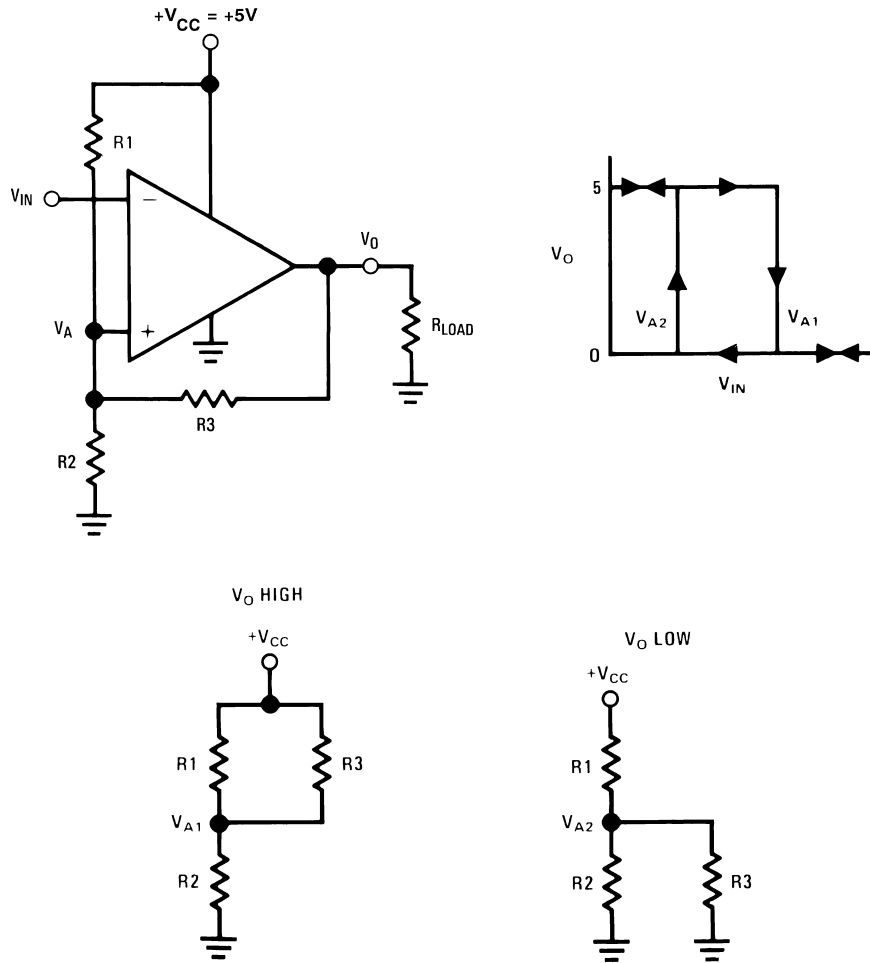


Figure 29. Inverting Comparator with Hysteresis

**Non-Inverting Comparator with Hysteresis**

A non-inverting comparator with hysteresis requires a two resistor network, and a voltage reference ( $V_{REF}$ ) at the inverting input (Figure 30). When  $V_{IN}$  is low, the output is also low. For the output to switch from low to high,  $V_{IN}$  must rise up to  $V_{IN1}$ , where  $V_{IN1}$  is calculated by

$$V_{in1} = \frac{V_{ref} (R_1 + R_2)}{R_2} \tag{4}$$

When  $V_{IN}$  is high, the output is also high. To make the comparator switch back to its low state,  $V_{IN}$  must equal  $V_{REF}$  before  $V_A$  will again equal  $V_{REF}$ .  $V_{IN}$  can be calculated by:

$$V_{in2} = \frac{V_{ref} (R_1 + R_2) - V_{CC} R_1}{R_2} \tag{5}$$

The hysteresis of this circuit is the difference between  $V_{IN1}$  and  $V_{IN2}$ .

