

# ADS54J69

## Dual-Channel, 16-Bit, 500-MSPS, Analog-to-Digital Converter

### 1 Features

- 16-Bit Resolution, Dual-Channel, 500-MSPS ADC
- Idle Channel Noise Floor:  $-159$  dBFS/Hz
- Spectral Performance ( $f_{IN} = 170$  MHz at  $-1$  dBFS):
  - SNR: 73 dBFS
  - NSD:  $-157$  dBFS/Hz
  - SFDR: 93 dBc
  - SFDR: 94 dBc (Except HD2, HD3, and Interleaving Tone)
- Spectral Performance ( $f_{IN} = 310$  MHz at  $-1$  dBFS):
  - SNR: 71.7 dBFS
  - NSD:  $-155.7$  dBFS/Hz
  - SFDR: 81 dBc
  - SFDR: 94 dBc (Except HD2, HD3, and Interleaving Tone)
- Channel Isolation: 100 dBc at  $f_{IN} = 170$  MHz
- Input Full-Scale:  $1.9 V_{PP}$
- Input Bandwidth (3 dB): 1.2 GHz
- On-Chip Dither
- Integrated Decimate-by-2 Filter
- JESD204B Interface with Subclass 1 Support:
  - 1 Lane per ADC at 10.0 Gbps
  - 2 Lanes per ADC at 5.0 Gbps
  - Support for Multi-Chip Synchronization
- Power Dissipation: 1.35 W/ch at 500 MSPS
- 72-Pin VQFN Package (10 mm x 10 mm)

### 2 Applications

- Radar and Antenna Arrays
- Broadband Wireless
- Cable CMTS, DOCSIS 3.1 Receivers
- Communications Test Equipment
- Microwave Receivers
- Software Defined Radio (SDR)
- Digitizers
- Medical Imaging and Diagnostics

### 3 Description

The ADS54J69 is a low-power, wide-bandwidth, 16-bit, 500-MSPS, dual-channel, analog-to-digital converter (ADC). Designed for high signal-to-noise ratio (SNR), the device delivers a noise floor of  $-159$  dBFS/Hz for applications aiming for highest dynamic range over a wide instantaneous bandwidth. The device supports the JESD204B serial interface with data rates up to 10.0 Gbps, supporting one or two lanes per ADC. The buffered analog input provides uniform input impedance across a wide frequency range and minimizes sample-and-hold glitch energy. Each ADC channel is directly connected to a wideband digital down-converter (DDC) block. The ADS54J69 provides excellent spurious-free dynamic range (SFDR) over a large input frequency range with very low power consumption.

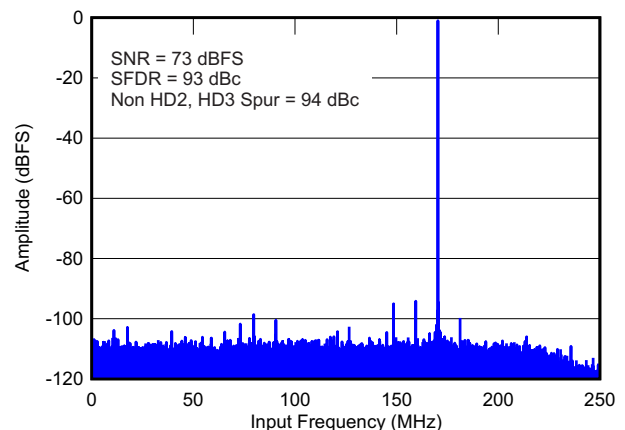
The JESD204B interface reduces the number of interface lines, allowing high system integration density. An internal phase-locked loop (PLL) multiplies the ADC sampling clock to derive the bit clock that is used to serialize the 16-bit data from each channel.

#### Device Information

PART NUMBER	SPEED GRADE (MSPS)	RESOLUTION (Bits)
<a href="#">ADS54J40</a>	1000	14
<a href="#">ADS54J60</a>	1000	16
<a href="#">ADS54J66</a>	500	14
<a href="#">ADS54J69</a>	500	16

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### Spectrum at 170-MHz IF



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## 4 Revision History

### Changes from Revision A (January 2016) to Revision B

**Page**

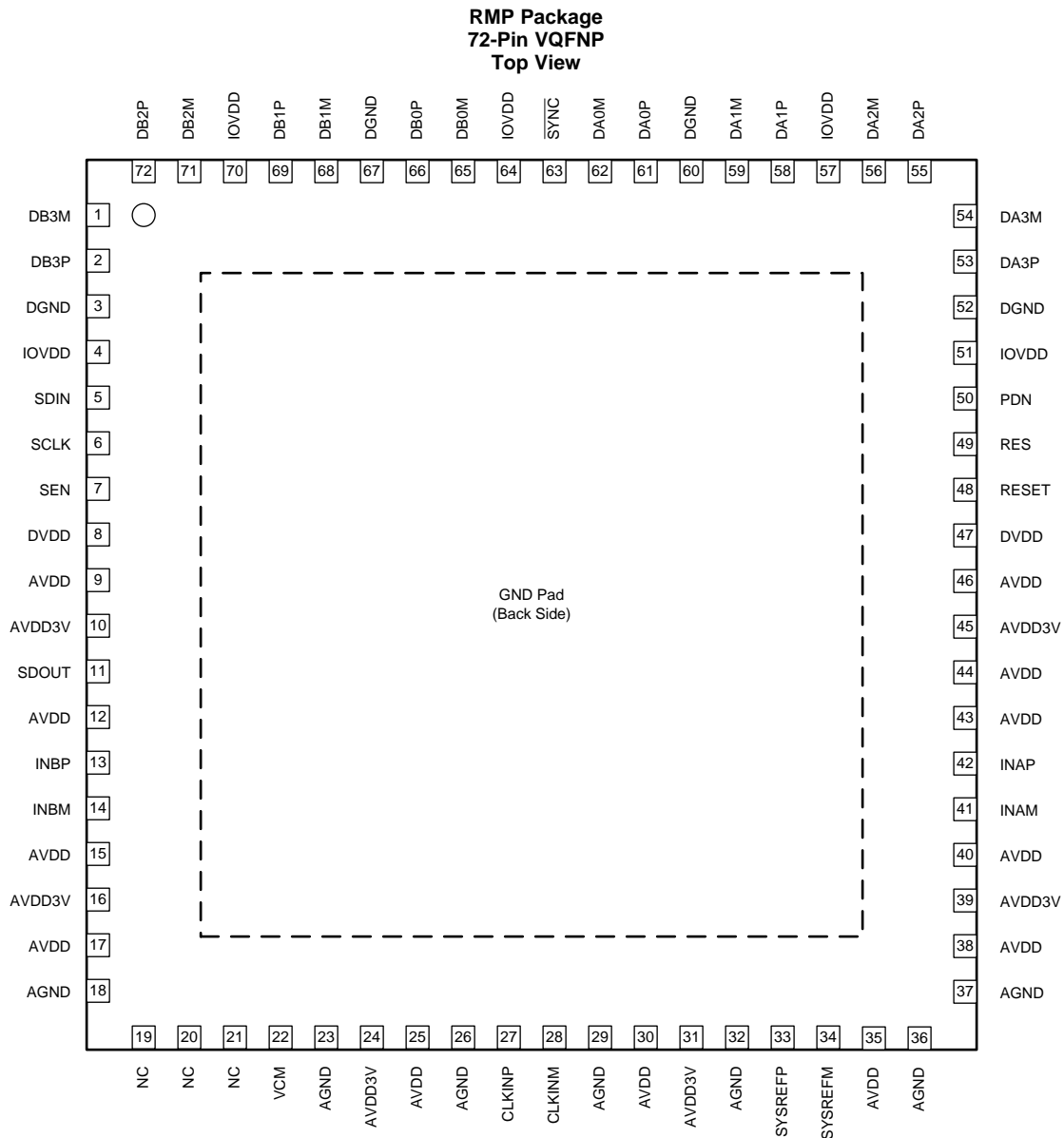
• Changed Sample Timing, <i>Aperture jitter</i> parameter in <i>Timing Characteristics</i> table .....	<b>10</b>
• Changed <a href="#">Table 19</a> and <a href="#">Table 21</a> .....	<b>43</b>
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• Changed <a href="#">Table 40</a> .....	<b>50</b>
• Changed <a href="#">Table 42</a> .....	<b>51</b>
• Changed <i>SNR and Clock Jitter</i> section: changed <a href="#">Figure 129</a> and last sentence of section .....	<b>63</b>
• Changed <i>Application Curves</i> section .....	<b>66</b>

### Changes from Original (May 2015) to Revision A

**Page**

• Released to production .....	<b>1</b>
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## 5 Pin Configuration and Functions



### Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
<b>CLOCK, SYSREF</b>			
CLKINM	28	I	Negative differential clock input for the ADC
CLKINP	27	I	Positive differential clock input for the ADC
SYSREFM	34	I	Negative external SYSREF input
SYSREFP	33	I	Positive external SYSREF input
<b>CONTROL, SERIAL INTERFACE</b>			
PDN	50	I/O	Power-down. Can be configured via an SPI register setting. Can be configured to fast overrange output for channel A via the SPI.
RESET	48	I	Hardware reset; active high. This pin has an internal 20-kΩ pulldown resistor.
SCLK	6	I	Serial interface clock input
SDIN	5	I	Serial interface data input
SDOUT	11	O	Serial interface data output. Can be configured to fast overrange output for channel B via the SPI.
SEN	7	I	Serial interface enable
<b>DATA INTERFACE</b>			
DA0M	62	O	JESD204B serial data negative outputs for channel A
DA1M	59		
DA2M	56		
DA3M	54		
DA0P	61	O	JESD204B serial data positive outputs for channel A
DA1P	58		
DA2P	55		
DA3P	53		
DB0M	65	O	JESD204B serial data negative outputs for channel B
DB1M	68		
DB2M	71		
DB3M	1		
DB0P	66	O	JESD204B serial data positive outputs for channel B
DB1P	69		
DB2P	72		
DB3P	2		
SYN $\bar{C}$	63	I	Synchronization input for JESD204B port
<b>INPUT, COMMON MODE</b>			
INAM	41	I	Differential analog negative input for channel A
INAP	42	I	Differential analog positive input for channel A
INBM	14	I	Differential analog negative input for channel B
INBP	13	I	Differential analog positive input for channel B
VCM	22	O	Common-mode voltage, 2.1 V. Note that analog inputs are internally biased to this pin through 600 Ω (effective), no external connection from the VCM pin to the INxP or INxM pin is required.
<b>POWER SUPPLY</b>			
AGND	18, 23, 26, 29, 32, 36, 37	I	Analog ground
AVDD	9, 12, 15, 17, 25, 30, 35, 38, 40, 43, 44, 46	I	Analog 1.9-V power supply
AVDD3V	10, 16, 24, 31, 39, 45	I	Analog 3.0-V power supply for the analog buffer
DGND	3, 52, 60, 67	I	Digital ground
DVDD	8, 47	I	Digital 1.9-V power supply
IOVDD	4, 51, 57, 64, 70	I	Digital 1.15-V power supply for the JESD204B transmitter
<b>NC, RES</b>			
NC	19, 20, 21	—	Unused pins, do not connect
RES	49	I	Reserved pin. Connect to DGND.

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
Supply voltage range	AVDD3V	-0.3	3.6	V
	AVDD	-0.3	2.1	
	DVDD	-0.3	2.1	
	IOVDD	-0.2	1.4	
Voltage between AGND and DGND		-0.3	0.3	V
Voltage applied to input pins	INAP, INBP, INAM, INBM	-0.3	3	V
	CLKINP, CLKINM	-0.3	AVDD + 0.3	
	SYSREFP, SYSREFM	-0.3	AVDD + 0.3	
	SCLK, SEN, SDIN, RESET, $\overline{\text{SYNC}}$ , PDN	-0.2	2.1	
Storage temperature, T <sub>stg</sub>		-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

		VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±1000
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±500

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.  
(2) JEDEC document JEP157 states that 250-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)</sup>

		MIN	NOM	MAX	UNIT
Supply voltage range	AVDD3V	2.85	3.0	3.6	V
	AVDD	1.8	1.9	2.0	
	DVDD	1.7	1.9	2.0	
	IOVDD	1.1	1.15	1.2	
Analog inputs	Differential input voltage range	1.9			V <sub>PP</sub>
	Input common-mode voltage	2.0			V
	Maximum analog input frequency for 1.9-V <sub>PP</sub> input amplitude <sup>(3)(4)</sup>	400			MHz
Clock inputs	Input clock frequency, device clock frequency		500	1000	MHz
	Input clock amplitude differential (V <sub>CLKP</sub> - V <sub>CLKM</sub> )	Sine wave, ac-coupled	0.75	1.5	V <sub>PP</sub>
		LVPECL, ac-coupled	0.8	1.6	
		LVDS, ac-coupled	0.7		
Input device clock duty cycle		45%	50%	55%	
Temperature	Operating free-air, T <sub>A</sub>		-40	85	°C
	Operating junction, T <sub>J</sub>		105 <sup>(5)</sup>		

- (1) SYSREF must be applied for the device to initialize; see the *SYSREF Signal* section for details.  
(2) After power-up, always use a hardware reset to reset the device for the first time; see [Table 58](#) for details.  
(3) Operating 0.5 dB below the maximum-supported amplitude is recommended to accommodate gain mismatch in interleaving ADCs.  
(4) At high frequencies, the maximum supported input amplitude reduces; see [Figure 51](#) for details.  
(5) Prolonged use above the nominal junction temperature can increase the device failure-in-time (FIT) rate.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		ADS54J69	
		RMP (VQFNP)	
		72 PINS	
			UNIT
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	22.3	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	5.1	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	2.4	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	0.1	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	2.3	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	0.4	°C/W

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

## 6.5 Electrical Characteristics

typical values are at T<sub>A</sub> = 25°C, full temperature range is from T<sub>MIN</sub> = -40°C to T<sub>MAX</sub> = 85°C, device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and -1-dBFS differential input (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>GENERAL</b>						
	Device clock frequency				1000	MSPS
	Output sample rate				500	MSPS
	Resolution		16			Bits
<b>POWER SUPPLIES</b>						
AVDD3V	3.0-V analog supply		2.85	3.0	3.6	V
AVDD	1.9-V analog supply		1.8	1.9	2.0	V
DVDD	1.9-V digital supply		1.7	1.9	2.0	V
IOVDD	1.15-V SERDES supply		1.1	1.15	1.2	V
I <sub>AVDD3V</sub>	3.0-V analog supply current	V <sub>IN</sub> = full-scale on both channels		293	360	mA
I <sub>AVDD</sub>	1.9-V analog supply current	V <sub>IN</sub> = full-scale on both channels		354	510	mA
I <sub>DVDD</sub>	1.9-V digital supply current	Four-lane output mode (default after reset)		188	260	mA
I <sub>IOVDD</sub>	1.15-V SERDES supply current			512	920	mA
P <sub>dis</sub>	Total power dissipation			2.66	3.1	W
I <sub>DVDD</sub>	1.9-V digital supply current	Two-lane output mode		195		mA
I <sub>IOVDD</sub>	1.15-V SERDES supply current			559		mA
P <sub>dis</sub>	Total power dissipation			2.73		W
	Global power-down power dissipation	Using the GLOBAL PDN register bit in the master page		204	315	mW

## 6.6 AC Characteristics

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SNR	Signal-to-noise ratio	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		74.2		dBFS
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		73.4		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	71.3	73		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		72.7		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		71.7		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		70.3		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		70.5		
NSD	Noise spectral density	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		158.2		dBFS/Hz
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		157.4		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	155.3	157.0		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		156.7		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		155.7		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		154.3		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		154.5		
SINAD	Signal-to-noise and distortion ratio	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		73.8		dBFS
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		73.3		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	69.8	72.9		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		72.5		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		71.2		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		70.2		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		69.4		
SFDR	Spurious free dynamic range (excluding IL spurs)	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		86		dBc
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		95		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	79	94		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		89		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		81		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		87		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		73		
HD2	Second-order harmonic distortion	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		86		dBc
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		104		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	85	102		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		95		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		81		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		87		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		96		
HD3	Third-order harmonic distortion	$f_{\text{IN}} = 10\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		89		dBc
		$f_{\text{IN}} = 140\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		103		
		$f_{\text{IN}} = 170\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$	86	101		
		$f_{\text{IN}} = 210\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		100		
		$f_{\text{IN}} = 310\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		98		
		$f_{\text{IN}} = 370\text{ MHz}, A_{\text{IN}} = -1\text{ dBFS}$		95		
		$f_{\text{IN}} = 470\text{ MHz}, A_{\text{IN}} = -3\text{ dBFS}$		73		

**AC Characteristics (continued)**

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle,  $\text{AVDD}3\text{V} = 3.0\text{ V}$ ,  $\text{AVDD} = \text{DVDD} = 1.9\text{ V}$ ,  $\text{IOVDD} = 1.15\text{ V}$ , and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Non HD2, HD3	Spurious-free dynamic range (excluding HD2, HD3, and IL spur)	$f_{\text{IN}} = 10\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		98		dBc
		$f_{\text{IN}} = 140\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		95		
		$f_{\text{IN}} = 170\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$	84	94		
		$f_{\text{IN}} = 210\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		89		
		$f_{\text{IN}} = 310\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		92		
		$f_{\text{IN}} = 370\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		97		
		$f_{\text{IN}} = 470\text{ MHz}$ , $A_{\text{IN}} = -3\text{ dBFS}$		92		
ENOB	Effective number of bits	$f_{\text{IN}} = 10\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		12		Bits
		$f_{\text{IN}} = 140\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		11.9		
		$f_{\text{IN}} = 170\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$	11.3	11.9		
		$f_{\text{IN}} = 210\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		11.8		
		$f_{\text{IN}} = 310\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		11.5		
		$f_{\text{IN}} = 370\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		11.4		
		$f_{\text{IN}} = 470\text{ MHz}$ , $A_{\text{IN}} = -3\text{ dBFS}$		11.2		
THD	Total harmonic distortion	$f_{\text{IN}} = 10\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		84		dBc
		$f_{\text{IN}} = 140\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		95		
		$f_{\text{IN}} = 170\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$	79	89		
		$f_{\text{IN}} = 210\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		85		
		$f_{\text{IN}} = 310\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		80		
		$f_{\text{IN}} = 370\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		85		
		$f_{\text{IN}} = 470\text{ MHz}$ , $A_{\text{IN}} = -3\text{ dBFS}$		72		
SFDR_IL	Interleaving spur	$f_{\text{IN}} = 10\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		90		dBc
		$f_{\text{IN}} = 140\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		90		
		$f_{\text{IN}} = 170\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$	75	87		
		$f_{\text{IN}} = 210\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		85		
		$f_{\text{IN}} = 310\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		85		
		$f_{\text{IN}} = 370\text{ MHz}$ , $A_{\text{IN}} = -1\text{ dBFS}$		86		
		$f_{\text{IN}} = 470\text{ MHz}$ , $A_{\text{IN}} = -3\text{ dBFS}$		82		
IMD3	Two-tone, third-order intermodulation distortion	$f_{\text{IN}1} = 185\text{ MHz}$ , $f_{\text{IN}2} = 190\text{ MHz}$ , $A_{\text{IN}} = -7\text{ dBFS}$		86		dBFS
		$f_{\text{IN}1} = 365\text{ MHz}$ , $f_{\text{IN}2} = 370\text{ MHz}$ , $A_{\text{IN}} = -7\text{ dBFS}$		79		
		$f_{\text{IN}1} = 465\text{ MHz}$ , $f_{\text{IN}2} = 470\text{ MHz}$ , $A_{\text{IN}} = -10\text{ dBFS}$		78		



## 6.7 Digital Characteristics

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle,  $AVDD3V = 3.0\text{ V}$ ,  $AVDD = DVDD = 1.9\text{ V}$ ,  $IOVDD = 1.15\text{ V}$ , and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNITS
<b>DIGITAL INPUTS (RESET, SCLK, SEN, SDIN, <math>\overline{\text{SYNC}}</math>, PDN)<sup>(1)</sup></b>						
$V_{\text{IH}}$	High-level input voltage	All digital inputs support 1.2-V and 1.8-V logic levels	0.8			V
$V_{\text{IL}}$	Low-level input voltage	All digital inputs support 1.2-V and 1.8-V logic levels			0.4	V
$I_{\text{IH}}$	High-level input current	SEN		0		$\mu\text{A}$
		RESET, SCLK, SDIN, PDN, $\overline{\text{SYNC}}$		50		
$I_{\text{IL}}$	Low-level input current	SEN		50		$\mu\text{A}$
		RESET, SCLK, SDIN, PDN, $\overline{\text{SYNC}}$		0		
<b>DIGITAL INPUTS (SYSREFP, SYSREFM)</b>						
$V_{\text{D}}$	Differential input voltage		0.35	0.45	1.4	V
$V_{\text{(CM\_DIG)}}$	Common-mode voltage for SYSREF <sup>(2)</sup>			1.3		V
<b>DIGITAL OUTPUTS (SDOUT, PDN)<sup>(3)</sup></b>						
$V_{\text{OH}}$	High-level output voltage		$DVDD - 0.1$	$DVDD$		V
$V_{\text{OL}}$	Low-level output voltage				0.1	V
<b>DIGITAL OUTPUTS (JESD204B Interface: DxP, DxM)<sup>(4)</sup></b>						
$V_{\text{OD}}$	Output differential voltage	With default swing setting		700		$\text{mV}_{\text{PP}}$
$V_{\text{OC}}$	Output common-mode voltage			450		mV
	Transmitter short-circuit current	Transmitter pins shorted to any voltage between $-0.25\text{ V}$ and $1.45\text{ V}$	$-100$		100	mA
$Z_{\text{OS}}$	Single-ended output impedance			50		$\Omega$
	Output capacitance	Output capacitance inside the device, from either output to ground		2		pF

- (1) The RESET, SCLK, SDIN, and PDN pins have a 20-k $\Omega$  (typical) internal pulldown resistor to ground, and the SEN pin has a 20-k $\Omega$  (typical) pullup resistor to IOVDD.
- (2) The SYSREFP and SYSREFM pins are internally biased to the 1.3-V common-mode voltage through a 5-k $\Omega$  resistor.
- (3) When functioning as an OVR pin for channel B.
- (4) 100- $\Omega$  differential termination.

### 6.8 Timing Characteristics

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle,  $\text{AVDD3V} = 3.0\text{ V}$ ,  $\text{AVDD} = \text{DVDD} = 1.9\text{ V}$ ,  $\text{IOVDD} = 1.15\text{ V}$ , and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

	MIN	TYP	MAX	UNITS
<b>SAMPLE TIMING</b>				
Aperture delay	0.75		1.6	ns
Aperture delay matching between two channels on the same device		$\pm 70$		ps
Aperture delay matching between two devices at the same temperature and supply voltage		$\pm 270$		ps
Aperture jitter	Actual jitter of sampling clock buffer		145	$f_s$ rms
	Effective jitter after decimation filtering		102	
<b>WAKE-UP TIMING</b>				
Wake-up time to valid data after coming out of global power-down		150		$\mu\text{s}$
<b>LATENCY</b>				
Data latency <sup>(1)</sup> : ADC sample to digital output		134 <sup>(2)</sup>		Input clock cycles
OVR latency: ADC sample to OVR bit		62		Input clock cycles
$t_{\text{PD}}$ Propagation delay: logic gates and output buffers delay (does not change with $f_s$ )		4		ns
<b>SYSREF TIMING</b>				
$t_{\text{SU\_SYSREF}}$ Setup time for SYSREF, referenced to the input clock falling edge	300		900	ps
$t_{\text{H\_SYSREF}}$ Hold time for SYSREF, referenced to the input clock falling edge	100			ps
<b>JESD OUTPUT INTERFACE TIMING CHARACTERISTICS</b>				
Unit interval	100		400	ps
Serial output data rate	2.5		10	Gbps
Total jitter for BER of 1E-15 and lane rate = 10 Gbps		26		ps
Random jitter for BER of 1E-15 and lane rate = 10 Gbps		0.75		ps rms
Deterministic jitter for BER of 1E-15 and lane rate = 10 Gbps		12		ps, pk-pk
$t_{\text{R}}, t_{\text{F}}$ Data rise time, data fall time: rise and fall times are measured from 20% to 80%, differential output waveform, 2.5 Gbps $\leq$ bit rate $\leq$ 10 Gbps		35		ps

- (1) Overall latency = data latency + decimation filter delay +  $t_{\text{PD}}$ .
- (2) Decimation filter latency is not included in this specification.