$$\Delta V_{IN} = V_{CC} R_1 / R_2 \tag{6}$$

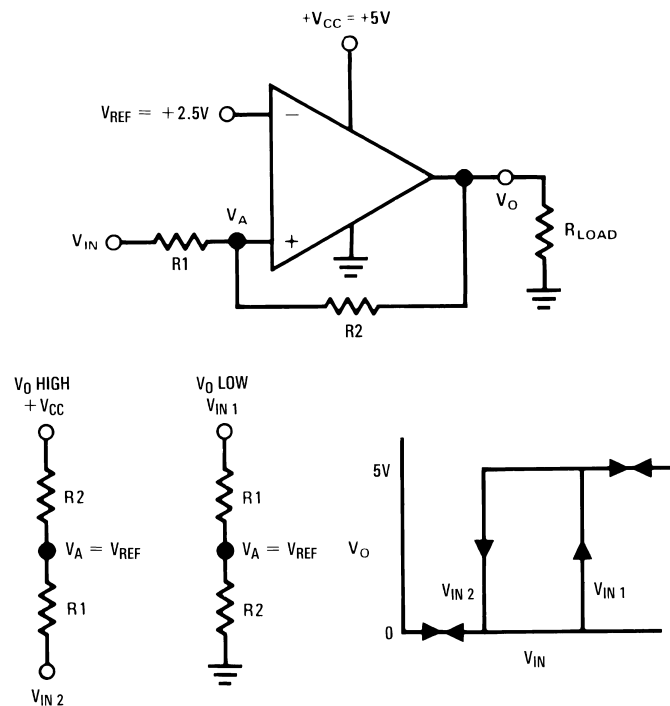
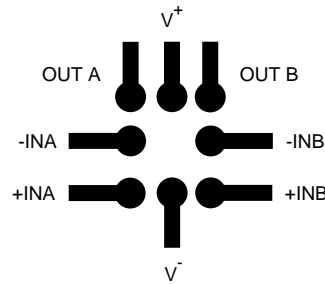


Figure 30. Non-Inverting Comparator with Hysteresis

## CIRCUIT TECHNIQUES FOR AVOIDING OSCILLATIONS IN COMPARATOR APPLICATIONS

Feedback to almost any pin of a comparator can result in oscillation. In addition, when the input signal is a slow voltage ramp or sine wave, the comparator may also burst into oscillation near the crossing point. To avoid oscillation or instability, PCB layout should be engineered thoughtfully. Several precautions are recommended:

1. Power supply bypassing is critical, and will improve stability and transient response. Resistance and inductance from power supply wires and board traces increase power supply line impedance. When supply current changes, the power supply line will move due to its impedance. Large enough supply line shift will cause the comparator to mis-operate. To avoid problems, a small bypass capacitor, such as 0.1 $\mu$ F ceramic, should be placed immediately adjacent to the supply pins. An additional 6.8 $\mu$ F or greater tantalum capacitor should be placed at the point where the power supply for the comparator is introduced onto the board. These capacitors act as an energy reservoir and keep the supply impedance low. In dual supply application, a 0.1 $\mu$ F capacitor is recommended to be placed across  $V^+$  and  $V^-$  pins.
2. Keep all leads short to reduce stray capacitance and lead inductance. It will also minimize any unwanted coupling from any high-level signals (such as the output). The comparators can easily oscillate if the output lead is inadvertently allowed to capacitively couple to the inputs via stray capacitance. This shows up only during the output voltage transition intervals as the comparator changes states. Try to avoid a long loop which could act as an inductor (coil).
3. It is a good practice to use an unbroken ground plane on a printed circuit board to provide all components with a low inductive ground connection. Make sure ground paths are low-impedance where heavier currents are flowing to avoid ground level shift. Preferably there should be a ground plane under the component.
4. The output trace should be routed away from inputs. The ground plane should extend between the output and inputs to act as a guard. This can be achieved by running a topside ground plane between the output and inputs. A typical PCB layout is shown in [Figure 31](#).