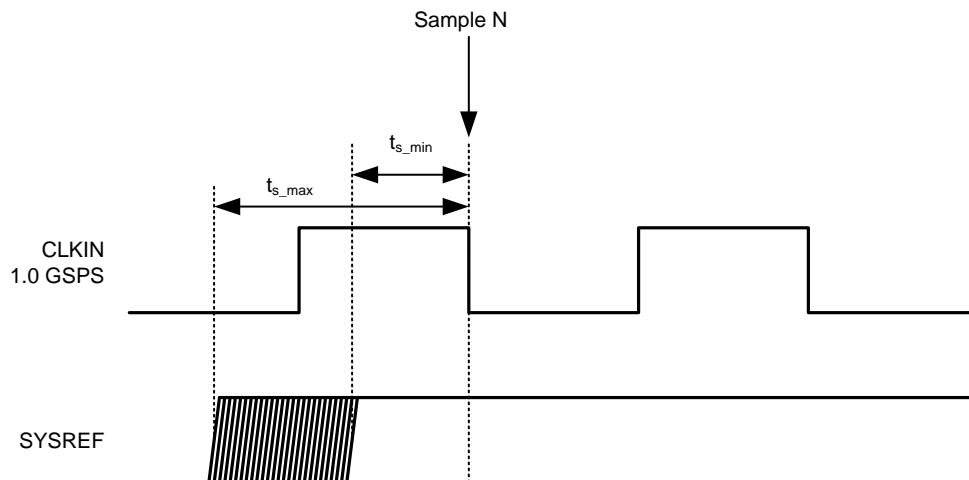
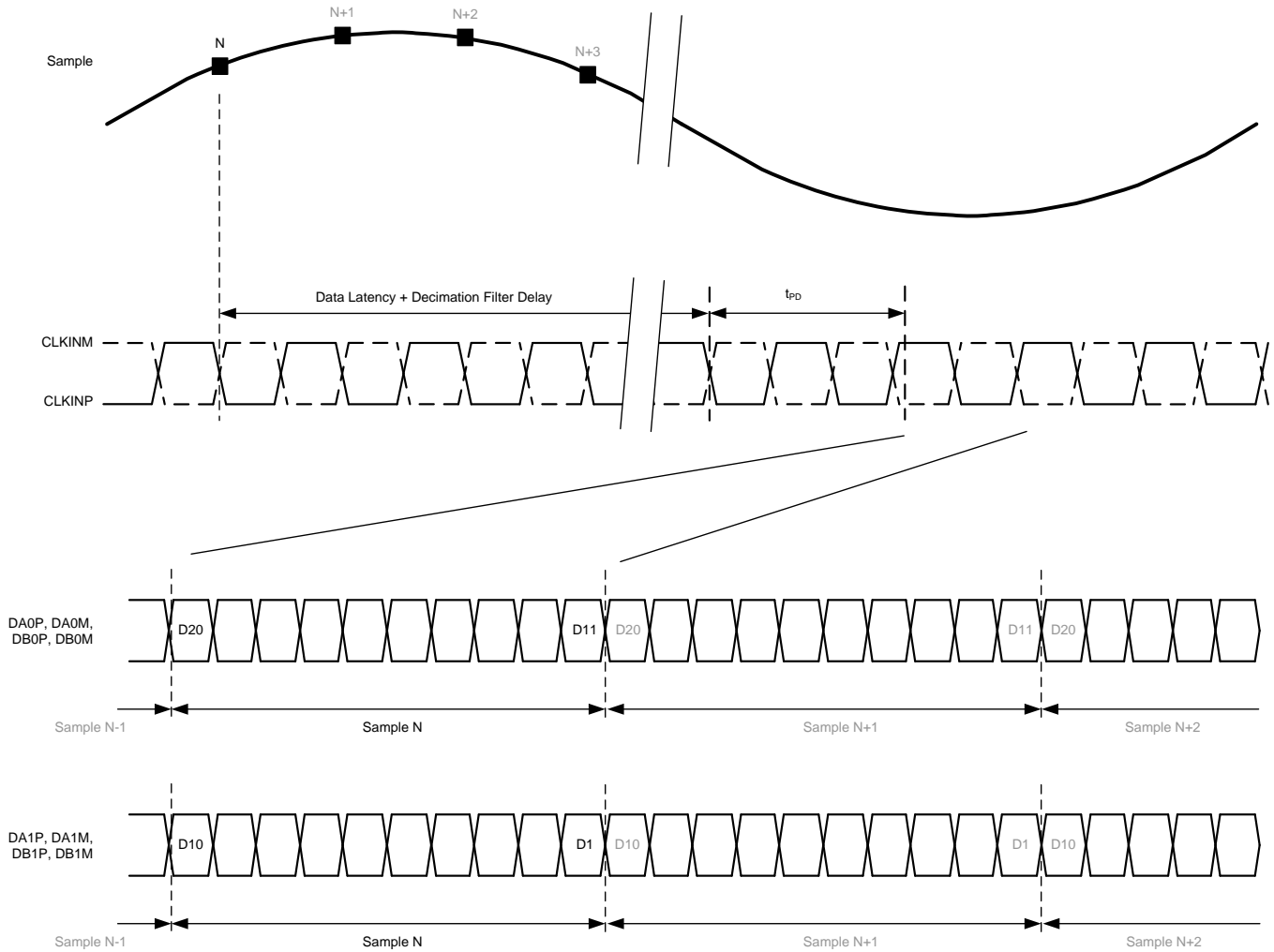


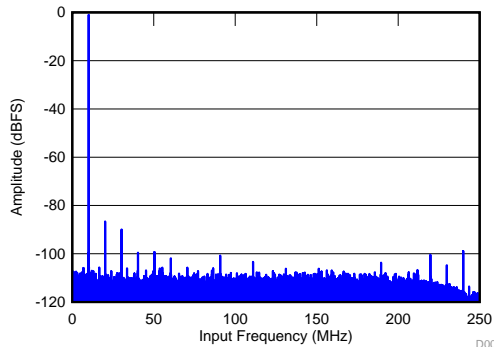
Figure 1. SYSREF Timing



**Figure 2. Sample Timing Requirements**

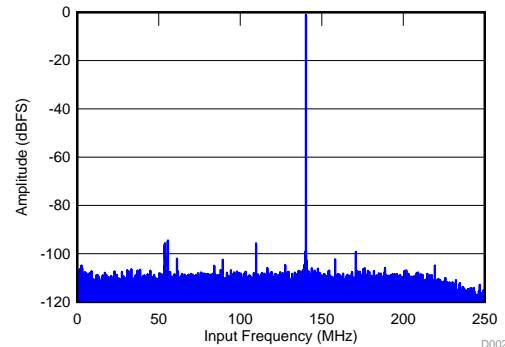
## 6.9 Typical Characteristics

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle,  $\text{AVDD}3\text{V} = 3.0\text{ V}$ ,  $\text{AVDD} = \text{DVDD} = 1.9\text{ V}$ ,  $\text{IOVDD} = 1.15\text{ V}$ , and  $-1\text{-dBFS}$  differential input (unless otherwise noted)



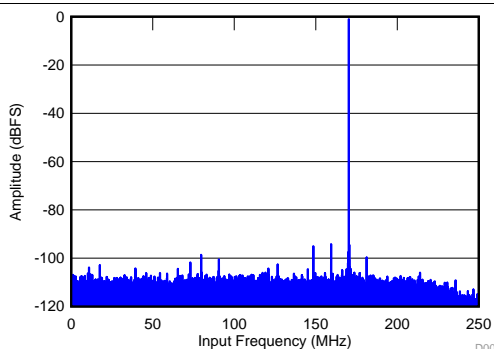
SNR = 74.2 dBFS; SFDR = 86 dBc; SINAD = 73.8 dBFS;  
 THD = 83 dBc; HD2 = 86 dBc; HD3 = 89 dBc;  
 IL spur = 99 dBc; non HD2, HD3 spur = 98 dBc

**Figure 3. FFT for 10-MHz Input Signal**



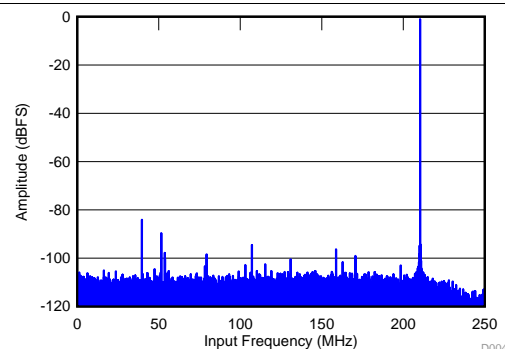
SNR = 73.3 dBFS; SFDR = 94 dBc; SINAD = 73.25 dBFS;  
 THD = 93 dBc; HD2 = 104 dBc; HD3 = 111 dBc;  
 IL spur = 95 dBc; non HD2, HD3 spur = 94 dBc

**Figure 4. FFT for 140-MHz Input Signal**



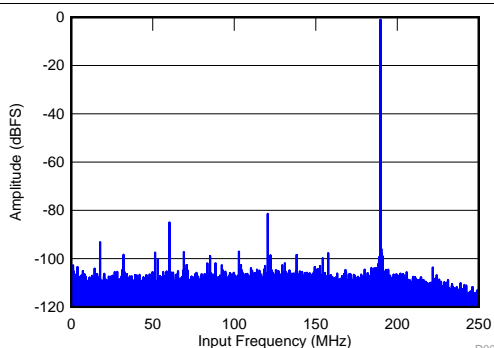
SNR = 73 dBFS; SFDR = 93 dBc; SINAD = 73.18 dBFS;  
 THD = 89 dBc; HD2 = 93 dBc; HD3 = 103 dBc;  
 IL spur = 99 dBc; non HD2, HD3 spur = 94 dBc

**Figure 5. FFT for 170-MHz Input Signal**



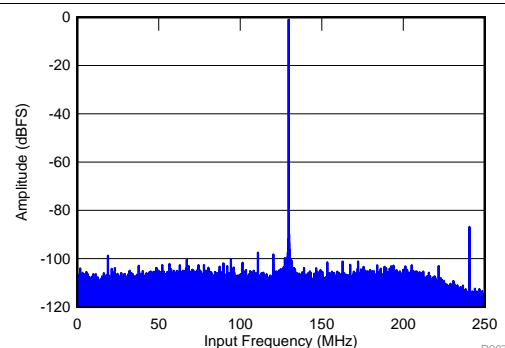
SNR = 72.8 dBFS; SFDR = 89 dBc; SINAD = 72.63 dBFS;  
 THD = 86 dBc; HD2 = 97 dBc; HD3 = 99 dBc;  
 IL spur = 84 dBc; non HD2, HD3 spur = 89 dBc

**Figure 6. FFT for 210-MHz Input Signal**



SNR = 71.6 dBFS; SFDR = 80 dBc; SINAD = 71 dBFS;  
 THD = 79 dBc; HD2 = -80 dBc; HD3 = -96 dBc;  
 IL spur = 85 dBc; non HD2, HD3 spur = 92 dBc

**Figure 7. FFT for 310-MHz Input Signal**

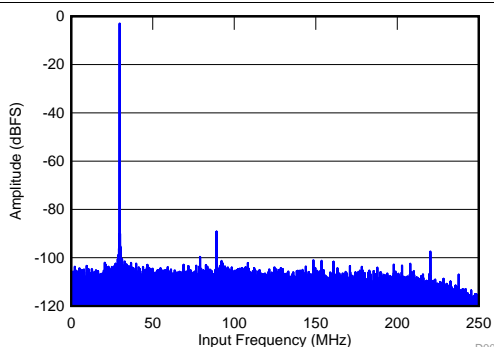


SNR = 70.5 dBFS; SFDR = 86 dBc; SINAD = 70.4 dBFS;  
 THD = 85 dBc; HD2 = -86 dBc; HD3 = -96 dBc;  
 IL spur = 98 dBc; non HD2, HD3 spur = 98 dBc

**Figure 8. FFT for 370-MHz Input Signal**

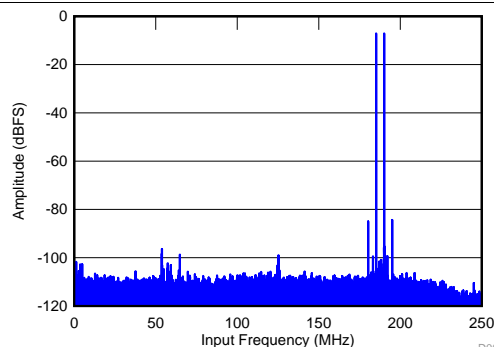
**Typical Characteristics (continued)**

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and  $-1\text{-dBFS}$  differential input (unless otherwise noted)



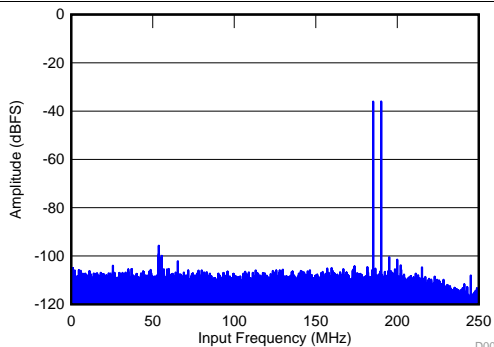
SNR = 70.6 dBFS; SFDR = 86 dBc; SINAD = 70.55 dBFS;  
THD = 85 dBc; tone at  $-3\text{ dBFS}$ ; HD2 = 102 dBc; HD3 = 86 dBc;  
IL spur = 97 dBc; non HD2, HD3 spur = 96 dBc

**Figure 9. FFT for 470-MHz Input Signal**



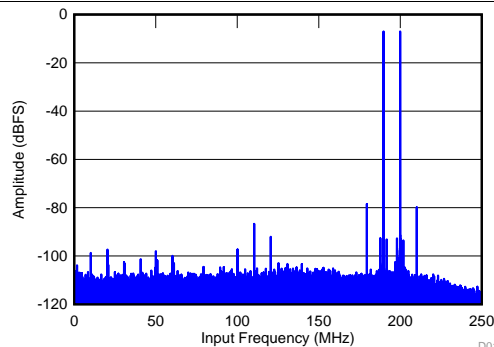
$f_{\text{IN1}} = 185\text{ MHz}$ ,  $f_{\text{IN2}} = 190\text{ MHz}$ , each tone at  $-7\text{ dBFS}$ ,  
IMD = 86 dBFS

**Figure 10. FFT for Two-Tone Input Signal ( $-7\text{ dBFS}$ )**



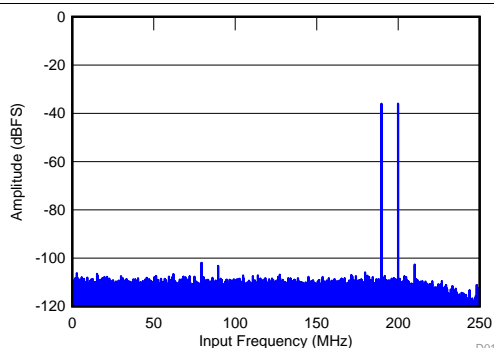
$f_{\text{IN1}} = 185\text{ MHz}$ ,  $f_{\text{IN2}} = 190\text{ MHz}$ , each tone at  $-36\text{ dBFS}$ ,  
IMD = 101 dBFS

**Figure 11. FFT for Two-Tone Input Signal ( $-36\text{ dBFS}$ )**



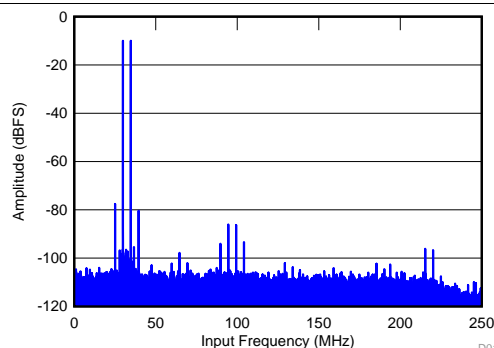
$f_{\text{IN1}} = 300\text{ MHz}$ ,  $f_{\text{IN2}} = 310\text{ MHz}$ , each tone at  $-7\text{ dBFS}$ ,  
IMD = 79 dBFS

**Figure 12. FFT for Two-Tone Input Signal ( $-7\text{ dBFS}$ )**



$f_{\text{IN1}} = 300\text{ MHz}$ ,  $f_{\text{IN2}} = 310\text{ MHz}$ , each tone at  $-36\text{ dBFS}$ ,  
IMD3 = 102 dBFS

**Figure 13. FFT for Two-Tone Input Signal ( $-36\text{ dBFS}$ )**

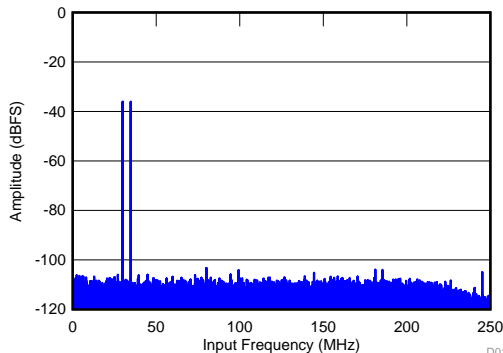


$f_{\text{IN1}} = 470\text{ MHz}$ ,  $f_{\text{IN2}} = 465\text{ MHz}$ , each tone at  $-10\text{ dBFS}$ ,  
IMD = 78 dBFS

**Figure 14. FFT for Two-Tone Input Signal ( $-10\text{ dBFS}$ )**

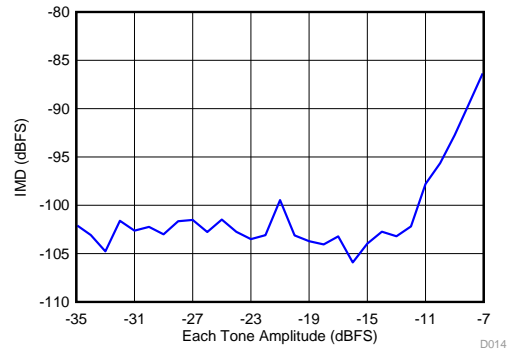
**Typical Characteristics (continued)**

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and  $-1\text{-dBFS}$  differential input (unless otherwise noted)



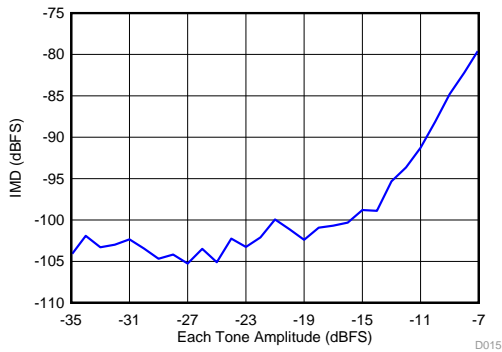
$f_{\text{IN1}} = 470\text{ MHz}$ ,  $f_{\text{IN2}} = 465\text{ MHz}$ , each tone at  $-36\text{ dBFS}$ ,  
IMD3 = 104 dBFS

**Figure 15. FFT for Two-Tone Input Signal ( $-36\text{ dBFS}$ )**



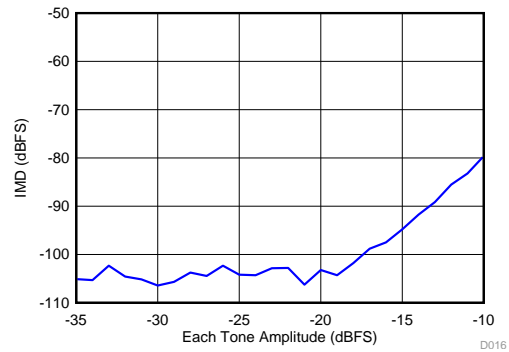
$f_{\text{IN1}} = 185\text{ MHz}$ ,  $f_{\text{IN2}} = 190\text{ MHz}$

**Figure 16. Intermodulation Distortion vs Input Amplitude (185 MHz and 190 MHz)**



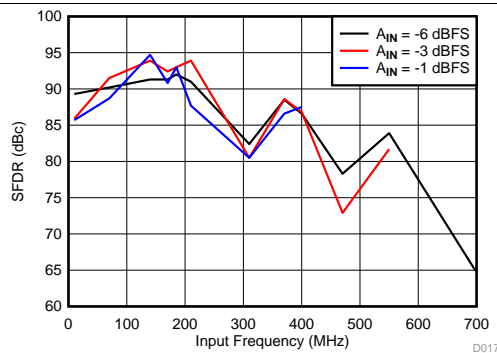
$f_{\text{IN1}} = 300\text{ MHz}$ ,  $f_{\text{IN2}} = 310\text{ MHz}$

**Figure 17. Intermodulation Distortion vs Input Amplitude (365 MHz and 370 MHz)**

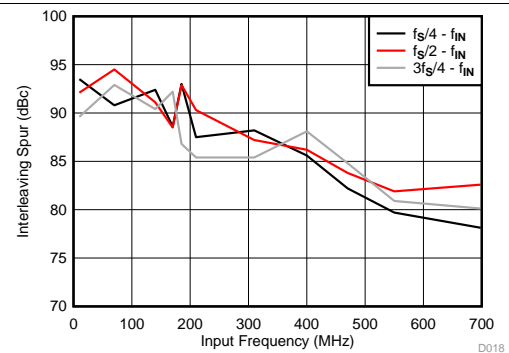


$f_{\text{IN1}} = 465\text{ MHz}$ ,  $f_{\text{IN2}} = 470\text{ MHz}$

**Figure 18. Intermodulation Distortion vs Input Amplitude (465 MHz and 470 MHz)**



**Figure 19. Spurious-Free Dynamic Range vs Input Frequency**



**Figure 20. IL Spur vs Input Frequency**

Typical Characteristics (continued)

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and -1-dBFS differential input (unless otherwise noted)

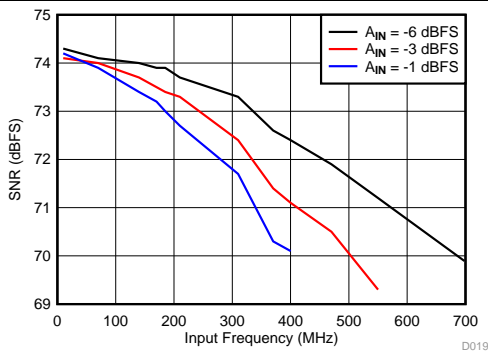


Figure 21. Signal-to-Noise Ratio vs Input Frequency

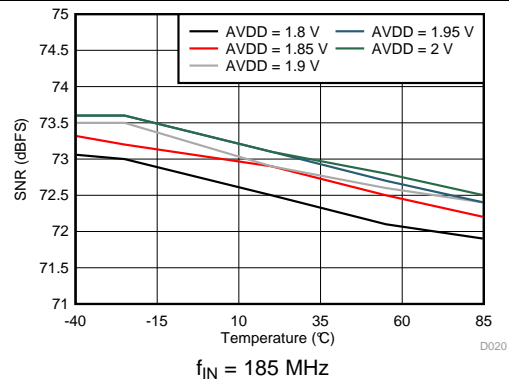


Figure 22. Signal-to-Noise Ratio vs AVDD Supply and Temperature

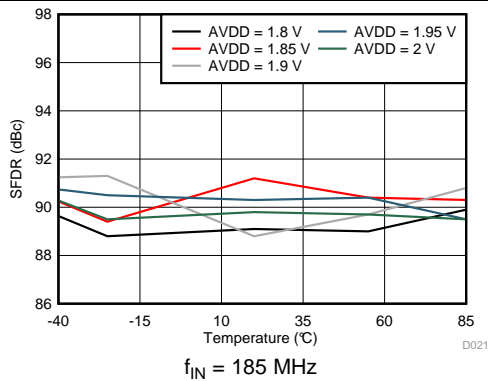


Figure 23. Spurious-Free Dynamic Range vs AVDD Supply and Temperature

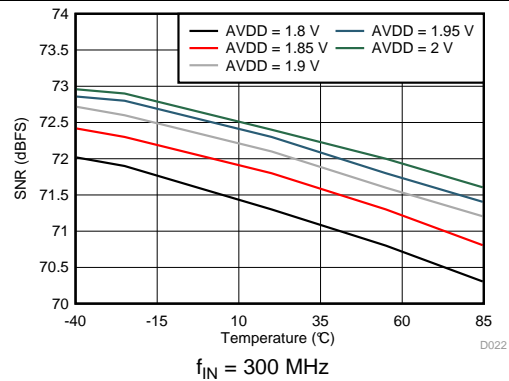


Figure 24. Signal-to-Noise Ratio vs AVDD Supply and Temperature

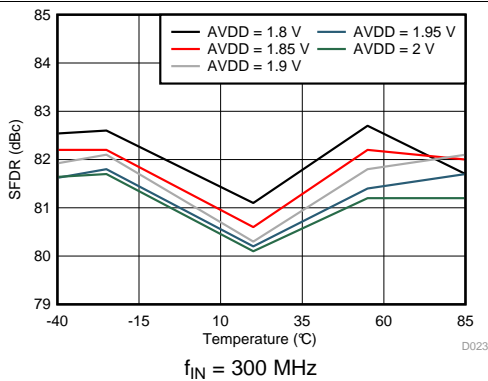


Figure 25. Spurious-Free Dynamic Range vs AVDD Supply and Temperature

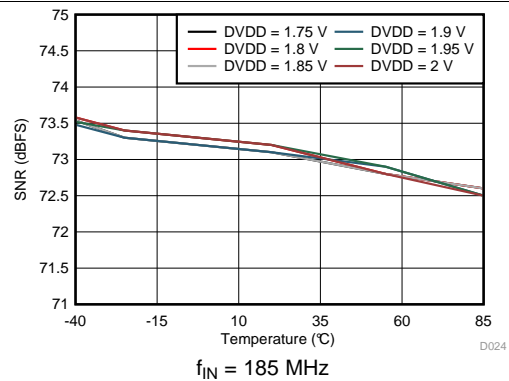


Figure 26. Signal-to-Noise Ratio vs DVDD Supply and Temperature

Typical Characteristics (continued)