**Figure 31. Typical PCB Layout**

5. When the signal source is applied through a resistive network to one input of the comparator, it is usually advantageous to connect the other input with a resistor with the same value, for both DC and AC consideration. Input traces should be laid out symmetrically if possible.
6. All pins of any unused comparators should be tied to the negative supply.

### DSBGA LIGHT SENSITIVITY

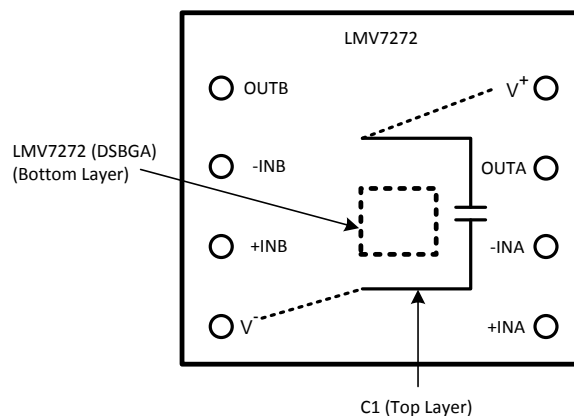
Exposing the DSBGA device to direct sunlight will cause mis-operation of the device. Light sources such as Halogen lamps can also affect electrical performance if brought near to the device. The wavelengths, which have the most detrimental effect, are reds and infrareds.

### DSBGA MOUNTING

The DSBGA package requires specific mounting techniques, which are detailed in Application Note AN-1112 ([SNVA009](#)).

### LMV7272 DSBGA to DIP Conversion Board

To facilitate characterization and testing, a DSBGA to DIP conversion board, LMV7272TLCONV, is available. It is a 2-layer board, with the LMV7272 mounted on the bottom layer, and a capacitor (C1, between the positive and negative supplies) added to the top layer.



**Figure 32. LMV7272TLCONV Diagram**

## Typical Applications

### UNIVERSAL LOGIC LEVEL SHIFTER

The output of LMV7275 is an unconnected drain of an NMOS device, which can be pulled up, through a resistor, to any desired output level within the permitted power supply range. Hence, the following simple circuit works as a universal logic level shifter, pulling up the signal to the desired level.



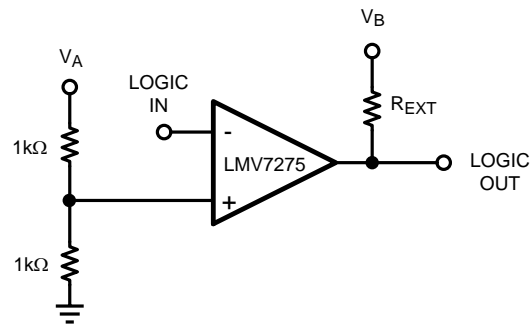


Figure 33. Logic Level Shifter

### POSITIVE PEAK DETECTOR

A positive peak detect circuit is basically a comparator operated in a unity gain follower configuration, with a capacitor as a load to maintain the highest voltage. A diode is added at the output to prevent the capacitor from discharging through the pull-up resistor, and a 1MΩ resistor added in parallel to the capacitor to provide a high impedance discharge path. When the input  $V_{IN}$  increases, the inverting input of the comparator follows it, thus charging the capacitor. When it decreases, the cap discharges through the 1MΩ resistor. The decay time can be modified by changing the resistor. The output should be accessed through a follower circuit to prevent loading.

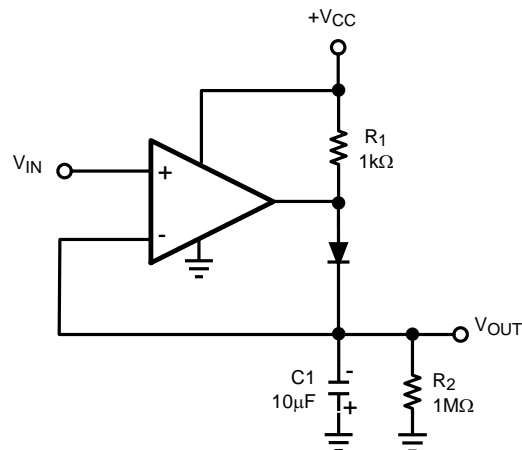


Figure 34. Positive Peak Detector

### OR'ING THE OUTPUT

Since the output is an unconnected NMOS drain, many drains can be tied together, pulled up to  $V_{DD}$  by a single resistor to provide an output OR'ing function. If any of the comparator outputs is pulled low the output  $V_O$  goes down.