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and -1-dBFS differential input (unless otherwise noted)

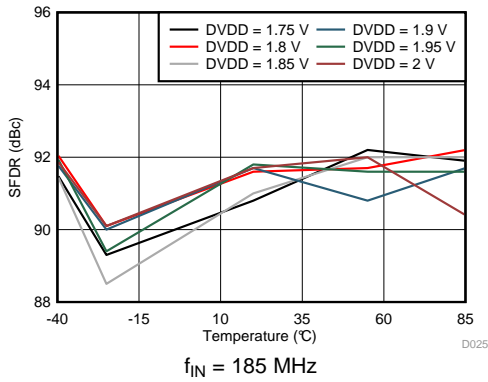


Figure 27. Spurious-Free Dynamic Range vs DVDD Supply and Temperature

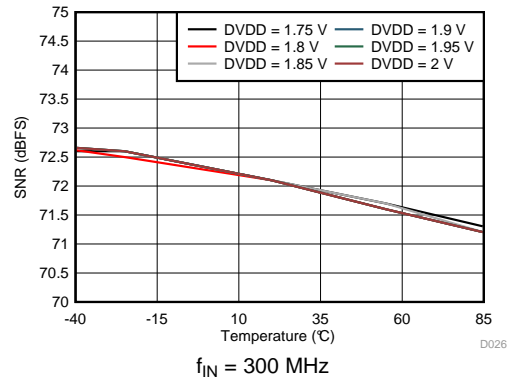


Figure 28. Signal-to-Noise Ratio vs DVDD Supply and Temperature

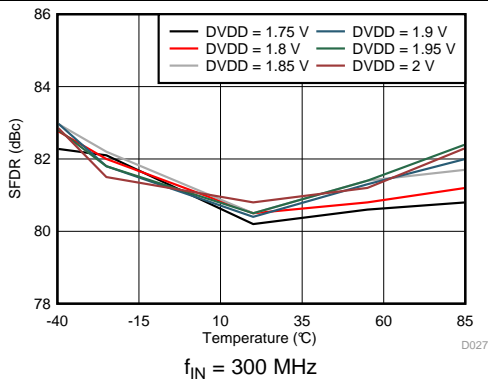


Figure 29. Spurious-Free Dynamic Range vs DVDD Supply and Temperature

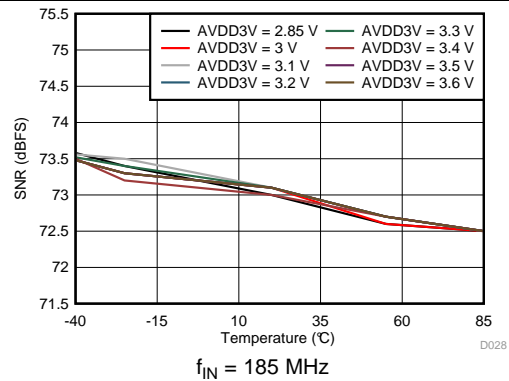


Figure 30. Signal-to-Noise Ratio vs AVDD3V Supply and Temperature

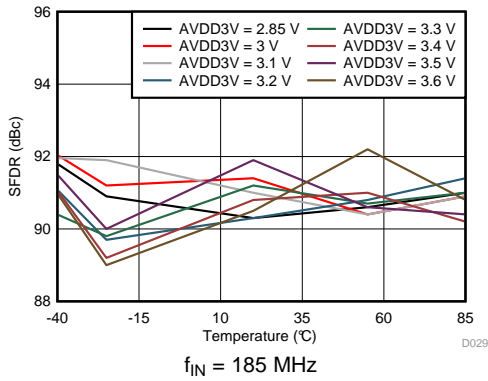


Figure 31. Spurious-Free Dynamic Range vs AVDD3V Supply and Temperature

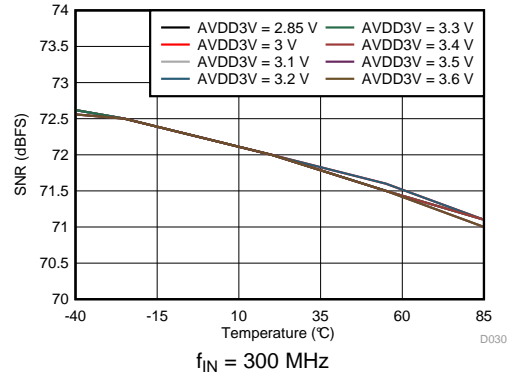


Figure 32. Signal-to-Noise Ratio vs AVDD3V Supply and Temperature



Typical Characteristics (continued)

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and -1-dBFS differential input (unless otherwise noted)

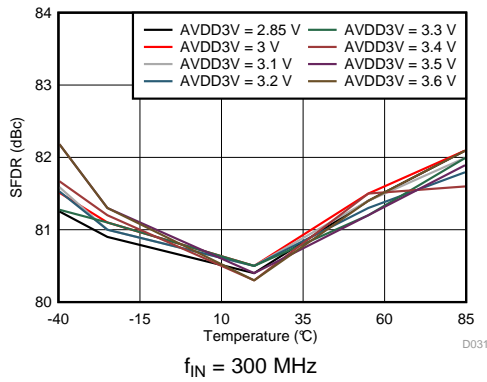


Figure 33. Spurious-Free Dynamic Range vs AVDD3V Supply and Temperature

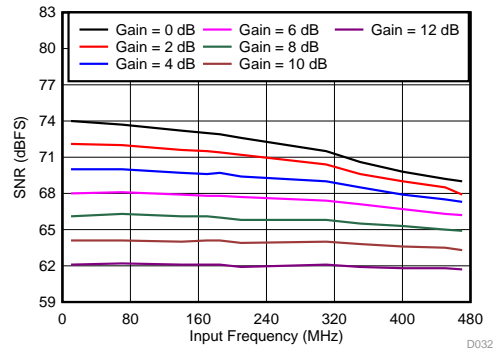


Figure 34. Signal-to-Noise Ratio vs Gain and Input Frequency

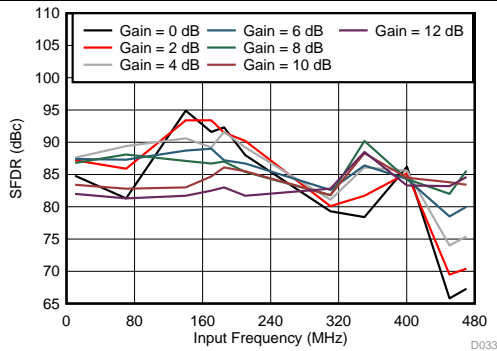


Figure 35. Spurious-Free Dynamic Range vs Gain and Input Frequency

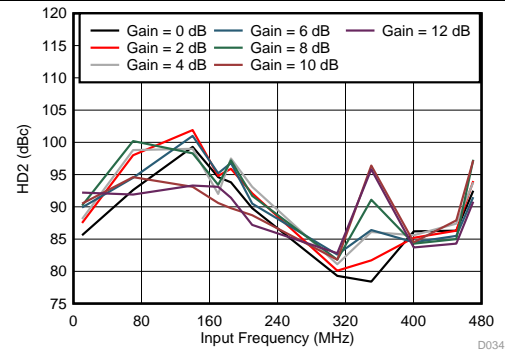


Figure 36. Second-Order Harmonic Distortion vs Gain and Input Frequency

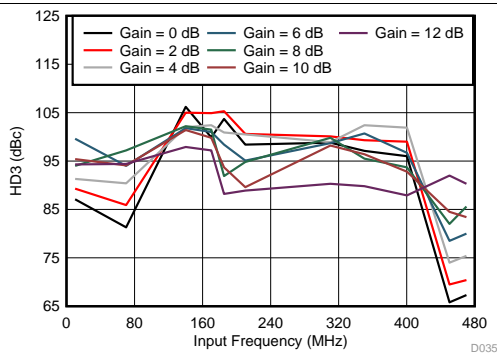


Figure 37. Third-Order Harmonic Distortion vs Gain and Input Frequency

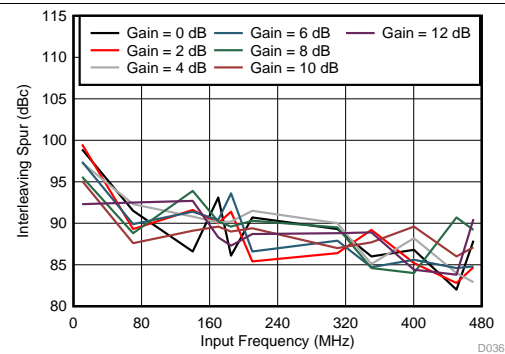
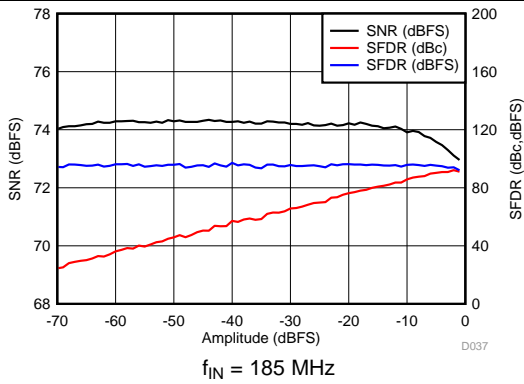


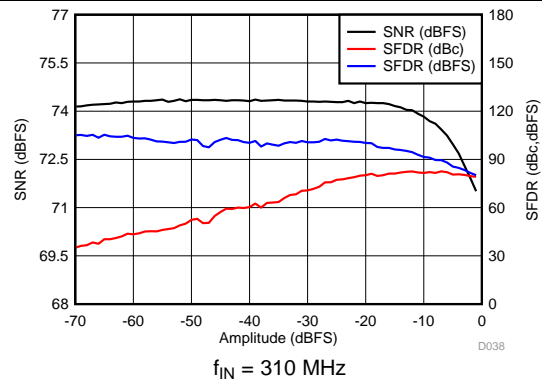
Figure 38. IL Spur vs Gain and Input Frequency

**Typical Characteristics (continued)**

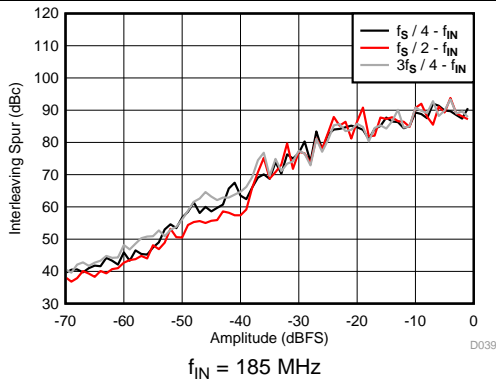
typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and  $-1\text{-dBFS}$  differential input (unless otherwise noted)



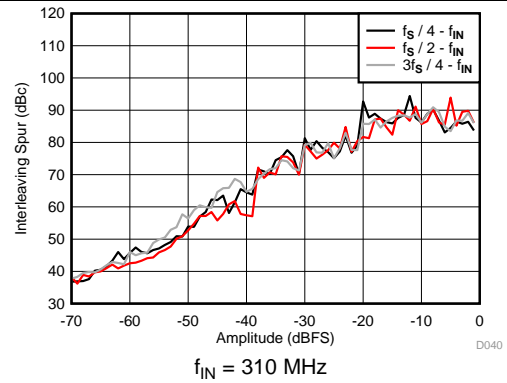
**Figure 39. Performance vs Amplitude**



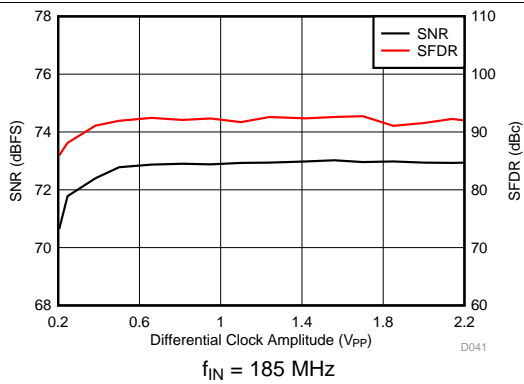
**Figure 40. Performance vs Amplitude**



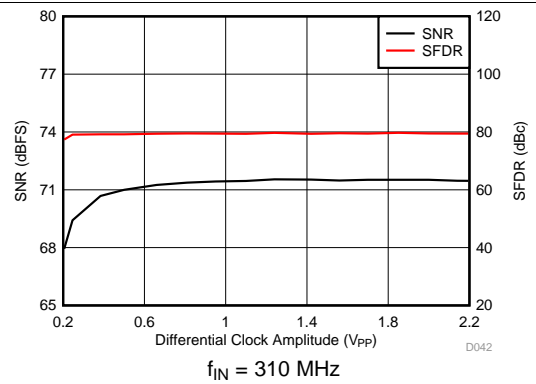
**Figure 41. IL Spur vs Amplitude**



**Figure 42. IL Spur vs Amplitude**



**Figure 43. Performance vs Differential Clock Amplitude**



**Figure 44. Performance vs Differential Clock Amplitude**

Typical Characteristics (continued)

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

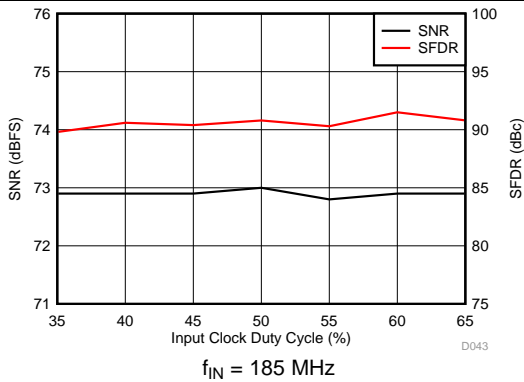


Figure 45. Performance vs Input Clock Duty Cycle

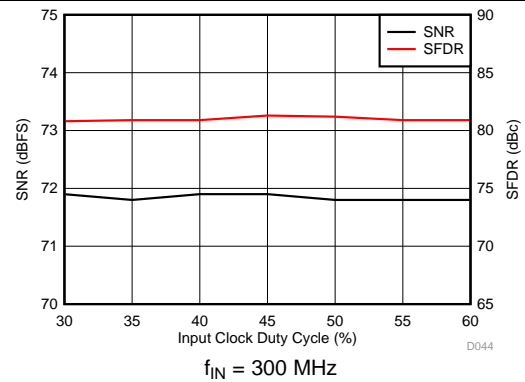


Figure 46. Performance vs Input Clock Duty Cycle

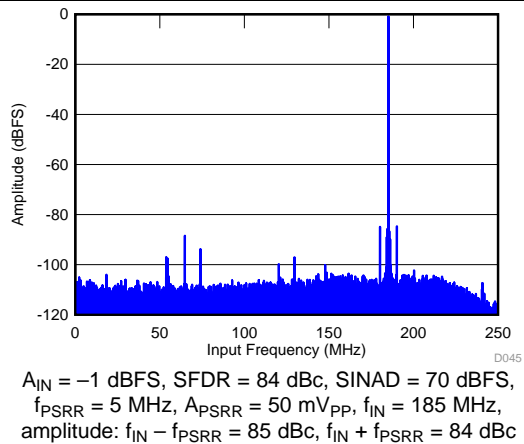


Figure 47. Power-Supply Rejection Ratio FFT for AVDD Supply

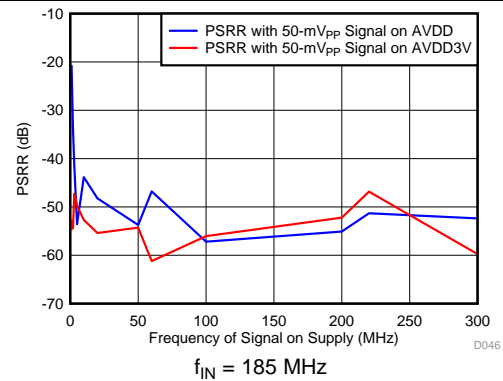


Figure 48. Power-Supply Rejection Ratio vs Noise Signal Frequency

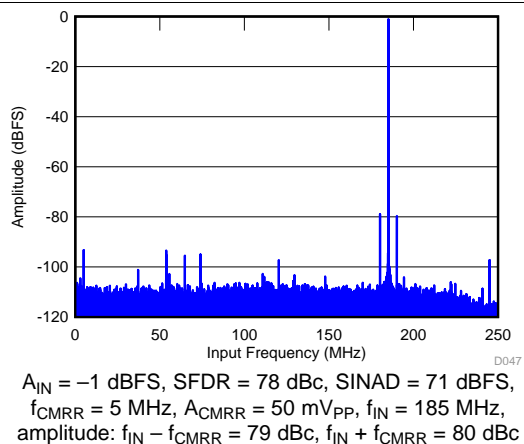


Figure 49. Common-Mode Rejection Ratio FFT

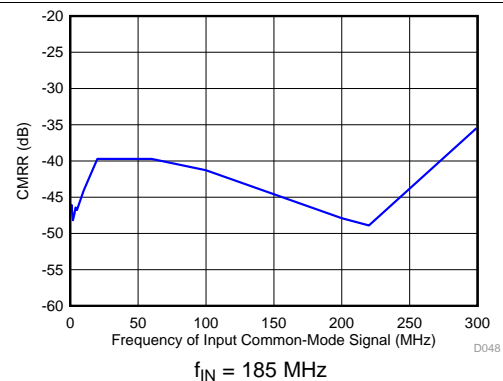
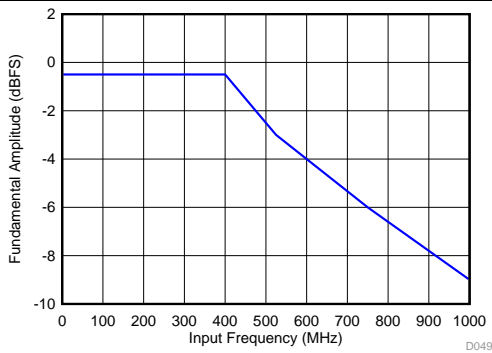


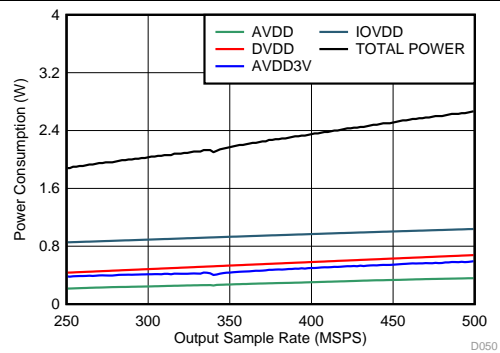
Figure 50. Common-Mode Rejection Ratio vs Common-Mode Signal Frequency

**Typical Characteristics (continued)**

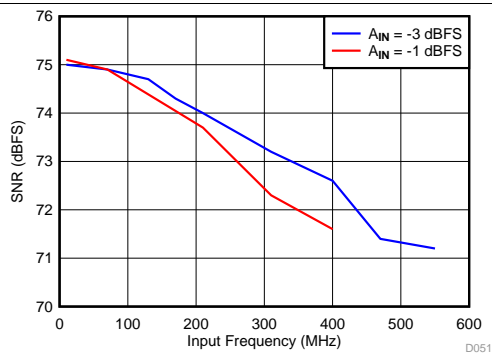
typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle, AVDD3V = 3.0 V, AVDD = DVDD = 1.9 V, IOVDD = 1.15 V, and -1-dBFS differential input (unless otherwise noted)



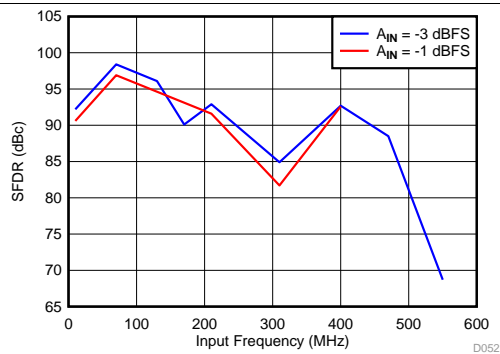
**Figure 51. Maximum-Supported Amplitude vs Input Frequency**



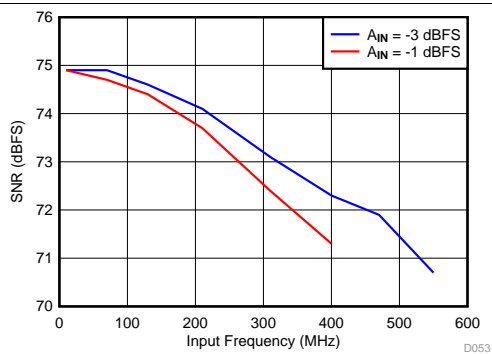
**Figure 52. Power Consumption vs Sampling Speed**



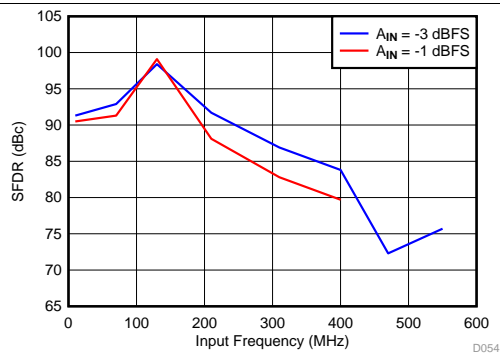
**Figure 53. Signal-to-Noise Ratio vs Input Frequency (Output Sample Rate = 300 MSPS)**



**Figure 54. Spurious-Free Dynamic Range vs Input Frequency (Output Sample Rate = 300 MSPS)**



**Figure 55. Signal-to-Noise Ratio vs Input Frequency (Output Sample Rate = 350 MSPS)**



**Figure 56. Spurious-Free Dynamic Range vs Input Frequency (Output Sample Rate = 350 MSPS)**

### Typical Characteristics (continued)

typical values are at  $T_A = 25^\circ\text{C}$ , full temperature range is from  $T_{\text{MIN}} = -40^\circ\text{C}$  to  $T_{\text{MAX}} = 85^\circ\text{C}$ , device clock frequency = 1 GSPS, output sampling rate = 500 MSPS, 50% clock duty cycle,  $AVDD3V = 3.0\text{ V}$ ,  $AVDD = DVDD = 1.9\text{ V}$ ,  $IOVDD = 1.15\text{ V}$ , and  $-1\text{-dBFS}$  differential input (unless otherwise noted)

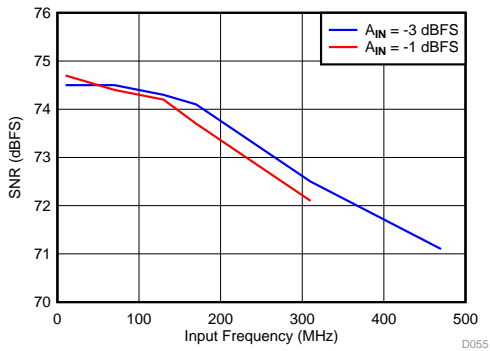


Figure 57. Signal-to-Noise Ratio vs Input Frequency (Output Sample Rate = 400 MSPS)

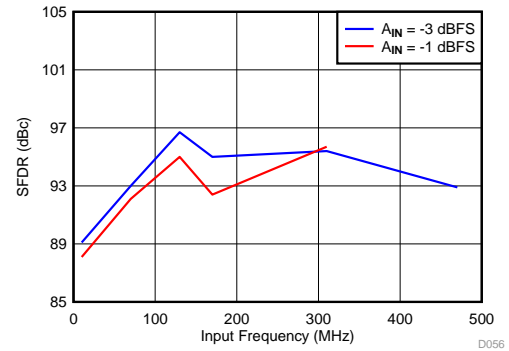


Figure 58. Spurious-Free Dynamic Range vs Input Frequency (Output Sample Rate = 400 MSPS)