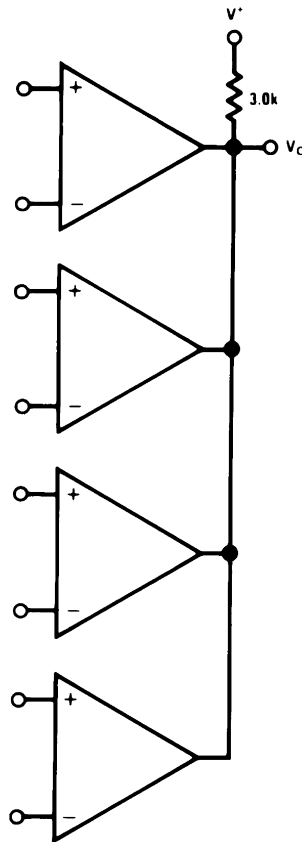


Figure 35. OR'ing the Outputs

### NEGATIVE PEAK DETECTOR

For the negative detector, the output transistor of the comparator acts as a low impedance current sink. Since there is no pull-up resistor, the only discharge path will be the  $1M\Omega$  resistor and any load impedance used. Decay time is changed by varying the  $1M\Omega$  resistor.

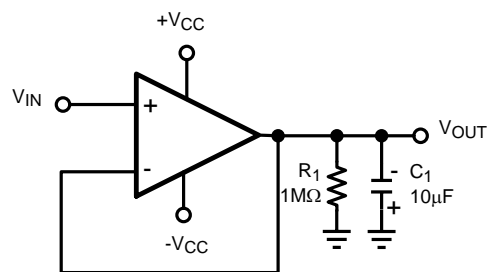


Figure 36. Negative Peak Detector

### SQUARE WAVE GENERATOR

A typical application for a comparator is as a square wave oscillator. The circuit below generates a square wave whose period is set by the RC time constant of the capacitor  $C_1$  and resistor  $R_4$ . The maximum frequency is limited by the large signal propagation delay of the comparator, and by the capacitive loading at the output, which limits the output slew rate.

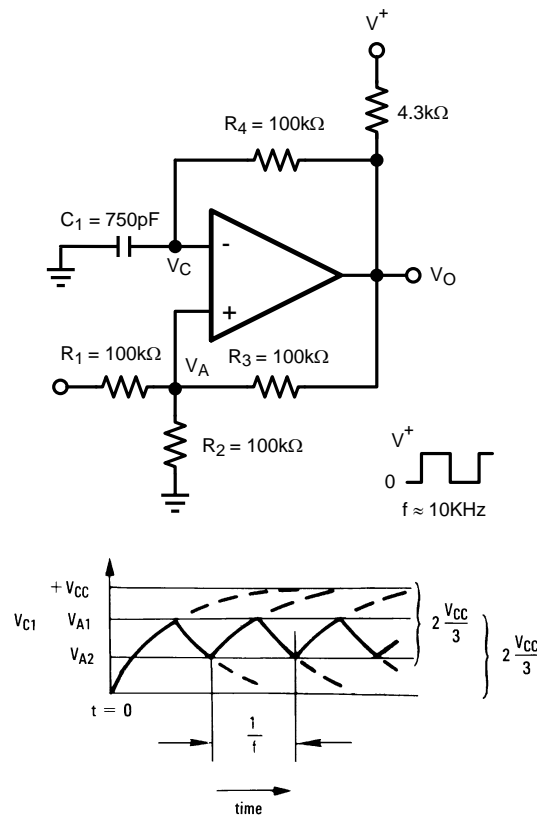


Figure 37. Squarewave Oscillator

To analyze the circuit, consider it when the output is high. That implies that the inverted input ( $V_C$ ) is lower than the non-inverting input ( $V_A$ ). This causes the  $C_1$  to get charged through  $R_4$ , and the voltage  $V_C$  increases till it is equal to the non-inverting input. The value of  $V_A$  at this point is