## 7 Detailed Description

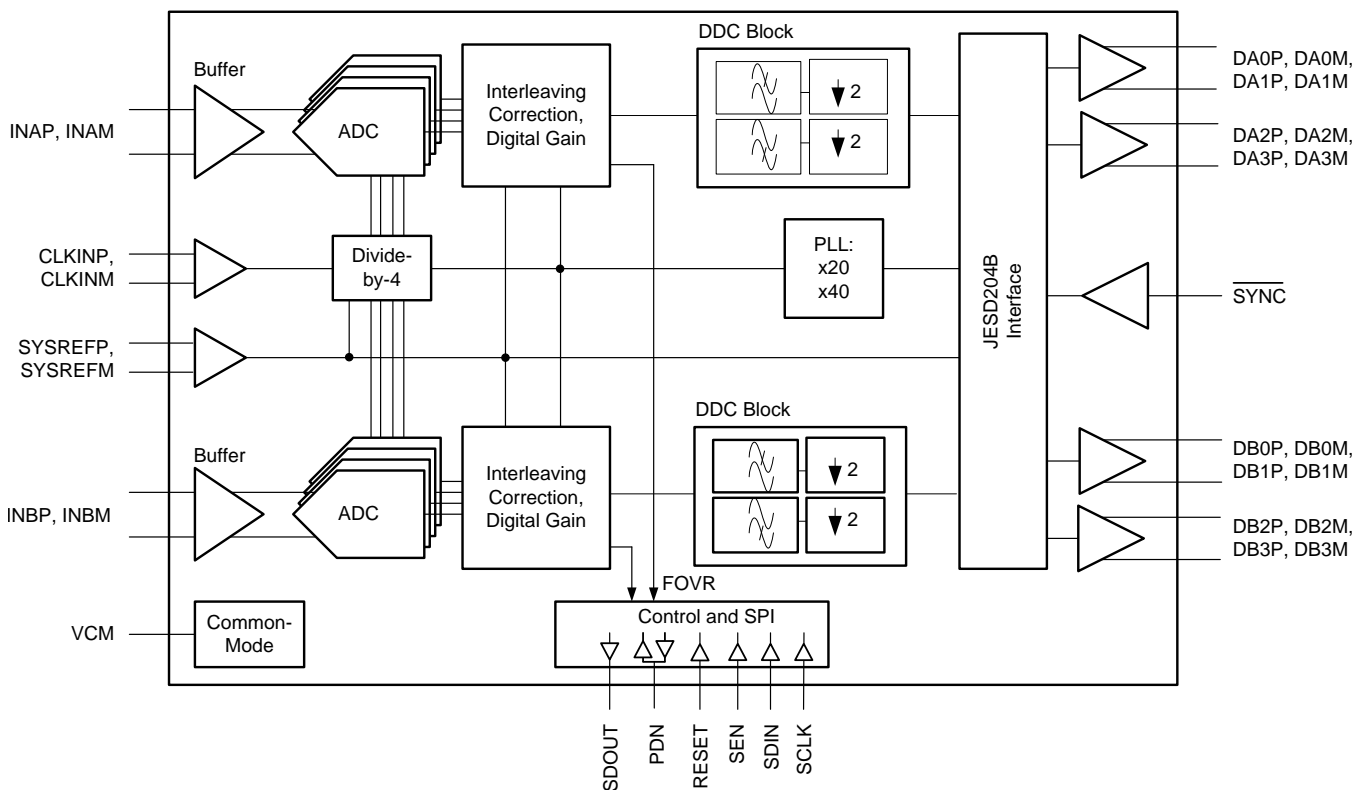
### 7.1 Overview

The ADS54J69 is a low-power, wide-bandwidth, 16-bit, 500-MSPS, dual-channel, analog-to-digital converter (ADC). The ADS54J69 employs four interleaving ADCs for each channel to achieve a noise floor of  $-159$  dBFS/Hz.

The ADS54J69 uses TI's proprietary interleaving and dither algorithms to achieve a clean spectrum with high spurious-free dynamic range (SFDR). Built-in, half-band, decimate-by-2 filters further enhance the capability of the ADS54J69 to deliver excellent signal-to-noise ratio (SNR) and SFDR over a wide range of frequencies. Analog input buffers isolate the ADC driver from glitch energy generated from sampling process, thereby simplify the driving network on-board.

The JESD204B interface reduces the number of interface lines with two-lane and four-lane options, allowing a high system integration density. The JESD204B interface operates in subclass-1, enabling multi-chip synchronization with the SYSREF input.

### 7.2 Functional Block Diagram



## 7.3 Feature Description

### 7.3.1 Analog Inputs

The ADS54J69 analog signal inputs are designed to be driven differentially. The analog input pins have internal analog buffers that drive the sampling circuit. As a result of the analog buffer, the input pins present a high impedance input across a very wide frequency range to the external driving source that enables great flexibility in the external analog filter design as well as excellent 50-Ω matching for RF applications. The buffer also helps isolate the external driving circuit from the internal switching currents of the sampling circuit, resulting in a more constant SFDR performance across input frequencies.

The common-mode voltage of the signal inputs is internally biased to VCM using 600-Ω resistors, allowing for ac-coupling of the input drive network. Each input pin (INP, INM) must swing symmetrically between (VCM + 0.475 V) and (VCM – 0.475 V), resulting in a 1.9-V<sub>PP</sub> (default) differential input swing. The input sampling circuit has a 3-dB bandwidth that extends up to 1.2 GHz. An equivalent analog input network diagram is shown in Figure 59.

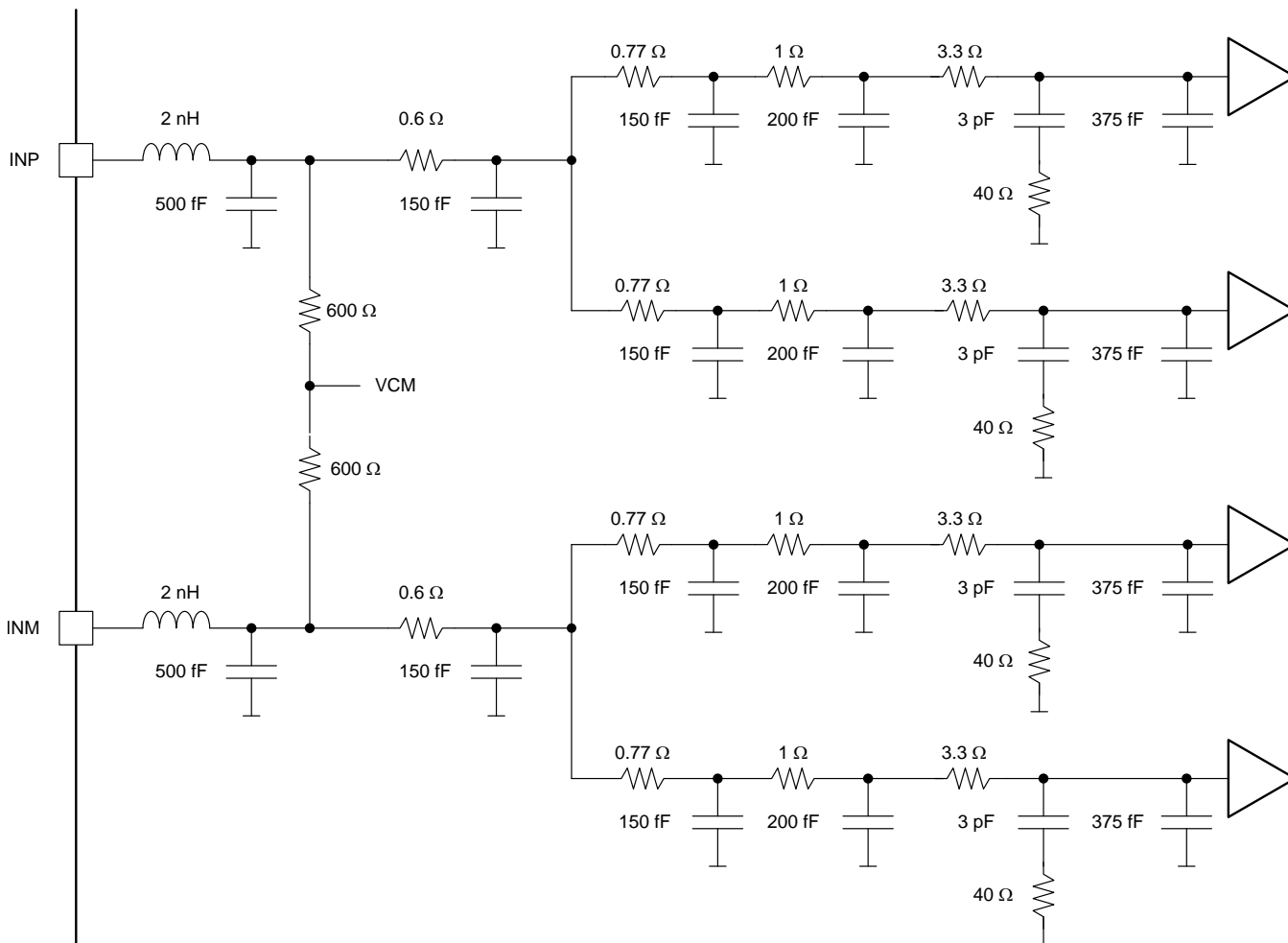


Figure 59. Analog Input Network

## Feature Description (continued)

The input bandwidth shown in Figure 60 is measured with respect to a 50-Ω differential input termination at the ADC input pins.

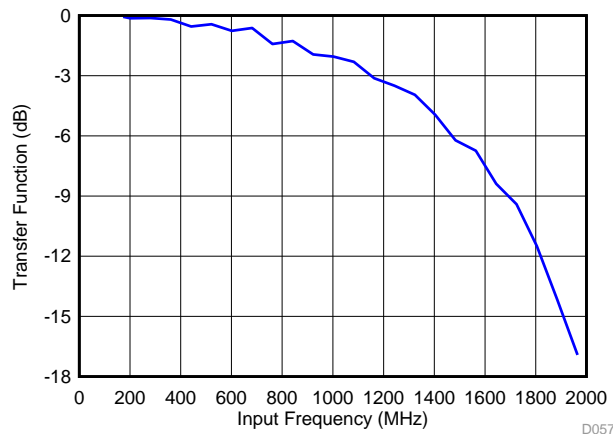


Figure 60. Transfer Function versus Frequency

### 7.3.2 DDC Block

The ADS54J69 has an optional DDC block that can be enabled via an SPI register write. Each ADC channel is followed by a DDC block consisting of a decimate-by-2, half-band, finite impulse response (FIR) filter with low-pass and high-pass options programmable via the SPI interface.

#### 7.3.2.1 Decimate-by-2 Filter

This decimation filter has 41 taps. The stop-band attenuation is approximately 90 dB and the pass-band flatness is ±0.05 dB. Table 1 shows corner frequencies for the low-pass and high-pass filter options.

Table 1. Corner Frequencies for the Decimate-by-2 Filter

CORNERS (dB)	LOW PASS	HIGH PASS
-0.1	$0.202 \times f_S$	$0.298 \times f_S$
-0.5	$0.210 \times f_S$	$0.290 \times f_S$
-1	$0.215 \times f_S$	$0.285 \times f_S$
-3	$0.227 \times f_S$	$0.273 \times f_S$

Figure 61 and Figure 62 show the frequency response of the decimate-by-2 filter from dc to  $f_S / 2$ .

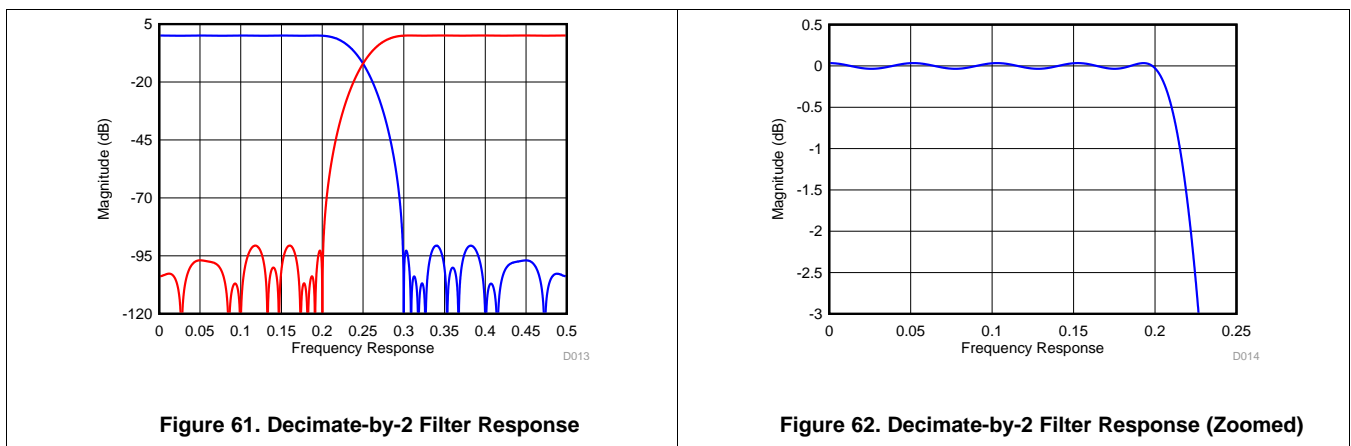


Figure 61. Decimate-by-2 Filter Response

Figure 62. Decimate-by-2 Filter Response (Zoomed)



### 7.3.3 SYSREF Signal

The SYSREF signal is a periodic signal that is sampled by the ADS54J69 device clock and used to align the boundary of the local multi-frame clock inside the data converter. SYSREF is required to be a sub-harmonic of the local multi-frame clock (LMFC) internal timing. To meet this requirement, the timing of SYSREF is dependent on the device clock frequency and the LMFC frequency, as determined by the selected DDC decimation and frames per multi-frame settings. The SYSREF signal is recommended be a low-frequency signal in the range of 1 MHz to 5 MHz in order to reduce coupling to the signal path both on the printed circuit board (PCB) as well as internal in the device.

The external SYSREF signal must be a sub-harmonic of the internal LMFC clock, as shown in Equation 1 and Table 2.

$$\text{SYSREF} = \text{LMFC} / 2^N$$

where

- N = 0, 1, 2, and so forth (1)

**Table 2. LMFC Clock Frequency**

LMFS CONFIGURATION	DECIMATION	LMFC CLOCK <sup>(1)(2)</sup>
4222	2X	$(f_s / 4) / k$
2242	2X	$(f_s / 4) / k$

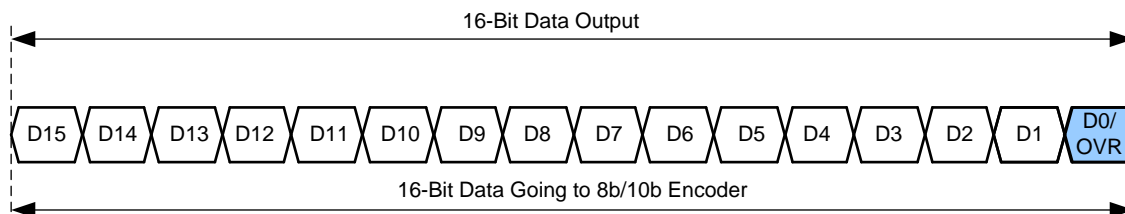
(1) K = Number of frames per multi-frame (JESD digital page 6900h, address 06h, bits 4-0).

(2)  $f_s$  = sampling (device) clock frequency.

### 7.3.4 Overrange Indication

The ADS54J69 provides a fast overrange indication that can be presented in the digital output data stream via an SPI configuration. Alternatively, if not used, the SDOOUT (pin 11) and PDN (pin 50) pins can be configured via the SPI to output the fast overrange (FOVR) indicator.

When the FOVR indication is embedded in the output data stream, the FOVR replaces the LSB of the 16-bit data stream going to the 8b/10b encoder, as shown in Figure 63.

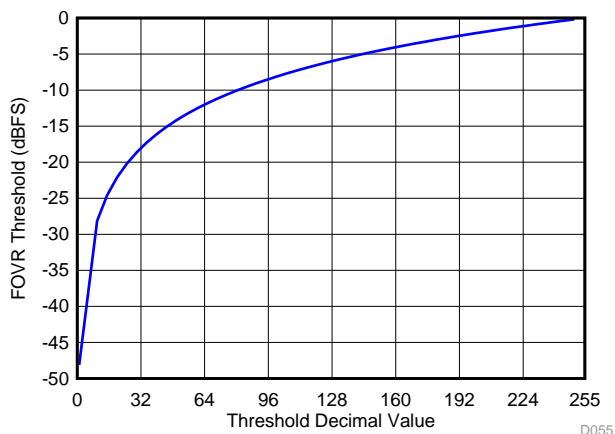


**Figure 63. Overrange Indication in a Data Stream**

### 7.3.4.1 Fast OVR

The fast OVR is triggered if the input voltage exceeds the programmable overrange threshold and is presented after only seven clock cycles, thus enabling a quicker reaction to an overrange event.

The input voltage level at which the overload is detected is referred to as the *threshold*. The threshold is programmable using the FOVR THRESHOLD bits, as shown in [Figure 64](#). The FOVR is triggered seven output clock cycles after the overload condition occurs.



**Figure 64. Programming Fast OVR Thresholds**

The input voltage level at which the fast OVR is triggered is defined by [Equation 2](#):

$$\text{Full-Scale} \times [\text{Decimal Value of the FOVR Threshold Bits}] / 255 \quad (2)$$

The default threshold is E3h (227d), corresponding to a threshold of –1 dBFS.

In terms of full-scale input, the fast OVR threshold can be calculated as [Equation 3](#):

$$20\log(\text{FOVR Threshold} / 255) \quad (3)$$

### 7.3.5 Power-Down Mode

The ADS54J69 provides a highly-configurable power-down mode. Power-down can be enabled by using the PDN pin or via SPI register writes.

A power-down mask can be configured that allows a trade-off between wake-up time and power consumption in power-down mode. Two independent power-down masks can be configured: MASK 1 and MASK 2, as shown in [Table 3](#). See the master page registers in [Table 9](#) for further details.

**Table 3. Register Address for Power-Down Modes**

REGISTER ADDRESS A[7:0] (Hex)	COMMENT	REGISTER DATA							
		7	6	5	4	3	2	1	0
<b>MASTER PAGE (80h)</b>									
20	MASK 1	PDN ADC CHA				PDN ADC CHB			
21		PDN BUFFER CHB	PDN BUFFER CHA	0	0	0	0		
23	MASK 2	PDN ADC CHA				PDN ADC CHB			
24		PDN BUFFER CHB	PDN BUFFER CHA	0	0	0	0		
26	CONFIG	GLOBAL PDN	OVERRIDE PDN PIN	PDN MASK SEL	0	0	0	0	0
53		0	MASK SYSREF	0	0	0	0	0	0
55		0	0	0	PDN MASK	0	0	0	0

To save power, the device can be put in complete power-down by using the GLOBAL PDN register bit. However, when the JESD link is required to be active during power-down, the ADC and analog buffer can be selectively powered down by using the PDN ADC CHx and PDN BUFFER CHx register bits after enabling the PDN MASK register bit. The PDN MASK SEL register bit can be used to select between MASK 1 or MASK 2. [Table 4](#) shows power consumption for different combinations of the GLOBAL PDN, PDN ADC CHx, and PDN BUFF CHx register bits.

**Table 4. Power Consumption in Different Power-Down Settings**

REGISTER BIT	COMMENT	I <sub>AVDD3V</sub> (mA)	I <sub>AVDD</sub> (mA)	I <sub>DVDD</sub> (mA)	I <sub>IOVDD</sub> (mA)	TOTAL POWER (W)
Default	After reset, with a full-scale input signal to both channels	346	354	188	512	2.66
GBL PDN = 1	The device is in a complete power-down state	3	6	21	127	0.2
GBL PDN = 0, PDN ADC CHx = 1 (x = A or B)	The ADC of one channel is powered down	284	221	130	461	2.05
GBL PDN = 0, PDN BUFF CHx = 1 (x = A or B)	The input buffer of one channel is powered down	270	352	188	516	2.43
GBL PDN = 0, PDN ADC CHx = 1, PDN BUFF CHx = 1 (x = A or B)	The ADC and input buffer of one channel are powered down	206	220	129	465	1.82
GBL PDN = 0, PDN ADC CHx = 1, PDN BUFF CHx = 1 (x = A and B)	The ADC and input buffer of both channels are powered down	64	84	67	389	0.93

## 7.4 Device Functional Modes

### 7.4.1 Device Configuration

The ADS54J69 can be configured by using a serial programming interface, as described in the [Serial Interface](#) section. In addition, the device has one dedicated parallel pin (PDN) for controlling the power-down mode.

The ADS54J69 supports a 24-bit (16-bit address, 8-bit data) SPI operation and uses paging (see the [Register Maps](#) section) to access all register bits.

#### 7.4.1.1 Serial Interface

The ADC has a set of internal registers that can be accessed by the serial interface formed by the SEN (serial interface enable), SCLK (serial interface clock), and SDIN (serial interface data) pins, as shown in [Figure 65](#). Legends used in [Figure 65](#) are explained in [Table 5](#). Serially shifting bits into the device is enabled when SEN is low. Serial data on SDIN are latched at every SCLK rising edge when SEN is active (low). The interface can function with SCLK frequencies from 2 MHz down to very low speeds (of a few hertz) and also with a non-50% SCLK duty cycle.

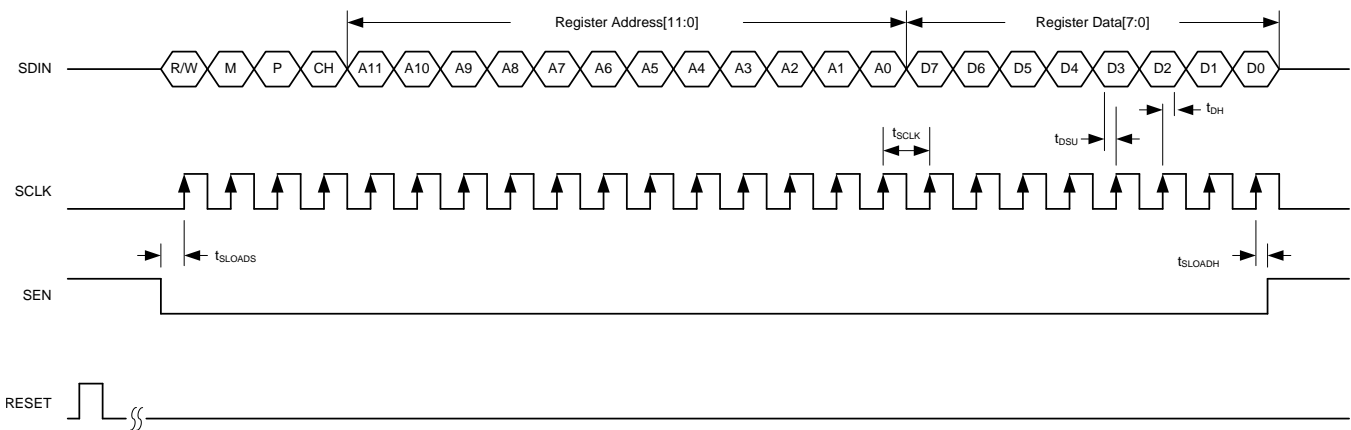


Figure 65. SPI Timing Diagram

Table 5. SPI Timing Diagram Legend

SPI BITS	DESCRIPTION	BIT SETTINGS
R/W	Read/write bit	0 = SPI write 1 = SPI read back
M	SPI bank access	0 = Analog SPI bank (master and ADC pages) 1 = JESD SPI bank (main digital, analog JESD, and digital JESD pages)
P	JESD page selection bit	0 = Page access 1 = Register access
CH	SPI access for a specific channel of the JESD SPI bank	0 = Channel A 1 = Channel B By default, both channels are being addressed.
A[11:0]	SPI address bits	—
D[7:0]	SPI data bits	—

Table 6 shows the timing requirements for the serial interface signals in Figure 65.