$$V_{A1} = \frac{V_{CC} \cdot R_2}{R_2 + R_1 \parallel R_3} \quad (7)$$

If  $R_1 = R_2 = R_3$ , then  $V_{A1} = 2V_{CC}/3$

At this point the comparator switches pulling down the output to the negative rail. The value of  $V_A$  at this point is

$$V_{A2} = \frac{V_{CC} (R_2 \parallel R_3)}{R_1 + (R_2 \parallel R_3)} \quad (8)$$

If  $R_1 = R_2 = R_3$ , then  $V_{A2} = V_{CC}/3$

The capacitor  $C_1$  now discharges through  $R_4$ , and the voltage  $V_C$  decreases till it is equal to  $V_{A2}$ , at which point the comparator switches again, bringing it back to the initial stage. The time period is equal to twice the time it takes to discharge  $C_1$  from  $2V_{CC}/3$  to  $V_{CC}/3$ , which is given by  $R_4 C_1 \cdot \ln 2$ . Hence the formula for the frequency is:

$$F = 1/(2 \cdot R_4 \cdot C_1 \cdot \ln 2)$$

### REVISION HISTORY

Changes from Revision G (February 2013) to Revision H	Page
• Changed layout of National Data Sheet to TI format .....	<a href="#">19</a>

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LMV7271MF	ACTIVE	SOT-23	DBV	5	1000	TBD	Call TI	Call TI	-40 to 85	C25A	<a href="#">Samples</a>
LMV7271MF/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C25A	<a href="#">Samples</a>
LMV7271MFX/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C25A	<a href="#">Samples</a>
LMV7271MG	ACTIVE	SC70	DCK	5	1000	TBD	Call TI	Call TI	-40 to 85	C34	<a href="#">Samples</a>
LMV7271MG/NOPB	ACTIVE	SC70	DCK	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C34	<a href="#">Samples</a>
LMV7271MGX	ACTIVE	SC70	DCK	5	3000	TBD	Call TI	Call TI	-40 to 85	C34	<a href="#">Samples</a>
LMV7271MGX/NOPB	ACTIVE	SC70	DCK	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C34	<a href="#">Samples</a>
LMV7272TL/NOPB	ACTIVE	DSBGA	YZR	8	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	C 01	<a href="#">Samples</a>
LMV7272TLX/NOPB	ACTIVE	DSBGA	YZR	8	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85	C 01	<a href="#">Samples</a>
LMV7275MF	ACTIVE	SOT-23	DBV	5	1000	TBD	Call TI	Call TI	-40 to 85	C26A	<a href="#">Samples</a>
LMV7275MF/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C26A	<a href="#">Samples</a>
LMV7275MFX	ACTIVE	SOT-23	DBV	5	3000	TBD	Call TI	Call TI	-40 to 85	C26A	<a href="#">Samples</a>
LMV7275MFX/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C26A	<a href="#">Samples</a>
LMV7275MG	ACTIVE	SC70	DCK	5	1000	TBD	Call TI	Call TI	-40 to 85	C35	<a href="#">Samples</a>
LMV7275MG/NOPB	ACTIVE	SC70	DCK	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C35	<a href="#">Samples</a>
LMV7275MGX	ACTIVE	SC70	DCK	5	3000	TBD	Call TI	Call TI	-40 to 85	C35	<a href="#">Samples</a>
LMV7275MGX/NOPB	ACTIVE	SC70	DCK	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	C35	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:  
**ACTIVE:** Product device recommended for new designs.