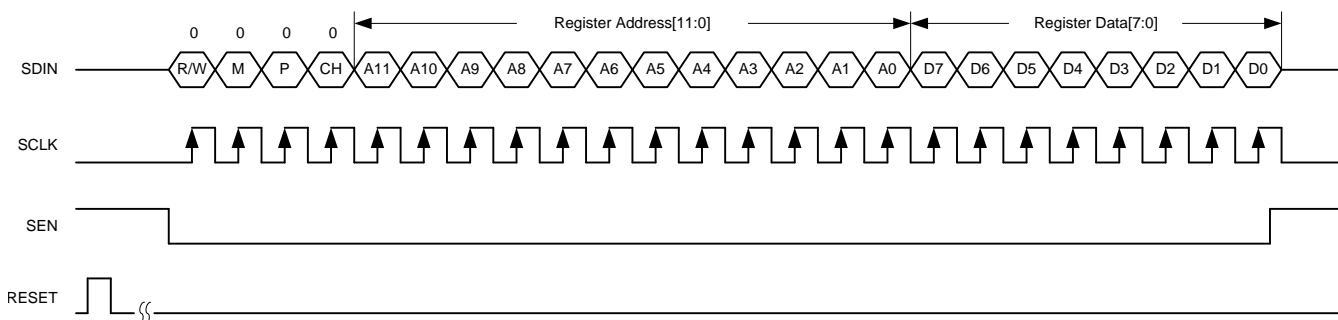
**Table 6. SPI Timing Requirements**

		MIN	TYP	MAX	UNIT
f <sub>SCLK</sub>	SCLK frequency (equal to 1 / t <sub>SCLK</sub> )	> dc		2	MHz
t <sub>SLOADS</sub>	SEN to SCLK setup time	100			ns
t <sub>SLOADH</sub>	SCLK to SEN hold time	100			ns
t <sub>DSU</sub>	SDIN setup time	100			ns
t <sub>DH</sub>	SDIN hold time	100			ns

**7.4.1.2 Serial Register Write: Analog Bank**

The analog SPI bank contains two pages (the master and ADC page). The internal register of the ADS54J69 analog SPI bank can be programmed by:

1. Driving the SEN pin low.
2. Initiating a serial interface cycle specifying the page address of the register whose content must be written.
  - Master page: write address 0011h with 80h.
  - ADC page: write address 0011h with 0Fh.
3. Writing the register content, as shown in Figure 66. When a page is selected, multiple writes into the same page can be done.



**Figure 66. Serial Register Write Timing Diagram**

### 7.4.1.3 Serial Register Readout: Analog Bank

The content from one of the two analog banks can be read out by:

1. Driving the SEN pin low.
2. Selecting the page address of the register whose content must be read.
  - Master page: write address 0011h with 80h.
  - ADC page: write address 0011h with 0Fh.
3. Setting the R/W bit to 1 and writing the address to be read back.
4. Reading back the register content on the SDOUT pin, as shown in Figure 67. When a page is selected, multiple read backs from the same page can be done.

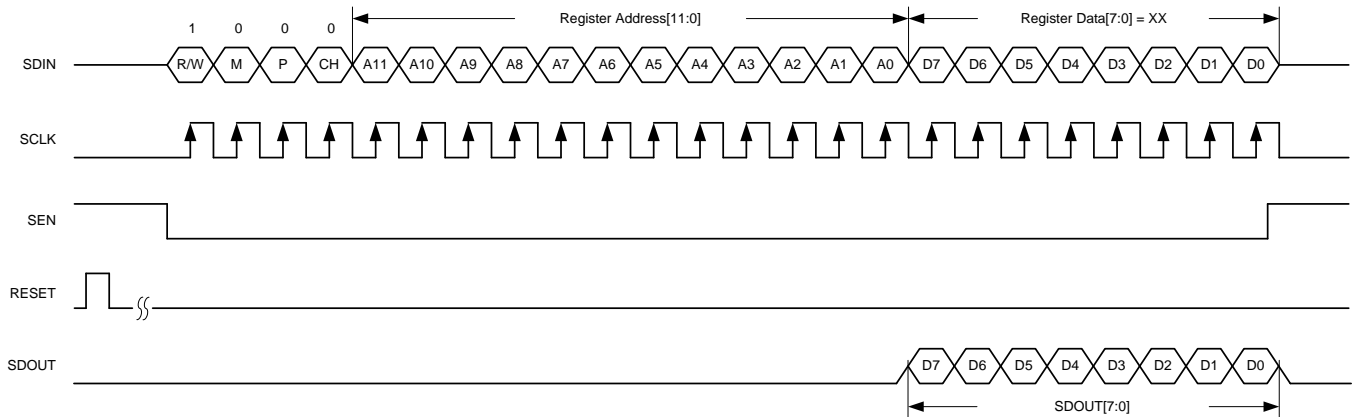


Figure 67. Serial Register Read Timing Diagram

### 7.4.1.4 JESD Bank SPI Page Selection

The JESD SPI bank contains four pages (main digital, digital, and analog JESD pages). The individual pages can be selected by:

1. Driving the SEN pin low.
2. Setting the M bit to 1 and specifying the page with two register writes. Note that the P bit must be set to 0, as shown in Figure 68.
  - Write address 4003h with 00h (LSB byte of page address).
  - Write address 4004h with the MSB byte of the page address.
    - For the main digital page: write address 4004h with 68h.
    - For the digital JESD page: write address 4004h with 69h.
    - For the analog JESD page: write address 4004h with 6Ah.

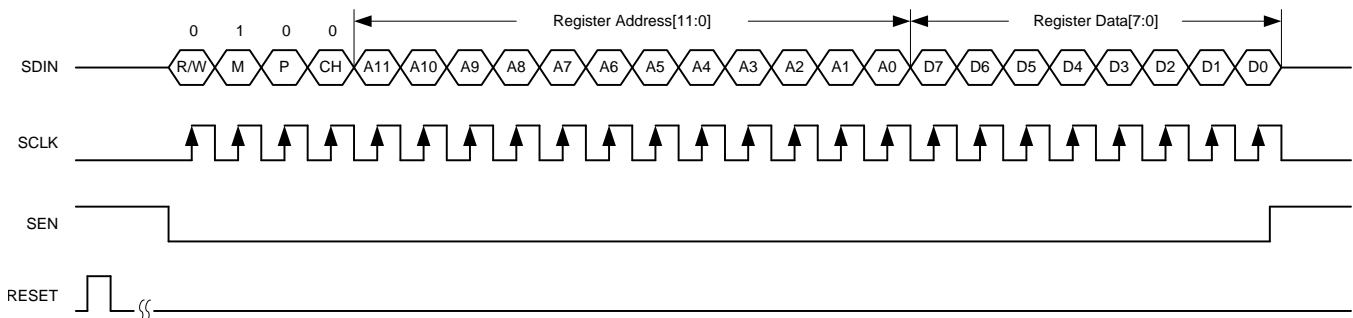


Figure 68. SPI Page Selection

### 7.4.1.5 Serial Register Write: JESD Bank

The ADS54J69 is a dual-channel device and the JESD204B portion is configured individually for each channel by using the CH bit. Note that the P bit must be set to 1 for register writes.

1. Drive the SEN pin low.
2. Select the JESD bank page. Note that the M bit = 1 and the P bit = 0.
  - Write address 4003h with 00h.
  - If separate control for both channels is desired, write address 4005h with 01h.
  - For the main digital page: write address 4004h with 68h.
  - For the digital JESD page: write address 4004h with 69h.
  - For the analog JESD page: write address 4004h with 6Ah.
3. Set the M and P bits to 1, select channel A (CH = 0) or channel B (CH = 1), and write the register content as shown in Figure 69. When a page is selected, multiple writes into the same page can be done.

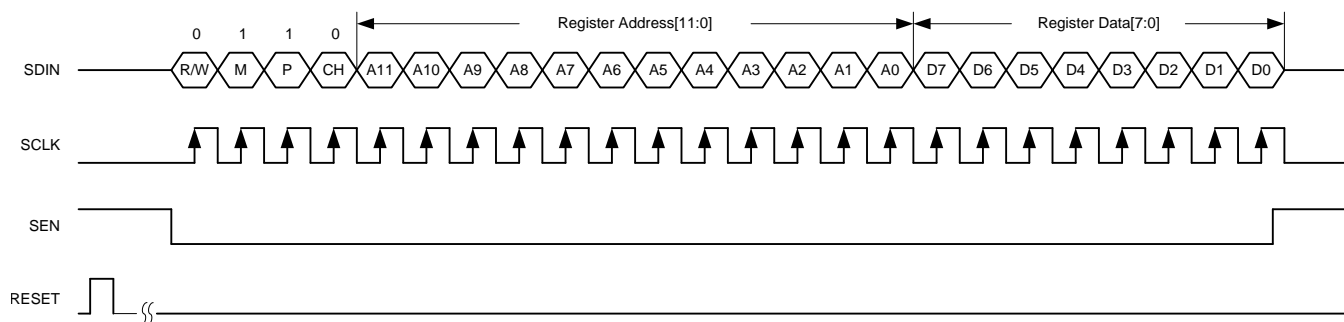


Figure 69. JESD Serial Register Write Timing Diagram

#### 7.4.1.5.1 Individual Channel Programming

By default, register writes are applied to both channels. To enable individual channel writes, write address 4005h with 01h (default is 00h).

### 7.4.1.6 Serial Register Readout: JESD Bank

The content from one of the pages of the JESD bank can be read out by:

1. Driving the SEN pin low.
2. Select the JESD bank page. Note that the M bit = 1 and the P bit = 0.
  - Write address 4003h with 00h.
  - If separate control for both channels is desired, write address 4005h with 01h.
  - For the main digital page: write address 4004h with 68h.
  - For the digital JESD page: write address 4004h with 69h.
  - For the analog JESD page: write address 4004h with 6Ah.
3. Setting the R/W, M, and P bits to 1, selecting channel A or channel B, and writing the address to be read back.
4. Reading back the register content on the SDO pin; see Figure 70. When a page is selected, multiple read backs from the same page can be done.

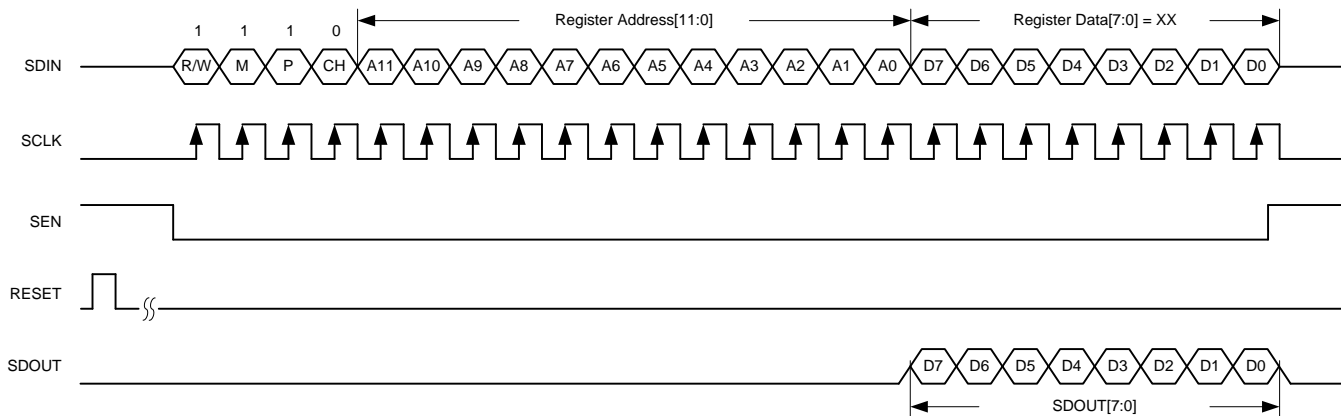


Figure 70. JESD Serial Register Read Timing Diagram

### 7.4.2 JESD204B Interface

The ADS54J69 supports device subclass 1 with a maximum output data rate of 10.0 Gbps for each serial transmitter.

An external SYSREF signal is used to align all internal clock phases and the local multi-frame clock to a specific sampling clock edge, allowing synchronization of multiple devices in a system and minimizing timing and alignment uncertainty. The SYNC input is used to control the JESD204B SERDES blocks.

Depending on the ADC output data rate, the JESD204B output interface can be operated with either two or four active lanes (out of total 8 lanes), as shown in Figure 71. The JESD204B setup and configuration of the frame assembly parameters is controlled via the SPI interface.

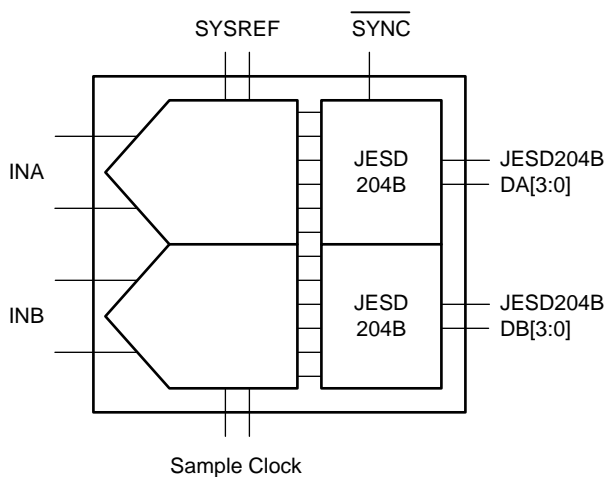


Figure 71. ADS54J69 Block Diagram



The JESD204B transmitter block shown in Figure 72 consists of the transport layer, the data scrambler, and the link layer. The transport layer maps the ADC output data into the selected JESD204B frame data format. The link layer performs the 8b/10b data encoding as well as the synchronization and initial lane alignment using the SYNC input signal. Optionally, data from the transport layer can be scrambled.

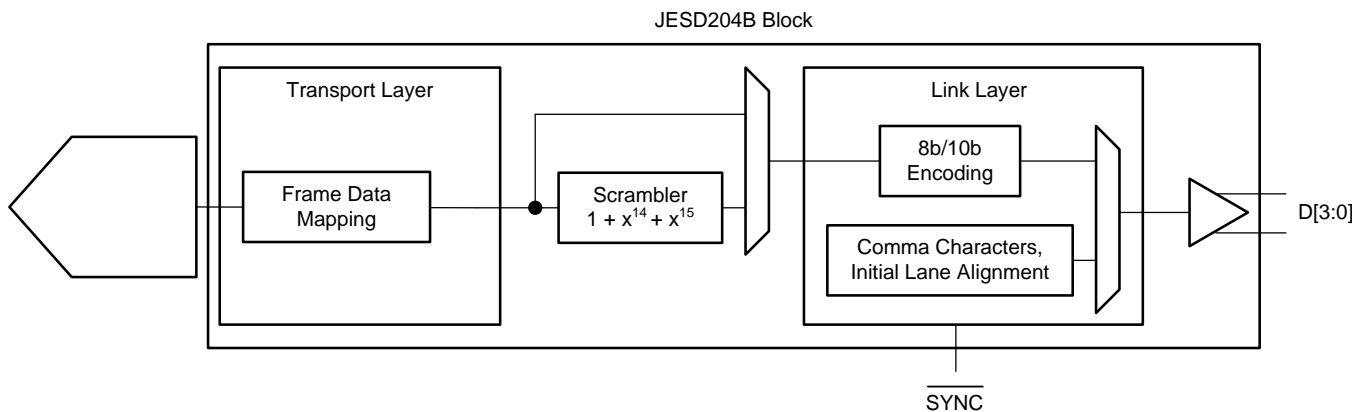


Figure 72. JESD204B Transmitter Block

7.4.2.1 JESD204B Initial Lane Alignment (ILA)

The initial lane alignment process is started when the receiving device de-asserts the SYNC signal, as shown in Figure 73. When a logic low is detected on the SYNC input pin, the ADS54J69 starts transmitting comma (K28.5) characters to establish a code group synchronization.

When synchronization is complete, the receiving device asserts the SYNC signal and the ADS54J69 starts the initial lane alignment sequence with the next local multi-frame clock boundary. The ADS54J69 transmits four multi-frames, each containing K frames (K is SPI programmable). Each of the multi-frames contains the frame start and end symbols and the second multi-frame also contains the JESD204 link configuration data.

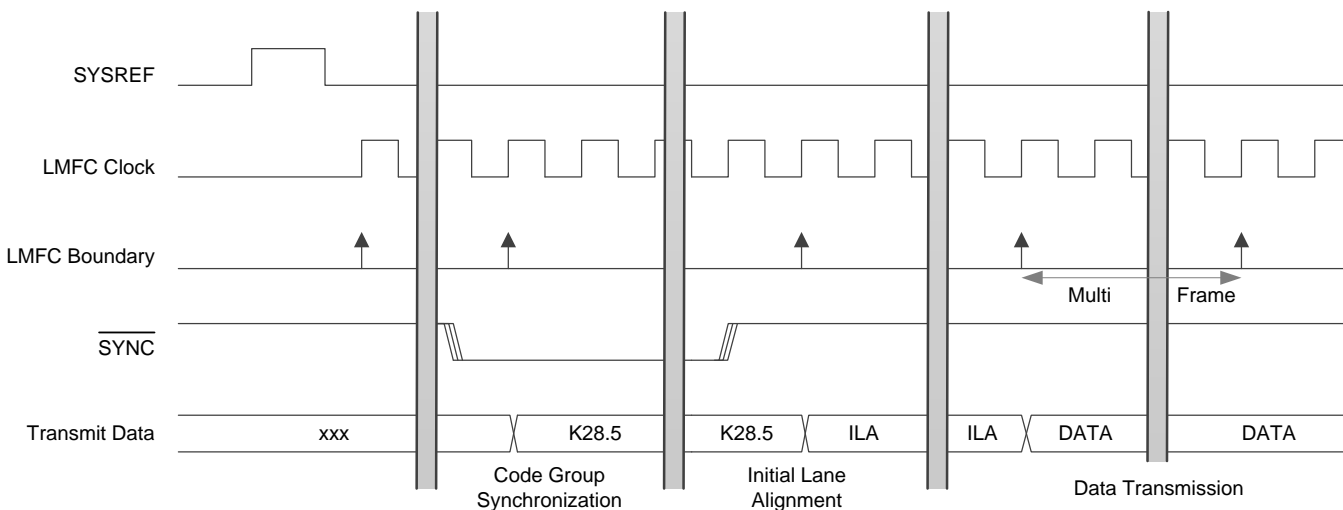


Figure 73. Lane Alignment Sequence

7.4.2.2 JESD204B Test Patterns

There are three different test patterns available in the transport layer of the JESD204B interface. The ADS54J69 supports a clock output, encoded, and a PRBS (2<sup>15</sup> – 1) pattern. These test patterns can be enabled via an SPI register write and are located in the JESD digital page of the JESD bank.

### 7.4.2.3 JESD204B Frame

The JESD204B standard defines the following parameters:

- L is the number of lanes per link
- M is the number of converters per device
- F is the number of octets per frame clock period, per lane
- S is the number of samples per frame per converter

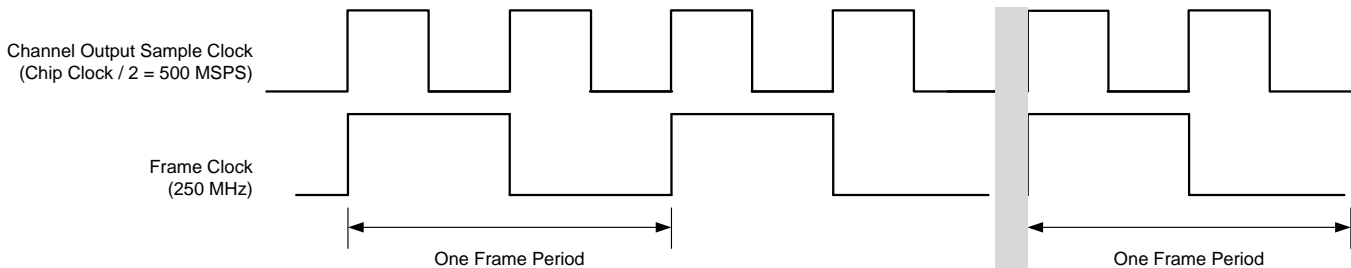
### 7.4.2.4 JESD204B Frame Assembly with Decimation

Table 7 lists the available JESD204B formats and interface rate at maximum sampling frequency. At lower sampling frequencies, interface rates scale down proportionally.

Figure 74 shows the detailed frame assembly for the decimated output.

**Table 7. Interface Rates with Decimation Filter**

L	M	F	S	JESD MODE REGISTER BIT	JESD PLL MODE SETTING	DECIMATION	MAX ADC OUTPUT RATE (MSPS)	MAX $f_{SERDES}$ (Gbps)
4	2	2	2	001	20x	2X	500	5.0
2	2	4	2	010	40x	2X	500	10.0



PIN	LMFS = 4222, 2X DECIMATION				LMFS = 2242, 2X DECIMATION			
DA0	A1[15:8]	A1[7:0]	A3[15:8]	A3[7:0]	Lane DA0 is Unused			
DA1	A0[15:8]	A0[7:0]	A2[15:8]	A2[7:0]	A0[15:8]	A0[7:0]	A1[15:8]	A1[7:0]
DA2	Lane DA2 is Unused				Lane DA2 is Unused			
DA3	Lane DA3 is Unused				Lane DA3 is Unused			
DB0	B1[15:8]	B1[7:0]	B3[15:8]	B3[7:0]	Lane DB0 is Unused			
DB1	B0[15:8]	B0[7:0]	B2[15:8]	B2[7:0]	B0[15:8]	B0[7:0]	B1[15:8]	B1[7:0]
DB2	Lane DB2 is Unused				Lane DB2 is Unused			
DB3	Lane DB3 is Unused				Lane DB3 is Unused			

**Figure 74. Frame Assembly with Decimation Filter**

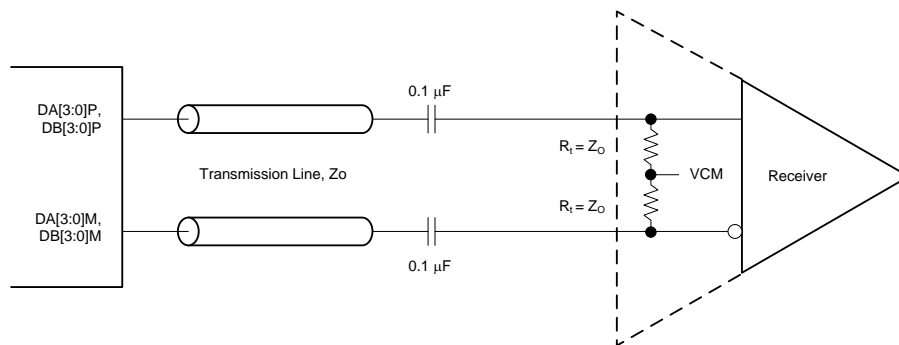
Note that after power-up, the JESD output bus must be reordered to obtain correct link parameters in the ILA sequence. [Table 8](#) shows the required register writes to reorder the JESD output bus.

**Table 8. Configuring LMFS for the JESD Link**

L	M	F	S	DECIMATION	JESD PLL MODE (In JESD Analog Page)	JESD MODE REGISTER BIT (In JESD Digital Page)	DA_BUS_REORDER REGISTER BIT (In JESD Digital Page)	DB_BUS_REORDER REGISTER BIT (In JESD Digital Page)	REGISTER 52 (In Main Digital Page)	REGISTER 72 (In Main Digital Page)
4	2	2	2	2X	00h	01h	0Ah	0Ah	80h	08h
2	2	4	2	2X	10h	02h	0Ah	0Ah	80h	08h

**7.4.2.4.1 JESD Transmitter Interface**

Each of the 10.0-Gbps SERDES JESD transmitter outputs requires ac-coupling between the transmitter and receiver. The differential pair must be terminated with 100-Ω resistors as close to the receiving device as possible to avoid unwanted reflections and signal degradation, as shown in [Figure 75](#).



**Figure 75. Output Connection to Receiver**

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7.4.2.4.2 Eye Diagrams

Figure 76 to Figure 79 show the serial output eye diagrams of the ADS54J69 at 5.0 Gbps and 10 Gbps with default and increased output voltage swing against the JESD204B mask.