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**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

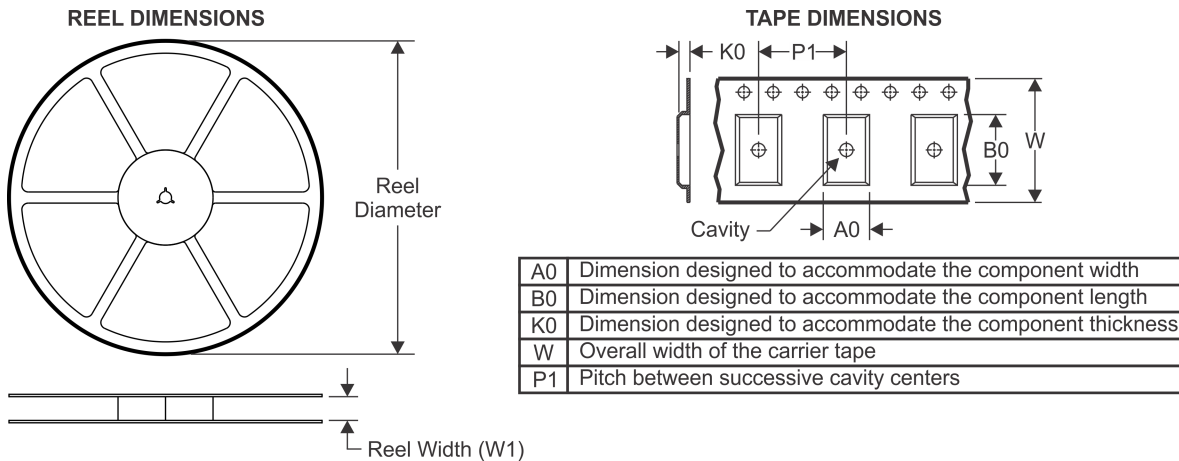
<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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## TAPE AND REEL INFORMATION



### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMV7271MF	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7271MF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7271MFX/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7271MG	SC70	DCK	5	1000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7271MG/NOPB	SC70	DCK	5	1000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7271MGX	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7271MGX/NOPB	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7272TL/NOPB	DSBGA	YZR	8	250	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LMV7272TLX/NOPB	DSBGA	YZR	8	3000	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LMV7275MF	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7275MF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7275MFX	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7275MFX/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMV7275MG	SC70	DCK	5	1000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7275MG/NOPB	SC70	DCK	5	1000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7275MGX	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3
LMV7275MGX/NOPB	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMV7271MF	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMV7271MF/NOPB	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMV7271MFX/NOPB	SOT-23	DBV	5	3000	210.0	185.0	35.0
LMV7271MG	SC70	DCK	5	1000	210.0	185.0	35.0
LMV7271MG/NOPB	SC70	DCK	5	1000	210.0	185.0	35.0
LMV7271MGX	SC70	DCK	5	3000	210.0	185.0	35.0
LMV7271MGX/NOPB	SC70	DCK	5	3000	210.0	185.0	35.0
LMV7272TL/NOPB	DSBGA	YZR	8	250	210.0	185.0	35.0
LMV7272TLX/NOPB	DSBGA	YZR	8	3000	210.0	185.0	35.0
LMV7275MF	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMV7275MF/NOPB	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMV7275MFX	SOT-23	DBV	5	3000	210.0	185.0	35.0
LMV7275MFX/NOPB	SOT-23	DBV	5	3000	210.0	185.0	35.0
LMV7275MG	SC70	DCK	5	1000	210.0	185.0	35.0
LMV7275MG/NOPB	SC70	DCK	5	1000	210.0	185.0	35.0
LMV7275MGX	SC70	DCK	5	3000	210.0	185.0	35.0
LMV7275MGX/NOPB	SC70	DCK	5	3000	210.0	185.0	35.0



DBV (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
  - D. Falls within JEDEC MO-178 Variation AA.