Figure 76. Eye Diagram at 5-Gbps Bit Rate with Default Output Swing

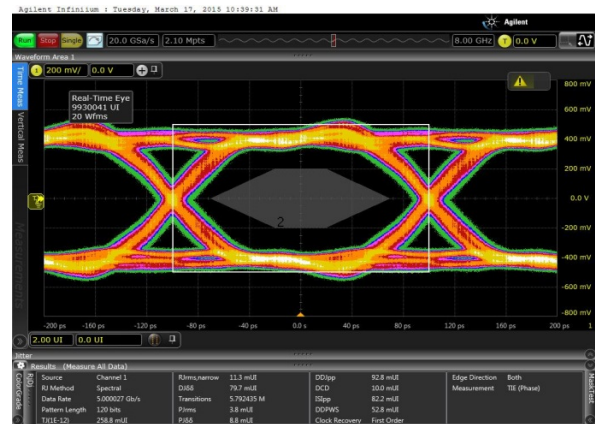


Figure 77. Eye Diagram at 5-Gbps Bit Rate with Increased Output Swing

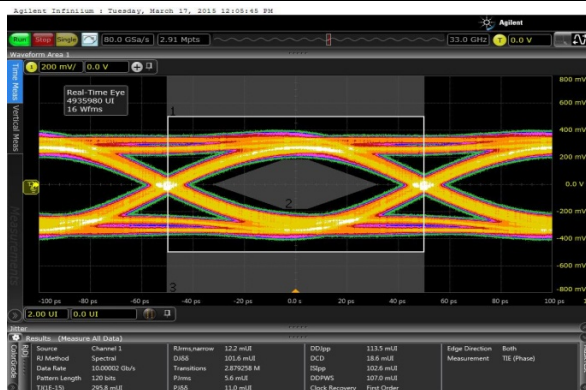


Figure 78. Eye Diagram at 10-Gbps Bit Rate with Default Output Swing

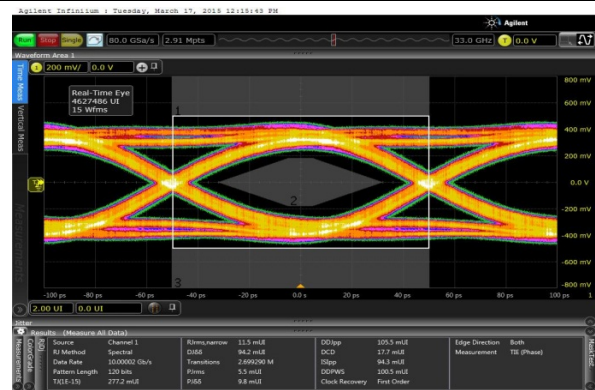


Figure 79. Eye Diagram at 10-Gbps Bit Rate with Increased Output Swing

## 7.5 Register Maps

Figure 80 shows a conceptual diagram of the serial registers.

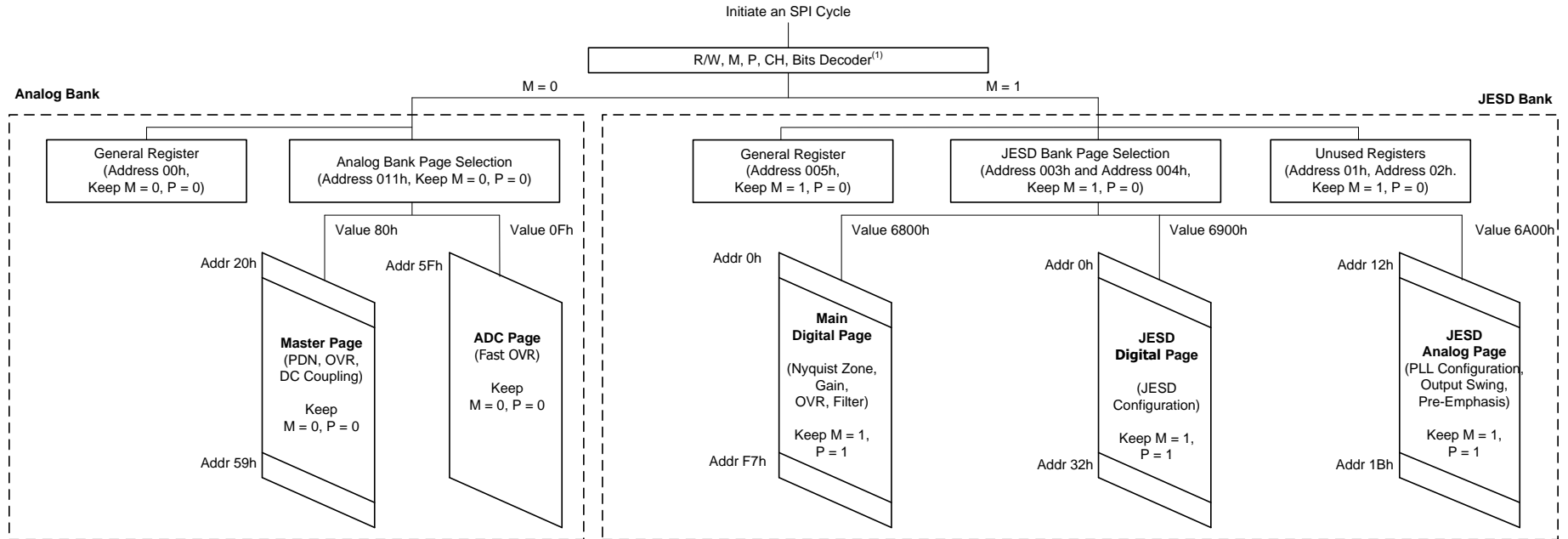


Figure 80. Serial Interface Registers

### 7.5.1 Detailed Register Info

The ADS54J69 contains two main SPI banks: the analog SPI bank and the digital SPI bank. The analog SPI bank gives access to the ADC analog blocks and the JESD SPI bank controls the digital features and anything related to the JESD204B serial interface. The analog SPI bank is divided into two pages (master and ADC) and the JESD SPI bank is divided into three pages (main digital, JESD digital, and JESD analog). Table 9 lists a register map for the ADS54J69.

**Register Maps (continued)**
**Table 9. Register Map**

REGISTER ADDRESS A[11:0] (Hex)	REGISTER DATA							
	7	6	5	4	3	2	1	0
<b>GENERAL REGISTERS</b>								
0	RESET	0	0	0	0	0	0	RESET
3	JESD BANK PAGE SEL[7:0]							
4	JESD BANK PAGE SEL[15:8]							
5	0	0	0	0	0	0	0	DISABLE BROADCAST
11	ANALOG BANK PAGE SEL							
<b>MASTER PAGE (80h)</b>								
20	PDN ADC CHA				PDN ADC CHB			
21	PDN BUFFER CHB		PDN BUFFER CHA		0	0	0	0
23	PDN ADC CHA				PDN ADC CHB			
24	PDN BUFFER CHB		PDN BUFFER CHA		0	0	0	0
26	GLOBAL PDN	OVERRIDE PDN PIN	PDN MASK SEL	0	0	0	0	0
39	PERF MODE[1:0]		0	0	0	0	0	0
3A	0	PERF MODE[2]	0	0	0	0	0	0
4F	0	0	0	0	0	0	0	EN INPUT DC COUPLING
53	0	MASK SYSREF	0	0	0	0	EN SYSREF DC COUPLING	0
55	0	0	0	PDN MASK	0	0	0	0
56	0	0	0	0	0	PERF MODE[3]	0	0
59	FOVR CHB	0	ALWAYS WRITE 1	0	0	0	0	0
<b>ADC PAGE (0Fh)</b>								
5F	FOVR THRESHOLD PROG							
<b>MAIN DIGITAL PAGE (6800h)</b>								
0	0	0	0	0	0	0	0	PULSE RESET
41	0	0	0	DECFIL MODE[3]	0	DECFIL MODE[2:0]		
42	0	0	0	0	0	NYQUIST ZONE		
43	0	0	0	0	0	0	0	FORMAT SEL
44	0	DIGITAL GAIN						
4B	0	0	FORMAT EN	0	0	0	0	0

**Register Maps (continued)**
**Table 9. Register Map (continued)**

REGISTER ADDRESS A[11:0] (Hex)	REGISTER DATA								
	7	6	5	4	3	2	1	0	
<b>MAIN DIGITAL PAGE (6800h) (continued)</b>									
4D	0	0	0	0	DEC MODE EN	0	0	0	
4E	CTRL NYQUIST	0	0	0	0	0	0	0	
52	ALWAYS WRITE 1	0	0	0	0	0	0	DIG GAIN EN	
72	0	0	0	0	ALWAYS WRITE 1	0	0	0	
AB	0	0	0	0	0	0	0	LSB SEL EN	
AD	0	0	0	0	0	0	LSB SELECT		
F7	0	0	0	0	0	0	0	DIG RESET	
<b>JESD DIGITAL PAGE (6900h)</b>									
0	CTRL K	0	0	TESTMODE EN	FLIP ADC DATA	LANE ALIGN	FRAME ALIGN	TX LINK DIS	
1	SYNC REG	SYNC REG EN	0	0	0	JESD MODE			
2	LINK LAYER TESTMODE			LINK LAYER RPAT	LMFC MASK RESET	0	0	0	
3	FORCE LMFC COUNT	LMFC COUNT INIT					RELEASE ILANE SEQ		
5	SCRAMBLE EN	0	0	0	0	0	0	0	
6	0	0	0	FRAMES PER MULTI FRAME (K)					
7	0	0	0	0	SUBCLASS	0	0	0	
31	DA_BUS_REORDER[7:0]								
32	DB_BUS_REORDER[7:0]								
<b>JESD ANALOG PAGE (6A00h)</b>									
12	SEL EMP LANE 1						0	0	
13	SEL EMP LANE 0						0	0	
14	SEL EMP LANE 2						0	0	
15	SEL EMP LANE 3						0	0	
16	0	0	0	0	0	0	JESD PLL MODE		
1A	0	0	0	0	0	0	FOVR CHA	0	
1B	JESD SWING			0	FOVR CHA EN	0	0	0	

## 7.5.2 Example Register Writes

This section provides three different example register writes. [Table 10](#) describes a global power-down register write, [Table 11](#) describes the register writes to enable the high-pass filter in the default four-lane output mode (LMFS = 4222), and [Table 12](#) describes the register writes to enable the high-pass filter in the two-lane output mode (LMFS = 2242).

Note that by default after reset, the low-pass filter and four-lane output mode are enabled and register writes are applied to both channels together.

**Table 10. Global Power-Down**

ADDRESS (Hex)	DATA (Hex)	COMMENT
11h	80h	Set the master page
26h	C0h	Set the global power-down

**Table 11. Selecting 2X Decimation with Four-Lane Mode (LMFS = 4222)**

ADDRESS (Hex)	DATA (Hex)	COMMENT
4-004h	68h	Select the main digital page (6800h)
4-003h	00h	
6-041h	16h	Set decimate-by-2 (high-pass filter)
6-04Dh	08h	Enable decimation filter control
6-072h	08h	Enable the ALWAYS WRITE 1 register bit (for output bus reorder)
6-052h	80h	Enable the ALWAYS WRITE 1 register bit (for output bus reorder)
6-000h	01h	Pulse the PULSE RESET register bit so that registers programmed in the main digital page (6800h) become effective.
6-000h	00h	
4-004h	69h	Select the JESD digital page (6900h)
4-003h	00h	
6-031h	0Ah	Output bus reorder for channel A
6-032h	0Ah	Output bus reorder for channel B
6-001h	01h	JESD filter mode + 4-lanes output selection

**Table 12. Selecting 2X Decimation with Two-Lane Mode (LMFS = 2242)**

ADDRESS (Hex)	DATA (Hex)	COMMENT
4-004h	68h	Select the main digital page (6800h)
4-003h	00h	
6-041h	16h	Set decimate-by-2 (high-pass filter)
6-04Dh	08h	Enable decimation filter control
6-072h	08h	Set the ALWAYS WRITE 1 register bit (for output bus reorder)
6-052h	80h	Set the ALWAYS WRITE 1 register bit (for output bus reorder)
6-000h	01h	Pulse the PULSE RESET register bit so that registers programmed in the main digital page (6800h) become effective.
6-000h	00h	
4-004h	69h	Select the JESD digital page (6900h)
4-003h	00h	
6-031h	0Ah	Output bus reorder for channel A
6-032h	0Ah	Output bus reorder for channel B
6-001h	02h	JESD filter mode + 2-lanes output selection
4-004h	6Ah	Select the JESD analog page (6A00h)
4-003h	00h	
6-016h	02h	JESD PLL MODE 40x selection in the analog page



## 7.5.3 Register Descriptions

### 7.5.3.1 General Registers

#### 7.5.3.1.1 Register 0h (address = 0h)

**Figure 81. Register 0h**

7	6	5	4	3	2	1	0
RESET	0	0	0	0	0	0	RESET
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: W = Write only; -n = value after reset

**Table 13. Register 0h Field Descriptions**

Bit	Field	Type	Reset	Description
7	RESET	W	0h	0 = Normal operation 1 = Internal software reset, clears back to 0
6-1	0	W	0h	Must write 0
0	RESET	W	0h	0 = Normal operation 1 = Internal software reset, clears back to 0

#### 7.5.3.1.2 Register 3h (address = 3h)

**Figure 82. Register 3h**

7	6	5	4	3	2	1	0
JESD BANK PAGE SEL[7:0]							
R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 14. Register 3h Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	JESD BANK PAGE SEL[7:0]	R/W	0h	Program these bits to access the desired page in the JESD bank. 6800h = Main digital page selected 6900h = JESD digital page selected 6A00h = JESD analog page selected

#### 7.5.3.1.3 Register 4h (address = 4h)

**Figure 83. Register 4h**

7	6	5	4	3	2	1	0
JESD BANK PAGE SEL[15:8]							
R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 15. Register 4h Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	JESD BANK PAGE SEL[15:8]	R/W	0h	Program these bits to access the desired page in the JESD bank. 6800h = Main digital page selected 6900h = JESD digital page selected 6A00h = JESD analog page selected

**7.5.3.1.4 Register 5h (address = 5h)**
**Figure 84. Register 5h**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	DISABLE BROADCAST
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 16. Register 5h Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	DISABLE BROADCAST	R/W	0h	0 = Normal operation; channel A and B are programmed as a pair 1 = Channel A and B can be individually programmed based on the CH bit (keep CH = 0 for channel A, CH = 1 for channel B).

**7.5.3.1.5 Register 11h (address = 11h)**
**Figure 85. Register 11h**

7	6	5	4	3	2	1	0
ANALOG PAGE SELECTION							
R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 17. Register 11h Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	ANALOG BANK PAGE SEL	R/W	0h	Program these bits to access the desired page in the analog bank. Master page = 80h ADC page = 0Fh

**7.5.3.2 Master Page (080h) Registers**
**7.5.3.2.1 Register 20h (address = 20h), Master Page (080h)**
**Figure 86. Register 20h**

7	6	5	4	3	2	1	0
PDN ADC CHA				PDN ADC CHB			
R/W-0h				R/W-0h			

LEGEND: R/W = Read/Write; -n = value after reset

**Table 18. Registers 20h Field Descriptions**

Bit	Field	Type	Reset	Description
7-4	PDN ADC CHA	R/W	0h	There are two power-down masks that are controlled via the PDN mask register bit in address 55h. Power-down mask 1 or mask 2 are selected via register bit 5 in address 26h. Power-down mask 1: addresses 20h and 21h Power-down mask 2: addresses 23h and 24h
3-0	PDN ADC CHB	R/W	0h	

**7.5.3.2.2 Register 21h (address = 21h), Master Page (080h)**
**Figure 87. Register 21h**

7	6	5	4	3	2	1	0
PDN BUFFER CHB		PDN BUFFER CHA		0	0	0	0
R/W-0h		R/W-0h		W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 19. Register 21h Field Descriptions**

Bit	Field	Type	Reset	Description
7-6	PDN BUFFER CHB	R/W	0h	There are two power-down masks that are controlled via the PDN mask register bit in address 55h. Power-down mask 1 or mask 2 are selected via register address 26h, bit 5. Power-down mask 1: addresses 20h and 21h Power-down mask 2: addresses 23h and 24h
5-4	PDN BUFFER CHA	R/W	0h	
3-0	0	W	0h	

**7.5.3.2.3 Register 23h (address = 23h), Master Page (080h)**
**Figure 88. Register 23h**

7	6	5	4	3	2	1	0
PDN ADC CHA				PDN ADC CHB			
R/W-0h				R/W-0h			

LEGEND: R/W = Read/Write; -n = value after reset

**Table 20. Register 23h Field Descriptions**

Bit	Field	Type	Reset	Description
7-4	PDN ADC CHA	R/W	0h	There are two power-down masks that are controlled via the PDN mask register bit in address 55h. Power-down mask 1 or mask 2 are selected via register address 26h, bit 5. Power-down mask 1: addresses 20h and 21h Power-down mask 2: addresses 23h and 24h
3-0	PDN ADC CHB	R/W	0h	

**7.5.3.2.4 Register 24h (address = 24h), Master Page (080h)**
**Figure 89. Register 24h**

7	6	5	4	3	2	1	0
PDN BUFFER CHB		PDN BUFFER CHA		0	0	0	0
R/W-0h		R/W-0h		W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 21. Register 24h Field Descriptions**

Bit	Field	Type	Reset	Description
7-6	PDN BUFFER CHB	R/W	0h	There are two power-down masks that are controlled via the PDN mask register bit in address 55h. Power-down mask 1 or mask 2 are selected via register address 26h, bit 5. Power-down mask 1: addresses 20h and 21h Power-down mask 2: addresses 23h and 24h
5-4	PDN BUFFER CHA	R/W	0h	
3-0	0	W	0h	

**7.5.3.2.5 Register 26h (address = 26h), Master Page (080h)**
**Figure 90. Register 26h**

7	6	5	4	3	2	1	0
GLOBAL PDN	OVERRIDE PDN PIN	PDN MASK SEL	0	0	0	0	0
R/W-0h	R/W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 22. Register 26h Field Descriptions**

Bit	Field	Type	Reset	Description
7	GLOBAL PDN	R/W	0h	Bit 6 (OVERRIDE PDN PIN) must be set before this bit can be programmed. 0 = Normal operation 1 = Global power-down via the SPI
6	OVERRIDE PDN PIN	R/W	0h	This bit ignores the power-down pin control. 0 = Normal operation 1 = Ignores inputs on the power-down pin
5	PDN MASK SEL	R/W	0h	This bit selects power-down mask 1 or mask 2. 0 = Power-down mask 1 1 = Power-down mask 2
4-0	0	W	0h	Must write 0

**7.5.3.2.6 Register 39h (address = 39h), Master Page (080h)**
**Figure 91. Register 39h**

7	6	5	4	3	2	1	0
PERF MODE[1:0]	0	0	0	0	0	0	0
R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 23. Register 39h Field Descriptions**

Bit	Field	Type	Reset	Description
7-6	PERF MODE[1:0]	R/W	0h	Set all four PERF MODE[3:0] bits together. Bits are located in register address 39h, 3Ah, and 56h in the master page.
5-0	0	W	0h	Must write 0

**7.5.3.2.7 Register 3Ah (address = 3Ah), Master Page (080h)**
**Figure 92. Register 3Ah**

7	6	5	4	3	2	1	0
0	PERF MODE[2]	0	0	0	0	0	0
W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 24. Register 3Ah Field Descriptions**

Bit	Field	Type	Reset	Description
7	0	W	0h	Must write 0
6	PERF MODE[2]	R/W	0h	Set all four PERF MODE[3:0] bits together. Bits are located in register address 39h, 3Ah, and 56h in the master page.
5-0	0	W	0h	Must write 0

**7.5.3.2.8 Register 4Fh (address = 4Fh), Master Page (080h)**
**Figure 93. Register 4Fh**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	EN INPUT DC COUPLING
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 25. Register 4Fh Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	EN INPUT DC COUPLING	R/W	0h	Enables dc-coupling between the analog inputs and driver by changing the internal biasing resistor between the analog inputs and VCM from 600 Ω to 5 kΩ. 0 = Disable dc-coupling support 1 = Enable dc-coupling support

**7.5.3.2.9 Register 53h (address = 53h), Master Page (080h)**
**Figure 94. Register 53h**

7	6	5	4	3	2	1	0
0	MASK SYSREF	0	0	0	0	EN SYSREF DC COUPLING	0
W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 26. Register 53h Field Descriptions**

Bit	Field	Type	Reset	Description
7	0	W	0h	Must write 0
6	MASK SYSREF	R/W	0h	0 = Normal operation 1 = Ignores the SYSREF input
5-2	0	W	0h	Must write 0
1	EN SYSREF DC COUPLING	R/W	0h	Enables a higher common-mode voltage input on the SYSREF signal (up to 1.6 V). 0 = Normal operation 1 = Enables a higher SYSREF common-mode voltage support
0	0	W	0h	Must write 0

**7.5.3.2.10 Register 55h (address = 55h), Master Page (080h)**
**Figure 95. Register 55h**

7	6	5	4	3	2	1	0
0	0	0	PDN MASK	0	0	0	0
W-0h	W-0h	W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 27. Register 55h Field Descriptions**

Bit	Field	Type	Reset	Description
7-5	0	W	0h	Must write 0
4	PDN MASK	R/W	0h	This bit enables power-down via a register bit. 0 = Normal operation 1 = Power-down is enabled by powering down internal blocks as specified in the selected power-down mask
3-0	0	W	0h	Must write 0

**7.5.3.2.11 Register 56h (address = 56h), Master Page (080h)**
**Figure 96. Register 56h**

7	6	5	4	3	2	1	0
0	0	0	0	0	PERF MODE[3]	0	0
W-0h	W-0h	W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: W = Write only; -n = value after reset

**Table 28. Register 56h Field Descriptions**

Bit	Field	Type	Reset	Description
7-3	0	W	0h	Must write 0
2	PERF MODE[3]	W	0h	Set all four PERF MODE[3:0] bits together. Bits are located in register address 39h, 3Ah, and 56h in the master page.
1-0	0	W	0h	Must write 0

**7.5.3.2.12 Register 59h (address = 59h), Master Page (080h)**
**Figure 97. Register 59h**

7	6	5	4	3	2	1	0
FOVR CHB	0	ALWAYS WRITE 1	0	0	0	0	0
W-0h	W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 29. Register 59h Field Descriptions**

Bit	Field	Type	Reset	Description
7	FOVR CHB	W	0h	Outputs the FOVR signal for channel B on the SDOUT pin. 0 = Normal operation 1 = Outputs FOVR on the SDOUT pin
6	0	W	0h	Must write 0
5	ALWAYS WRITE 1	R/W	0h	Must write 1
4-0	0	W	0h	Must write 0

**7.5.3.3 ADC Page (0Fh) Registers**
**7.5.3.3.1 Registers 5F (addresses = 5F), ADC Page (0Fh)**
**Figure 98. Register 5F**

7	6	5	4	3	2	1	0
FOVR THRESHOLD PROG							
R/W-E3h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 30. Registers 5F Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	FOVR THRESHOLD PROG	R/W	E3h	Program the fast OVR thresholds together for channel A and B, as described in the <a href="#">Overrange Indication</a> section.

### 7.5.3.4 Main Digital Page (6800h) Registers

#### 7.5.3.4.1 Register 0h (address = 0h), Main Digital Page (6800h)

**Figure 99. Register 0h**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	PULSE RESET
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 31. Register 0h Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	PULSE RESET	R/W	0h	Must be pulsed after power-up or after configuring registers in the main digital page of the JESD bank. Any register bits in the main digital page (6800h) take effect only after this bit is pulsed; see the <a href="#">Start-Up Sequence</a> section for the correct sequence. 0 = Normal operation 0 → 1 → 0 = Bit is pulsed

#### 7.5.3.4.2 Register 41h (address = 41h), Main Digital Page (6800h)

**Figure 100. Register 41h**

7	6	5	4	3	2	1	0
0	0	0	DECFIL MODE[3]	0	DECFIL MODE[2:0]		
W-0h	W-0h	W-0h	R/W-0h	W-0h	R/W-0h		

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 32. Register 41h Field Descriptions**

Bit	Field	Type	Reset	Description
7-5	0	W	0h	Must write 0
4	DECFIL MODE[3]	R/W	0h	Refer <a href="#">Table 33</a> .
3	0	W	0h	Must write 0
2-0	DECFIL MODE[2:0]	R/W	2h	These bits select the decimation filter mode. <a href="#">Table 33</a> lists the bit settings. Register bit DEC MODE EN (register 4Dh, bit 3) must also be enabled.

**Table 33. DECFIL MODE Bit Settings**

DEC MODE EN (REGISTER 4Dh, BIT 3)	BITS (4, 2-0)	FILTER MODE	DECIMATION
0	XXXX	Low-pass filter	2X
1	1010	Low-pass filter	2X
1	1110	High-pass filter	2X
1	Others	Do not use	—

**7.5.3.4.3 Register 42h (address = 42h), Main Digital Page (6800h)**
**Figure 101. Register 42h**

7	6	5	4	3	2	1	0
0	0	0	0	0	NYQUIST ZONE		
W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h		

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 34. Register 42h Field Descriptions**

Bit	Field	Type	Reset	Description
7-3	0	W	0h	Must write 0
2-0	NYQUIST ZONE	R/W	0h	The Nyquist zone must be selected for proper interleaving correction. Here Nyquist refers to Device Clock/2. For 1 GSPS Device clock, Nyquist frequency is 500 MHz. Also set register bit CTRL NYQUIST (4Eh, bit 7). 000 = 1st Nyquist zone (0 MHz to 500 MHz) 001 = 2nd Nyquist zone (500 MHz to 1000 MHz) 010 = 3rd Nyquist zone (1000 MHz to 1500 MHz) All others = Not used

**7.5.3.4.4 Register 43h (address = 43h), Main Digital Page (6800h)**
**Figure 102. Register 43h**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	FORMAT SEL
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 35. Register 43h Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	FORMAT SEL	R/W	0h	Changes the output format. Set the FORMAT EN bit (register 4Bh, bit 5) to enable control using this bit. 0 = Twos complement 1 = Offset binary

**7.5.3.4.5 Register 44h (address = 44h), Main Digital Page (6800h)**
**Figure 103. Register 44h**

7	6	5	4	3	2	1	0
0	DIGITAL GAIN						
R/W-0h	R/W-0h						

LEGEND: R/W = Read/Write; -n = value after reset

**Table 36. Register 44h Field Descriptions**

Bit	Field	Type	Reset	Description
7	0	R/W	0h	Must write 0
6-0	DIGITAL GAIN	R/W	0h	Digital gain setting. Digital gain must be enabled (register 52h, bit 0). Gain in dB = 20log (digital gain / 32). 7Fh = 127, equals digital gain of 9.5 dB.



**7.5.3.4.6 Register 4Bh (address = 4Bh), Main Digital Page (6800h)**
**Figure 104. Register 4Bh**

7	6	5	4	3	2	1	0
0	0	FORMAT EN	0	0	0	0	0
W-0h	W-0h	R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 37. Register 4Bh Field Descriptions**

Bit	Field	Type	Reset	Description
7-6	0	W	0h	Must write 0
5	FORMAT EN	R/W	0h	This bit enables control for data format selection using the FORMAT SEL register bit. 0 = Default, output is in twos complement format 1 = Output is in offset binary format after the FORMAT SEL bit is set
4-0	0	W	0h	Must write 0

**7.5.3.4.7 Register 4Dh (address = 4Dh), Main Digital Page (6800h)**
**Figure 105. Register 4Dh**

7	6	5	4	3	2	1	0
0	0	0	0	DEC MOD EN	0	0	0
W-0h	W-0h	W-0h	W-0h	R/W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 38. Register 4Dh Field Descriptions**

Bit	Field	Type	Reset	Description
7-4	0	W	0h	Must write 0
3	DEC MOD EN	R/W	0h	This bit enables control of decimation filter mode via the DECFIL MODE[3:0] register bits. 0 = Default 1 = Decimation modes control is enabled
2-0	0	W	0h	Must write 0

**7.5.3.4.8 Register 4Eh (address = 4Eh), Main Digital Page (6800h)**
**Figure 106. Register 4Eh**

7	6	5	4	3	2	1	0
CTRL NYQUIST	0	0	0	0	0	0	0
R/W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 39. Register 4Eh Field Descriptions**

Bit	Field	Type	Reset	Description
7	CTRL NYQUIST	R/W	0h	This bit enables selecting the Nyquist zone using register 42h, bits 2-0. 0 = Selection disabled 1 = Selection enabled
6-0	0	W	0h	Must write 0

**7.5.3.4.9 Register 52h (address = 52h), Main Digital Page (6800h)**
**Figure 107. Register 52h**

7	6	5	4	3	2	1	0
ALWAYS WRITE 1	0	0	0	0	0	0	DIG GAIN EN
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 40. Register 52h Field Descriptions**

Bit	Field	Type	Reset	Description
7	ALWAYS WRITE 1	W	0h	This bit enables output bus reorder using the Dx_BUS_REORDER[7:0] bits. Set this bit along with register 72h, bit 3 in the main digital page.
6-1	0	W	0h	Must write 0
0	DIG GAIN EN	R/W	0h	Enables selecting the digital gain for register 44h. 0 = Digital gain disabled 1 = Digital gain enabled

**7.5.3.4.10 Register 72h (address = 72h), Main Digital Page (6800h)**
**Figure 108. Register 72h**

7	6	5	4	3	2	1	0
0	0	0	0	ALWAYS WRITE 1	0	0	0
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: W = Write only; -n = value after reset

**Table 41. Register 72h Field Descriptions**

Bit	Field	Type	Reset	Description
7-4	0	W	0h	Must write 0
3	ALWAYS WRITE 1	W	0h	This bit enables output bus reorder using the Dx_BUS_REORDER[7:0] bits. Set this bit along with register 52h, bit 7 in the main digital page.
2-0	0	W	0h	Must write 0

**7.5.3.4.11 Register ABh (address = ABh), Main Digital Page (6800h)**
**Figure 109. Register ABh**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	LSB SEL EN
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 42. Register ABh Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	LSB SEL EN	R/W	0h	Enables control for the LSB SELECT register bit. 0 = Default 1 = The LSB of the 16-bit ADC data can be programmed as fast OVR using the LSB SELECT bit

**7.5.3.4.12 Register ADh (address = ADh), Main Digital Page (6800h)**
**Figure 110. Register ADh**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	LSB SELECT	
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h	

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 43. Register ADh Field Descriptions**

Bit	Field	Type	Reset	Description
7-2	0	W	0h	Must write 0
1-0	LSB SELECT	R/W	0h	Enables output of the FOVR flag instead of the output data LSB. 00 = Output is 16-bit data 11 = Output data LSB is replaced by the FOVR information for each channel

**7.5.3.4.13 Register F7h (address = F7h), Main Digital Page (6800h)**
**Figure 111. Register F7h**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	DIG RESET
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: W = Write only; -n = value after reset

**Table 44. Register F7h Field Descriptions**

Bit	Field	Type	Reset	Description
7-1	0	W	0h	Must write 0
0	DIG RESET	W	0h	Self-clearing reset for the digital block. Does not include the interleaving correction. 0 = Normal operation 1 = Digital reset

### 7.5.3.5 JESD Digital Page (6900h) Registers

#### 7.5.3.5.1 Register 0h (address = 0h), JESD Digital Page (6900h)

**Figure 112. Register 0h**

7	6	5	4	3	2	1	0
CTRL K	0	0	TESTMODE EN	FLIP ADC DATA	LANE ALIGN	FRAME ALIGN	TX LINK DIS
R/W-0h	W-0h	W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 45. Register 0h Field Descriptions**

Bit	Field	Type	Reset	Description
7	CTRL K	R/W	0h	Enable bit for a number of frames per multi-frame. 0 = Default is five frames per multi-frame 1 = Frames per multi-frame can be set in register 06h
6-5	0	W	0h	Must write 0
4	TESTMODE EN	R/W	0h	This bit generates the long transport layer test pattern mode, as per section 5.1.6.3 of the JESD204B specification. 0 = Test mode disabled 1 = Test mode enabled
3	FLIP ADC DATA	R/W	0h	0 = Normal operation 1 = Output data order is reversed: MSB to LSB
2	LANE ALIGN	R/W	0h	This bit inserts the lane alignment character (K28.3) for the receiver to align to the lane boundary, as per section 5.3.3.5 of the JESD204B specification. 0 = Normal operation 1 = Inserts lane alignment characters
1	FRAME ALIGN	R/W	0h	This bit inserts the lane alignment character (K28.7) for the receiver to align to the lane boundary, as per section 5.3.3.5 of the JESD204B specification. 0 = Normal operation 1 = Inserts frame alignment characters
0	TX LINK DIS	R/W	0h	This bit disables sending the initial link alignment (ILA) sequence when SYNC is de-asserted. 0 = Normal operation 1 = ILA disabled

**7.5.3.5.2 Register 1h (address = 1h), JESD Digital Page (6900h)**
**Figure 113. Register 1h**

7	6	5	4	3	2	1	0
SYNC REG	SYNC REG EN	0	0	0	JESD MODE		
R/W-0h	R/W-0h	W-0h	W-0h	W-0h	R/W-1h		

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 46. Register 1h Field Descriptions**

Bit	Field	Type	Reset	Description
7	SYNC REG	R/W	0h	Register control for sync request. 0 = Normal operation 1 = ADC output data are replaced with K28.5 characters; the SYNC REG EN register bit must also be set to 1
6	SYNC REG EN	R/W	0h	Enables register control for sync request. 0 = Use the SYNC pin for sync requests 1 = Use the SYNC REG register bit for sync requests
5-3	0	W	0h	Must write 0
2-0	JESD MODE	R/W	1h	These bits select the number of active output lanes. The JESD PLL MODE register bit located in the JESD analog page must also be set accordingly. Active lanes carry serial JESD data whereas inactive lanes don't carry any data. 001 = 20X mode, four active lanes per device (default) 010 = 40X mode, two active lanes per device All others = Not used

**7.5.3.5.3 Register 2h (address = 2h), JESD Digital Page (6900h)**
**Figure 114. Register 2h**

7	6	5	4	3	2	1	0
LINK LAYER TESTMODE		LINK LAYER RPAT		LMFC MASK RESET	0	0	0
R/W-0h		R/W-0h		R/W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 47. Register 2h Field Descriptions**

Bit	Field	Type	Reset	Description
7-5	LINK LAYER TESTMODE	R/W	0h	These bits generate a pattern according to section 5.3.3.8.2 of the JESD204B document. 000 = Normal ADC data 001 = D21.5 (high-frequency jitter pattern) 010 = K28.5 (mixed-frequency jitter pattern) 011 = Repeat initial lane alignment (generates a K28.5 character and continuously repeats lane alignment sequences) 100 = 12-octet RPAT jitter pattern All others = Not used
4	LINK LAYER RPAT	R/W	0h	This bit changes the running disparity in the modified RPAT pattern test mode (only when the link layer test mode = 100). 0 = Normal operation 1 = Changes disparity
3	LMFC MASK RESET	R/W	0h	Masks the LMFC reset coming to the digital block. 0 = LMFC reset is not masked 1 = Ignore the LMFC reset request
2-0	0	W	0h	Must write 0

**7.5.3.5.4 Register 3h (address = 3h), JESD Digital Page (6900h)**
**Figure 115. Register 3h**

7	6	5	4	3	2	1	0
FORCE LMFC COUNT	LMFC COUNT INIT					RELEASE ILANE SEQ	
R/W-0h	R/W-0h					R/W-0h	

LEGEND: R/W = Read/Write; -n = value after reset

**Table 48. Register 3h Field Descriptions**

Bit	Field	Type	Reset	Description
7	FORCE LMFC COUNT	R/W	0h	This bit forces the LMFC count. 0 = Normal operation 1 = Enables using a different starting value for the LMFC counter
6-2	MASK SYSREF	R/W	0h	When SYSREF transmits to the digital block, the LMFC count resets to 0 and K28.5 stops transmitting when the LMFC count reaches 31. The initial value that the LMFC count resets to can be set using LMFC COUNT INIT. In this manner, the receiver can be synchronized early because the LANE ALIGNMENT SEQUENCE is received early. The FORCE LMFC COUNT register bit must be enabled.
1-0	RELEASE ILANE SEQ	R/W	0h	These bits delay the generation of the lane alignment sequence by 0, 1, 2 or 3 multi-frames after the code group synchronization. 00 = 0 01 = 1 10 = 2 11 = 3

**7.5.3.5.5 Register 5h (address = 5h), JESD Digital Page (6900h)**
**Figure 116. Register 5h**

7	6	5	4	3	2	1	0
SCRAMBLE EN	0	0	0	0	0	0	0
R/W-Undefined	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 49. Register 5h Field Descriptions**

Bit	Field	Type	Reset	Description
7	SCRAMBLE EN	R/W	Undefined	Scrambles the enable bit in the JESD204B interface. 0 = Scrambling disabled 1 = Scrambling enabled
6-0	0	W	0h	Must write 0

**7.5.3.5.6 Register 6h (address = 6h), JESD Digital Page (6900h)**
**Figure 117. Register 6h**

7	6	5	4	3	2	1	0
0	0	0	FRAMES PER MULTI FRAME (K)				
W-0h	W-0h	W-0h	R/W-8h				

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 50. Register 6h Field Descriptions**

Bit	Field	Type	Reset	Description
7-5	0	W	0h	Must write 0
4-0	FRAMES PER MULTI FRAME (K)	R/W	8h	These bits set the number of multi-frames. Actual K is the value in hex + 1 (that is, 0Fh is K = 16).

**7.5.3.5.7 Register 7h (address = 7h), JESD Digital Page (6900h)**
**Figure 118. Register 7h**

7	6	5	4	3	2	1	0
0	0	0	0	SUBCLASS	0	0	0
W-0h	W-0h	W-0h	W-0h	R/W-1h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 51. Register 7h Field Descriptions**

Bit	Field	Type	Reset	Description
7-4	0	W	0h	Must write 0
3	SUBCLASS	R/W	1h	This bit sets the JESD204B subclass. 000 = Subclass 0 is backward compatible with JESD204A 001 = Subclass 1 deterministic latency using the SYSREF signal
2-0	0	W	0h	Must write 0

**7.5.3.5.8 Register 31h (address = 31h), JESD Digital Page (6900h)**
**Figure 119. Register 31h**

7	6	5	4	3	2	1	0
DA_BUS_REORDER[7:0]							
R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 52. Register 31h Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	DA_BUS_REORDER[7:0]	R/W	0h	Use these bits to program output connections between data streams and output lanes in decimate-by-2 mode. <a href="#">Table 11</a> lists the supported combinations of these bits.

**7.5.3.5.9 Register 32h (address = 32h), JESD Digital Page (6900h)**
**Figure 120. Register 32h**

7	6	5	4	3	2	1	0
DB_BUS_REORDER[7:0]							
R/W-0h							

LEGEND: R/W = Read/Write; -n = value after reset

**Table 53. Register 32h Field Descriptions**

Bit	Field	Type	Reset	Description
7-0	DB_BUS_REORDER[7:0]	R/W	0h	Use these bits to program output connections between data streams and output lanes in decimate-by-2 mode. <a href="#">Table 11</a> lists the supported combinations of these bits.



### 7.5.3.6 JESD Analog Page (6A00h) Register

#### 7.5.3.6.1 Registers 12h-5h (address = 12h-5h), JESD Analog Page (6A00h)

**Figure 121. Register 12h**

7	6	5	4	3	2	1	0
SEL EMP LANE 1						0	0
R/W-0h						W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Figure 122. Register 13h**

7	6	5	4	3	2	1	0
SEL EMP LANE 0						0	0
R/W-0h						W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Figure 123. Register 14h**

7	6	5	4	3	2	1	0
SEL EMP LANE 2						0	0
R/W-0h						W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Figure 124. Register 15h**

7	6	5	4	3	2	1	0
SEL EMP LANE 3						0	0
R/W-0h						W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 54. Registers 12h-15h Field Descriptions**

Bit	Field	Type	Reset	Description
7-2	SEL EMP LANE 1, 0, 2, or 3	R/W	0h	Selects the amount of de-emphasis for the JESD output transmitter. The de-emphasis value in dB is measured as the ratio between the peak value after the signal transition to the settled value of the voltage in one bit period. 0 = 0 dB 1 = -1 dB 3 = -2 dB 7 = -4.1 dB 15 = -6.2 dB 31 = -8.2 dB 63 = -11.5 dB
1-0	0	W-0h	0h	Must write 0

**7.5.3.6.2 Register 16h (address = 16h), JESD Analog Page (6A00h)**
**Figure 125. Register 16h**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	JESD PLL MODE	
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h	

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 55. Register 16h Field Descriptions**

Bit	Field	Type	Reset	Description
7-2	0	W	0h	Must write 0
1-0	JESD PLL MODE	R/W	0h	These bits select the JESD PLL multiplication factor and must match the JESD MODE setting. 00 = 20X mode, four active lanes per device 01 = Not used 10 = 40X mode, two active lanes per device 11 = Not used

**7.5.3.6.3 Register 1Ah (address = 1Ah), JESD Analog Page (6A00h)**
**Figure 126. Register 1Ah**

7	6	5	4	3	2	1	0
0	0	0	0	0	0	FOVR CHA	0
W-0h	W-0h	W-0h	W-0h	W-0h	W-0h	R/W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 56. Register 1Ah Field Descriptions**

Bit	Field	Type	Reset	Description
7-2	0	W	0h	Must write 0
1	FOVR CHA	R/W	0h	Outputs the FOVR signal for channel A on the PDN pin. FOVR CHA EN (register 1Bh, bit 3) must be enabled. 0 = Normal operation 1 = FOVR on the PDN pin
0	0	W	0h	Must write 0

**7.5.3.6.4 Register 1Bh (address = 1Bh), JESD Analog Page (6A00h)**
**Figure 127. Register 1Bh**

7	6	5	4	3	2	1	0
JESD SWING			0	FOVR CHA EN	0	0	0
R/W-0h			W-0h	R/W-0h	W-0h	W-0h	W-0h

LEGEND: R/W = Read/Write; W = Write only; -n = value after reset

**Table 57. Register 1Bh Field Descriptions**

Bit	Field	Type	Reset	Description
7-5	JESD SWING	R/W	0h	Selects the output differential amplitude $V_{OD}$ (mV <sub>PP</sub> ) of the JESD transmitter (for all lanes). 0 = 860 mV <sub>PP</sub> 1 = 810 mV <sub>PP</sub> 2 = 770 mV <sub>PP</sub> 3 = 745 mV <sub>PP</sub> 4 = 960 mV <sub>PP</sub> 5 = 930 mV <sub>PP</sub> 6 = 905 mV <sub>PP</sub> 7 = 880 mV <sub>PP</sub>
4	0	W	0h	Must write 0
3	FOVR CHA EN	R/W	0h	Enables overwriting the PDN pin with the FOVR signal from channel A. 0 = Normal operation 1 = PDN is overwritten
2-0	JESD PLL MODE	R/W	0h	Must write 0

## 8 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

#### 8.1.1 Start-Up Sequence

The steps described in [Table 58](#) are the recommended power-up sequence with the ADS54J69 in 20X or 40X mode.

**Table 58. Initialization Sequence**

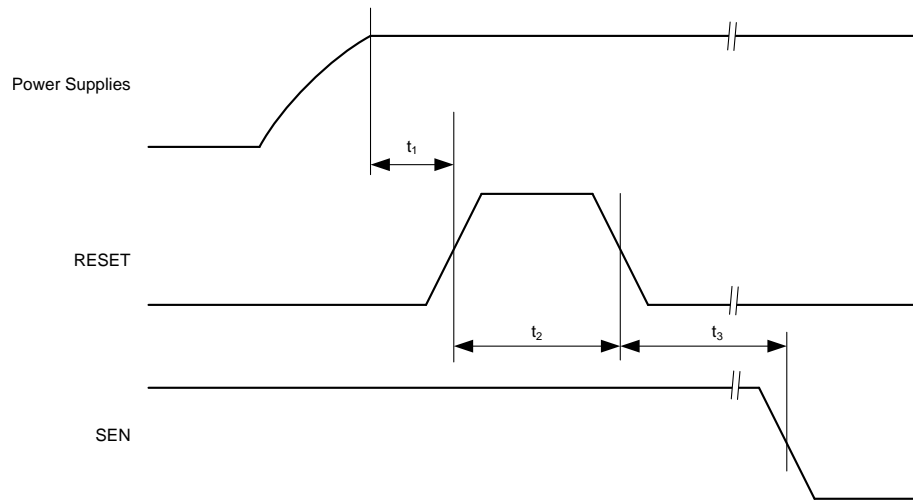
STEP	SEQUENCE	DESCRIPTION	PAGE BEING PROGRAMMED	COMMENT
1	Power-up the device	<b>Power up:</b> Bring up the supplies to IOVDD = 1.15 V, DVDD = AVDD = 1.9 V, and AVDD3V = 3.0 V.	—	These supplies can be brought up in any order.
2	Reset the device	<b>Hardware reset</b>		
		<b>Apply a hardware reset by pulsing pin 48 (low-&gt;high-&gt;low).</b>	—	A hardware reset clears all registers to their default values.
		<b>Software reset:</b> Register writes equivalent to a hardware reset are:		
		Write address 0-000h with 81h.	General register	Reset registers in the ADC page and master page of the analog bank This bit is a self-clearing bit. This bit is a self-clearing bit.
		Write address 4-001h with 00h and address 4-002h with 00h.	Unused page	Clear any unwanted content from the unused pages of the JESD bank.
		Write address 4-003h with 00h and address 4-004h with 68h.	—	Select the main digital page of the JESD bank.
		Write address 6-0F7h with 01h for channel A.	Main digital page (JESD bank)	Use the DIG RESET register bit to reset all pages in the JESD bank. This bit is a self-clearing bit.
Write address 6-000h with 01h, then address 6-000h with 00h.	Pulse the PULSE RESET register bit for both channels.			
3	Performance modes	Write address 0-011h with 80h.	—	Select the master page of the analog bank.
		Write address 0-039h with C0h, address 0-03Ah with 40h, and address 0-056h with 04h.	Master page (analog bank)	Set the PERF MODE[3:0] bits.
		Write address 0-059h with 20h.		Set the ALWAYS WRITE 1 bit.
4	Program registers for 20X or 40X serialization and program the HPF or LPF filter	<b>The JESD mode (in the JESD digital page) and JESD PLL mode (in the JESD analog page) register bits control 20X or 40X serialization. By default after reset, the device is in 20X serialization mode (4-lanes output).</b>		
		Write address 4-003h with 00h and address 4-004h with 69h.	—	Select the JESD digital page.
		Write address 6-000h with 80h.	JESD digital page (JESD bank)	Set the CTRL K bit for both channels to program K for the SYSREF signal frequency in step 5.
		Write address 6-001h with 01h.		Enable 20X serialization (4-lane output, default setting after reset).
		Write address 6-001h with 02h.		Enable 40X serialization (2-lane output).
		Write address 6-031h with 0Ah.		Set the ALWAYS WRITE 1 bit.
		Write address 6-032h with 0Ah		
		Write address 4-003h with 00h and address 4-004h with 6Ah.	JESD analog page (JESD bank)	Select the JESD analog page.
		Write address 6-016h with 00h		Enable 20X serialization (4-lane output, default setting after reset).
		Write address 6-016h with 02h		To enable 40X serialization (2-lane output).
		Write address 4-003h with 00h and address 4-004h with 68h.	Main digital page (JESD bank)	Select the main digital page.
		Write address 6-052h with 80h and address 6-072h with 08h.		Set the ALWAYS WRITE 1 bit (enables correct order of the JESD output lanes).
		Write address 6-04Dh with 08h		Enable the decimation filter programming.
		Write address 6-041h with 12h		Enable the low-pass filter (default setting after reset).
		Write address 6-041h with 16h		Enable the high-pass filter.
Write address 6-000h with 01h and address 6-000h with 00h.	Pulse the PULSE RESET register bit. All settings programmed in the main digital page take effect only after this bit is pulsed.			

**Table 58. Initialization Sequence (continued)**

STEP	SEQUENCE	DESCRIPTION	PAGE BEING PROGRAMMED	COMMENT
5	Set the value of K and the SYSREF signal frequency accordingly	Write address 4-003h with 00h and address 4-004h with 69h.	—	Select the JESD digital page.
		Write address 6-006h with XXh (choose the value of K).	JESD digital page (JESD bank)	Default value of K is 8 for 20X (4-lane) mode and 4 for 40X (2-lane) mode. However, K can be programmed for higher values than the default by using bits 4-0 of address 6-006 in the JESD digital page. For example, if K = 31 by writing address 6-006h with 1Fh in the JESD digital page, then the SYSREF signal frequency must be kept less than or equal to $250 \text{ MHz} / 32 = 7.8125 \text{ MHz}$ .
6	JESD lane alignment	Pull the $\overline{\text{SYNC}}$ pin (pin 63) low.	—	Transmit K28.5 characters.
		Pull the $\overline{\text{SYNC}}$ pin high.		After the receiver is synchronized, initiate an ILA phase and subsequent transmissions of ADC data.

### 8.1.2 Hardware Reset

Figure 128 and Table 59 show the timing for a hardware reset.