DBV (R-PDSO-G5)

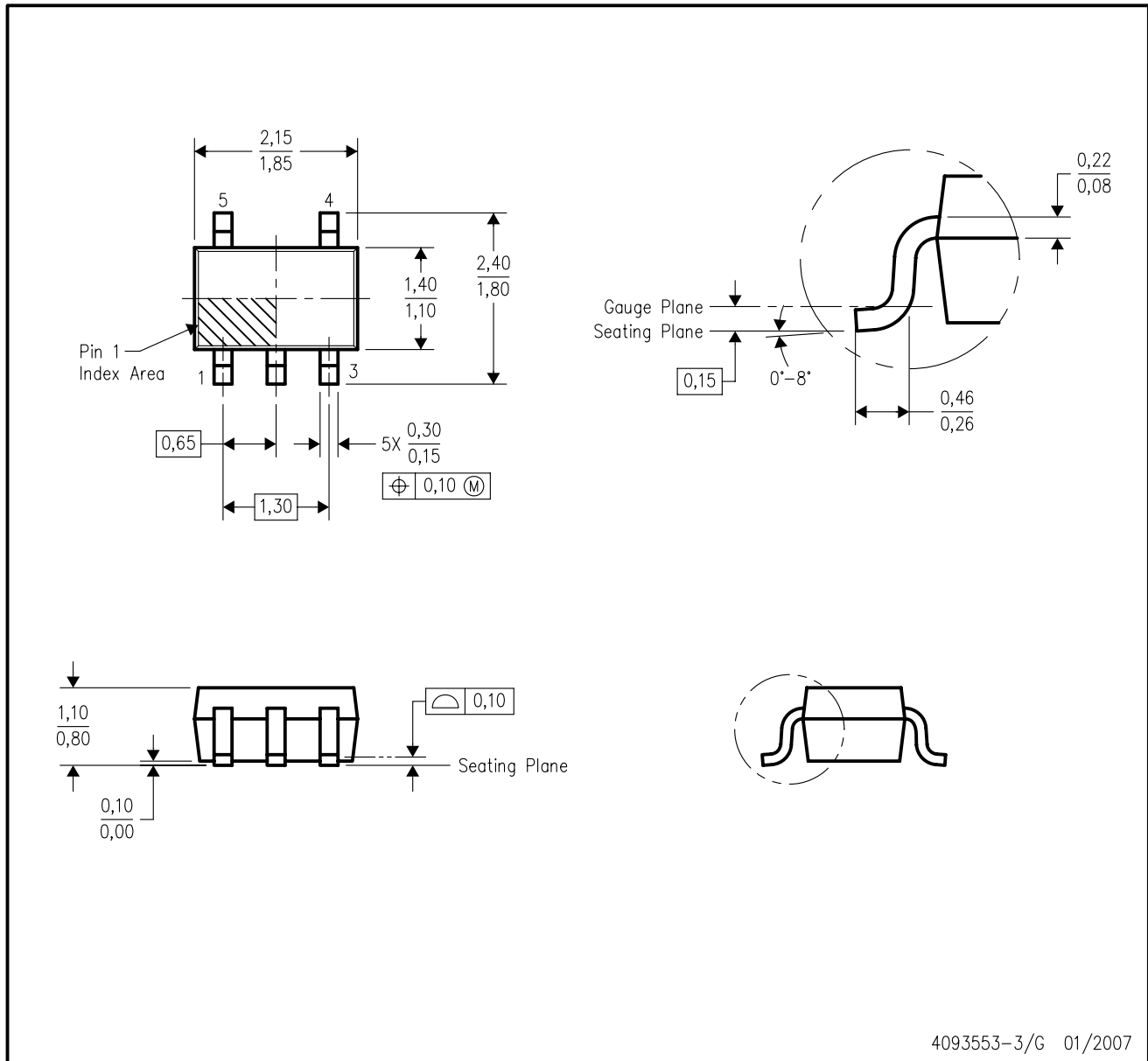
PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
  - D. Publication IPC-7351 is recommended for alternate designs.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.

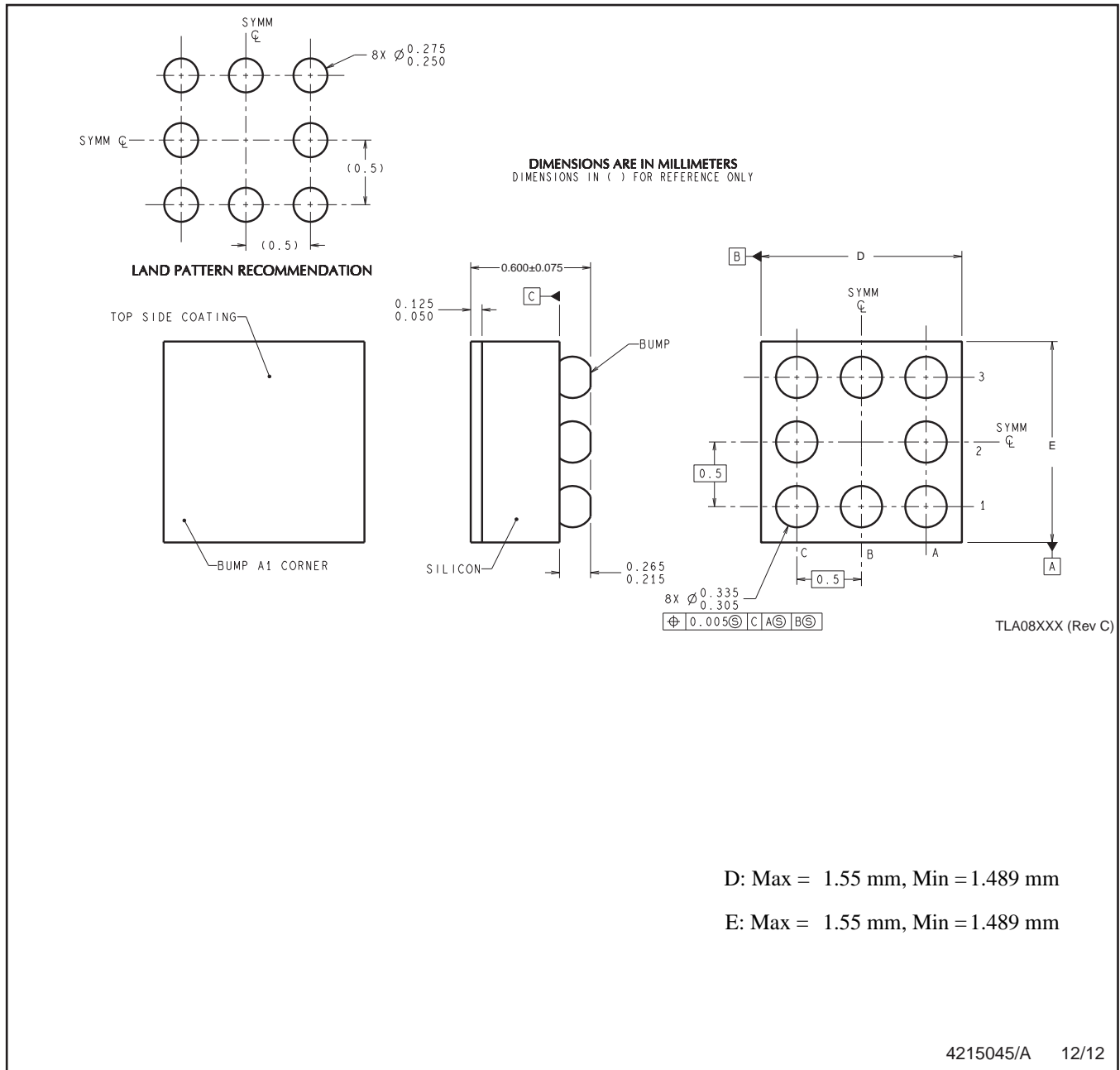
DCK (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
  - Falls within JEDEC MO-203 variation AA.

YZR0008



D: Max = 1.55 mm, Min = 1.489 mm

E: Max = 1.55 mm, Min = 1.489 mm

4215045/A 12/12

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.  
 B. This drawing is subject to change without notice.

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