**Figure 128. Hardware Reset Timing Diagram**

**Table 59. Timing Requirements for Figure 128**

		MIN	TYP	MAX	UNIT
t <sub>1</sub>	Power-on delay from power-up to active high RESET pulse	1			ms
t <sub>2</sub>	Reset pulse duration: active high RESET pulse duration	10			ns
t <sub>3</sub>	Register write delay: delay from RESET disable to SEN active	100			ns

### 8.1.3 SNR and Clock Jitter

The signal-to-noise ratio (SNR) of the ADC is limited by three different factors: quantization noise, thermal noise, and jitter, as shown in Equation 4. The quantization noise is typically not noticeable in pipeline converters and is 98 dB for a 16-bit ADC. The thermal noise limits SNR at low input frequencies and the clock jitter sets SNR for higher input frequencies. The decimation-by-2 process gives approximately an additional 3-dB improvement in SNR.

$$SNR_{ADC} [dBc] = -3 - 20 \log \sqrt{\left(10^{-\frac{SNR_{Quantization\ Noise}}{20}}\right)^2 + \left(10^{-\frac{SNR_{Thermal\ Noise}}{20}}\right)^2 + \left(10^{-\frac{SNR_{Jitter}}{20}}\right)^2} \quad (4)$$

The SNR limitation resulting from the sample clock jitter can be calculated by Equation 5:

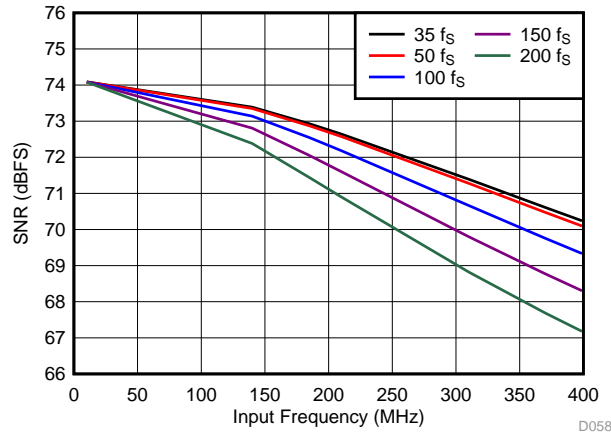
$$SNR_{jitter} [dBc] = -20 \log(2\pi \times f_{in} \times T_{jitter}) \quad (5)$$

The total clock jitter ( $T_{jitter}$ ) has two components: the internal aperture jitter ( $145 f_s$ ) is set by the noise of the clock input buffer and the external clock jitter.  $T_{jitter}$  can be calculated by Equation 6:

$$T_{jitter} = \sqrt{(T_{jitter, Ext\_Clock\_Input})^2 + (T_{Aperture\_ADC})^2} \quad (6)$$

External clock jitter can be minimized by using high-quality clock sources and jitter cleaners as well as band-pass filters at the clock input. A faster clock slew rate also improves the ADC aperture jitter.

The ADS54J69 has a thermal noise of approximately 71.1 dBFS and an internal aperture jitter of 145  $f_s$ . The SNR, depending on the amount of external jitter for different input frequencies, is shown in Figure 129.



**Figure 129. SNR vs Input Frequency and External Clock Jitter**

Half-band decimation filtering employed by the ADS54J69 reduces the affect of all contributors to SNR by 3 dB. Filtering makes the SNR curve in Figure 129 start at 74 dBFS despite a thermal noise of 71.1 dBFS.

Decimation filtering also improves the affect of jitter noise by 3 dB, and is equivalent to having 102  $f_s$  as the effective aperture jitter instead of 145  $f_s$ .

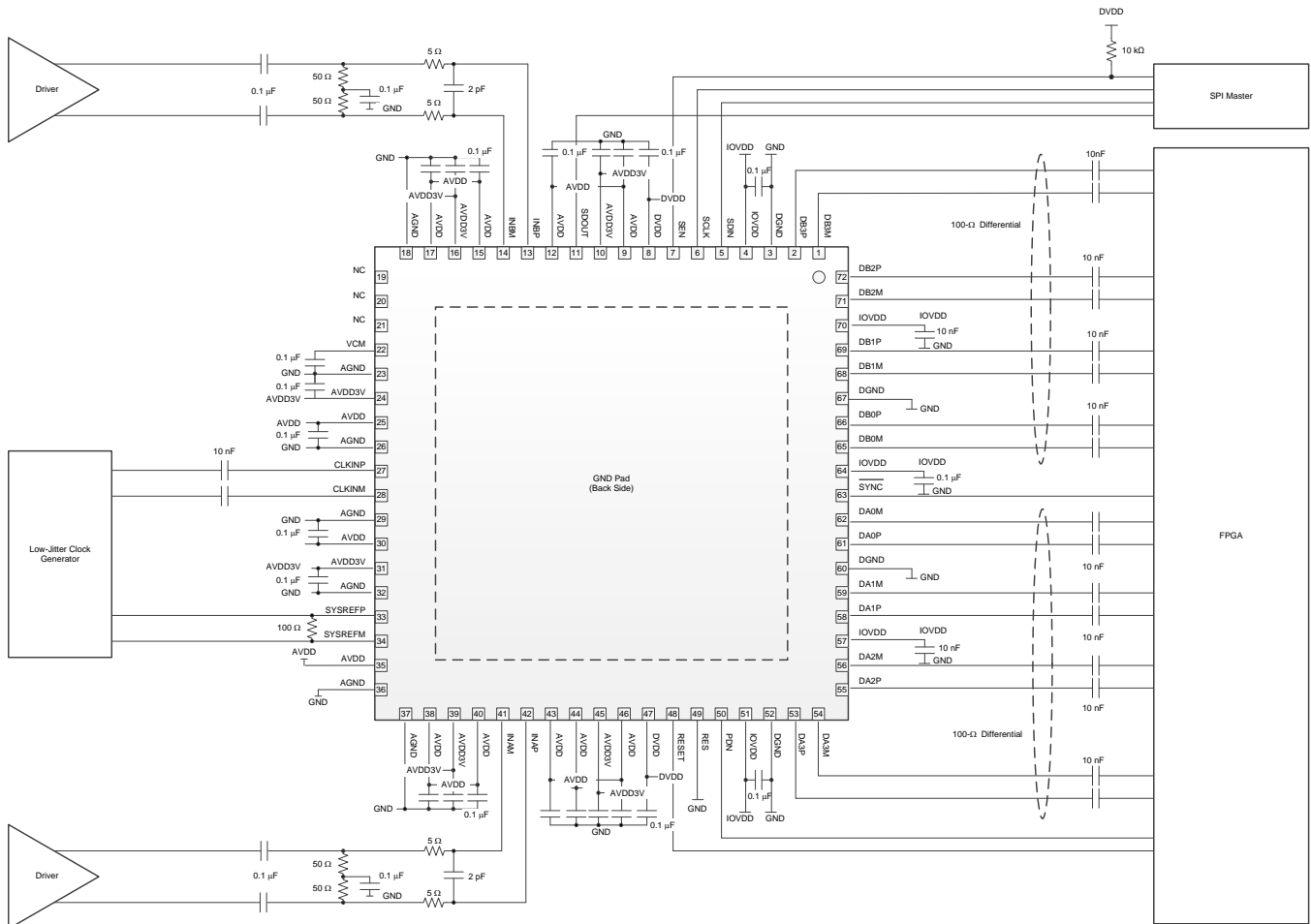
ADS54J69

SBAS713B –MAY 2015–REVISED FEBRUARY 2016

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8.2 Typical Application

The ADS54J69 is designed for wideband receiver applications demanding excellent dynamic range over a large input frequency range. A typical schematic for an ac-coupled receiver is shown in Figure 130.



NOTE: GND = AGND and DGND connected in the PCB layout.

Figure 130. AC-Coupled Receiver

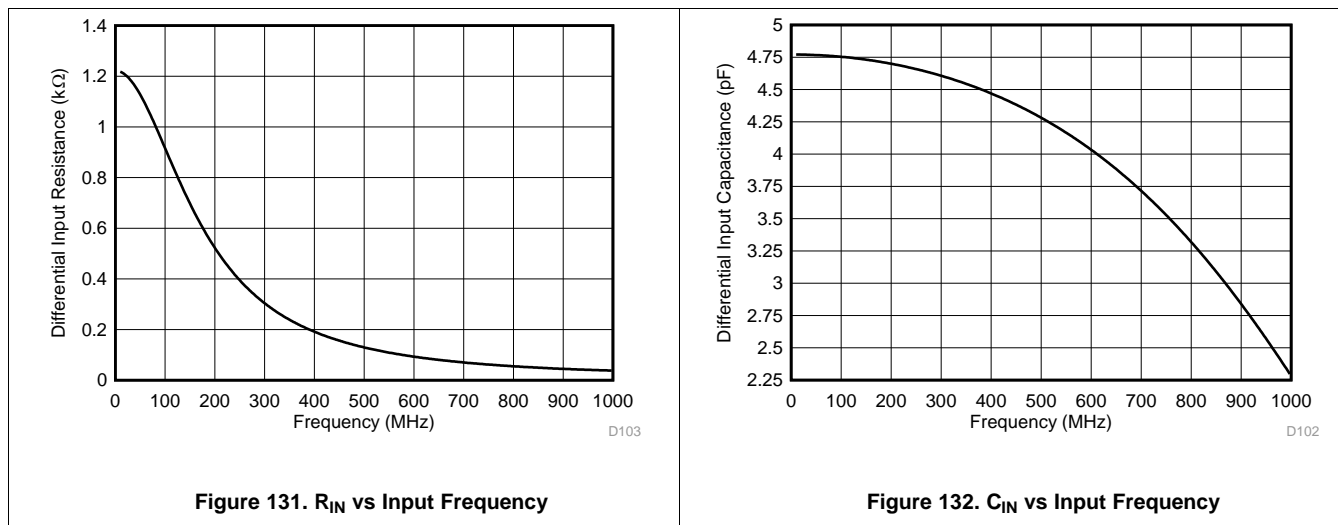


## Typical Application (continued)

### 8.2.1 Design Requirements

#### 8.2.1.1 Transformer-Coupled Circuits

Typical applications involving transformer-coupled circuits are discussed in this section. Transformers (such as ADT1-1WT or WBC1-1) can be used up to 300 MHz to achieve good phase and amplitude balances at the ADC inputs. When designing dc driving circuits, the ADC input impedance must be considered. Figure 131 and Figure 132 show the impedance ( $Z_{IN} = R_{IN} \parallel C_{IN}$ ) across the ADC input pins.



By using the simple drive circuit of Figure 133, uniform performance can be obtained over a wide frequency range. The buffers present at the analog inputs of the device help isolate the external drive source from the switching currents of the sampling circuit.

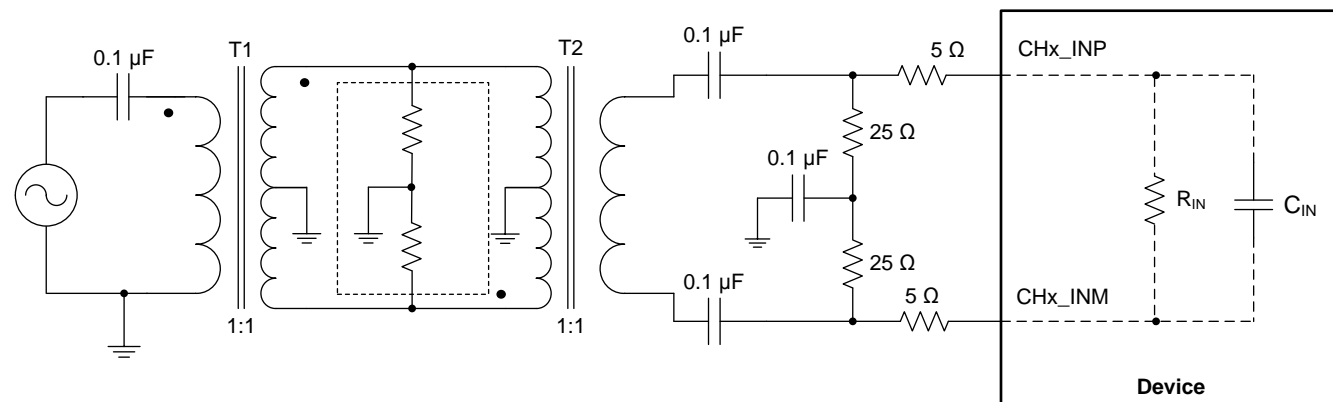


Figure 133. Input Drive Circuit

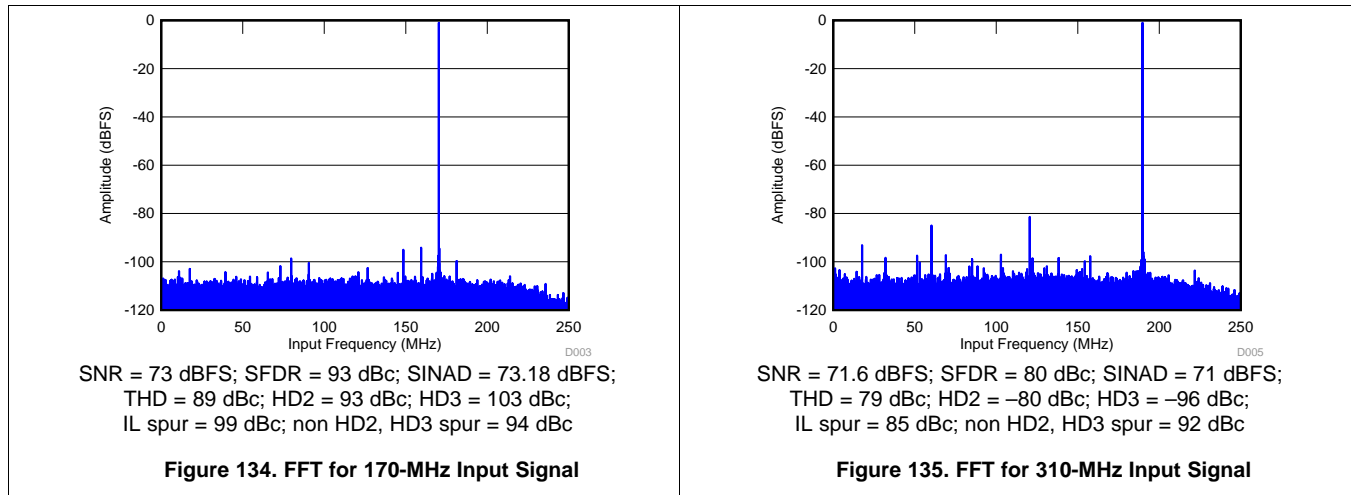
### 8.2.2 Detailed Design Procedure

For optimum performance, the analog inputs must be driven differentially. This architecture improves common-mode noise immunity and even-order harmonic rejection. A small resistor (5 Ω to 10 Ω) in series with each input pin is recommended to damp out ringing caused by package parasitics, as shown in Figure 133.

## Typical Application (continued)

### 8.2.3 Application Curves

and show the typical performance at 170 MHz and 230 MHz, respectively.



## 9 Power Supply Recommendations

The device requires a 1.8-V nominal supply for DVDD, a 1.9-V nominal supply for AVDD, and a 3.0-V nominal supply for AVDD3V. There is no specific sequence for power-supply requirements during device power-up. AVDD, DVDD, and AVDD3V can power-up in any order.

## 10 Layout

### 10.1 Layout Guidelines

The device evaluation module (EVM) layout can be used as a reference layout to obtain the best performance. A layout diagram of the EVM top layer is provided in [Figure 136](#). A complete layout of the EVM is available from the [ADS54J69EVM folder](#). Some important points to remember during board layout are:

- Analog inputs are located on opposite sides of the device pinout to ensure minimum crosstalk on the package level. To minimize crosstalk onboard, the analog inputs must exit the pinout in opposite directions, as illustrated in the reference layout of [Figure 136](#) as much as possible.
- In the device pinout, the sampling clock is located on a side perpendicular to the analog inputs in order to minimize coupling between them. This configuration is also maintained on the reference layout of [Figure 136](#) as much as possible.
- Keep digital outputs away from the analog inputs. When these digital outputs exit the pinout, the digital output traces must not be kept parallel to the analog input traces because this configuration can result in coupling from the digital outputs to the analog inputs and degrade performance. All digital output traces to the receiver [such as a field-programmable gate array (FPGA) or an application-specific integrated circuit (ASIC)] must be matched in length to avoid skew among outputs.
- At each power-supply pin (AVDD, DVDD, or AVDDD3V), keep a 0.1- $\mu$ F decoupling capacitor close to the device. A separate decoupling capacitor group consisting of a parallel combination of 10- $\mu$ F, 1- $\mu$ F, and 0.1- $\mu$ F capacitors can be kept close to the supply source.

## 10.2 Layout Example

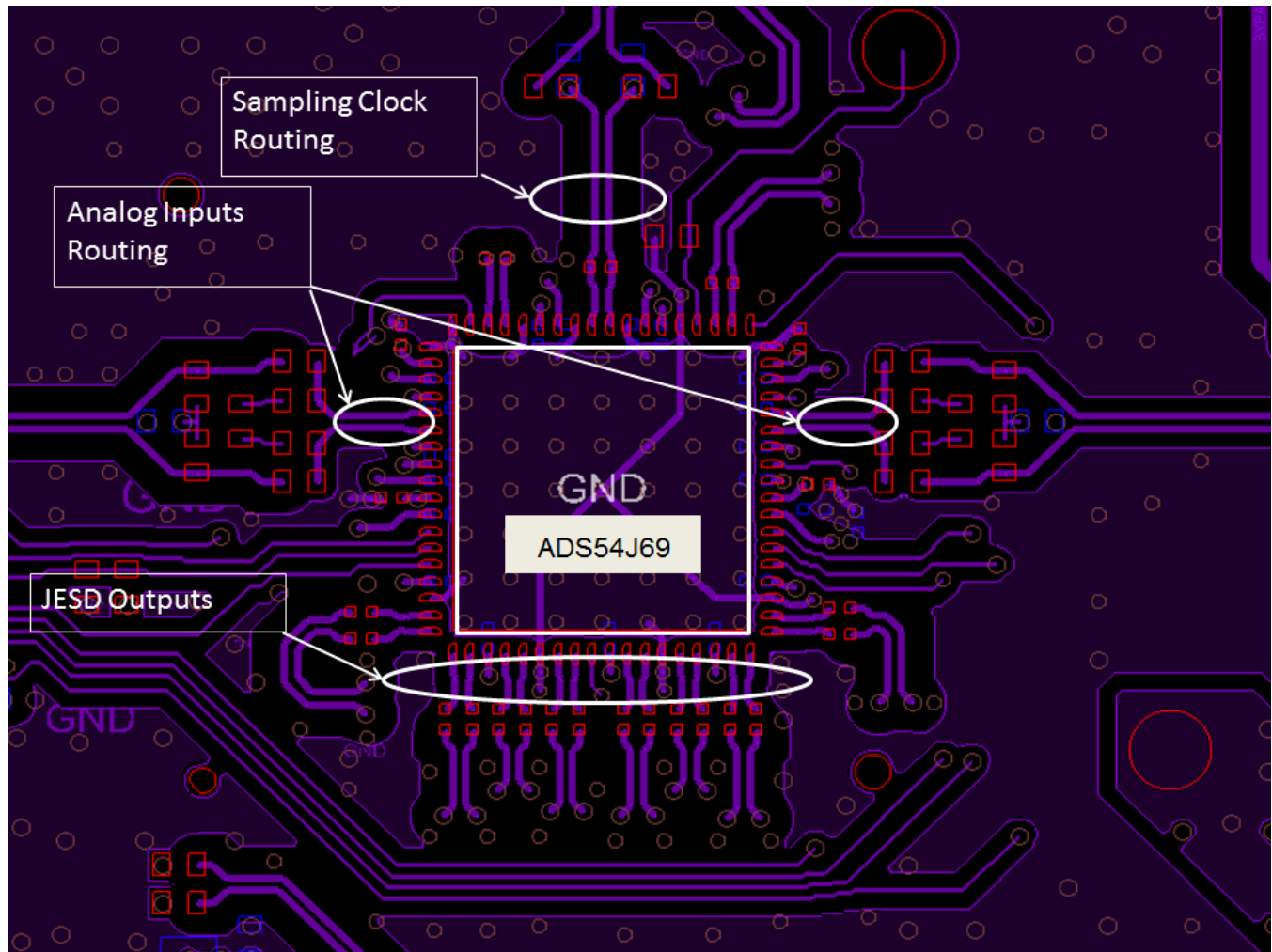


Figure 136. ADS54J69EVM layout

## 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

ADS54J40 Data Sheet, [SBAS714](#)

ADS54J60 Data Sheet, [SBAS706](#)

ADS54J69EVM User's Guide, [SLAU662](#)

### 11.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

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### 11.3 Trademarks

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### 11.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 11.5 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ADS54J69IRMP	ACTIVE	VQFN	RMP	72	168	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	AZ54J69	<a href="#">Samples</a>
ADS54J69IRMPT	ACTIVE	VQFN	RMP	72	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	AZ54J69	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

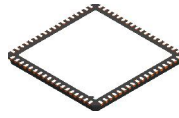
(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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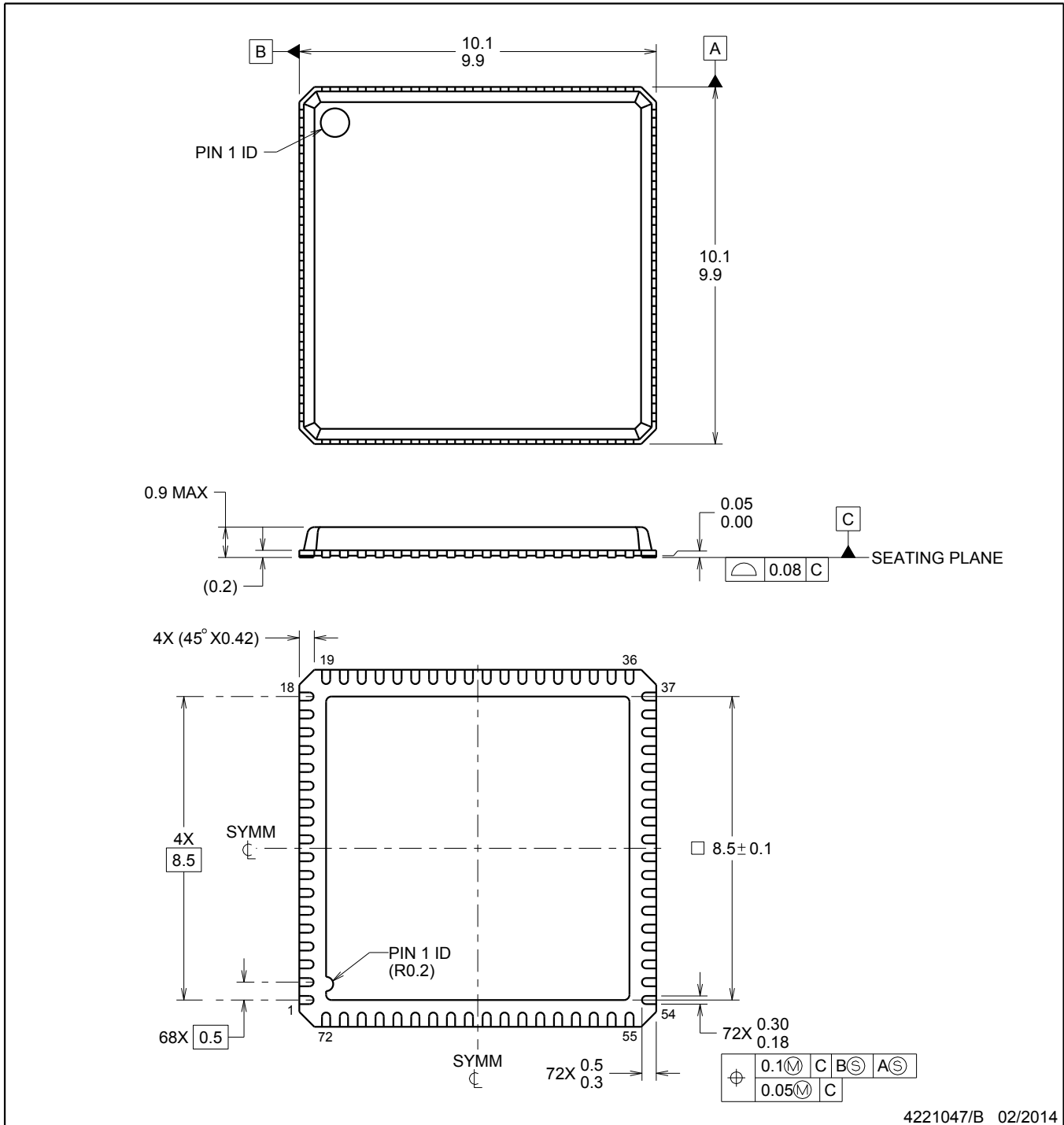
# RMP0072A



# PACKAGE OUTLINE

## VQFN - 0.9 mm max height

VQFN



4221047/B 02/2014

**NOTES:**

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

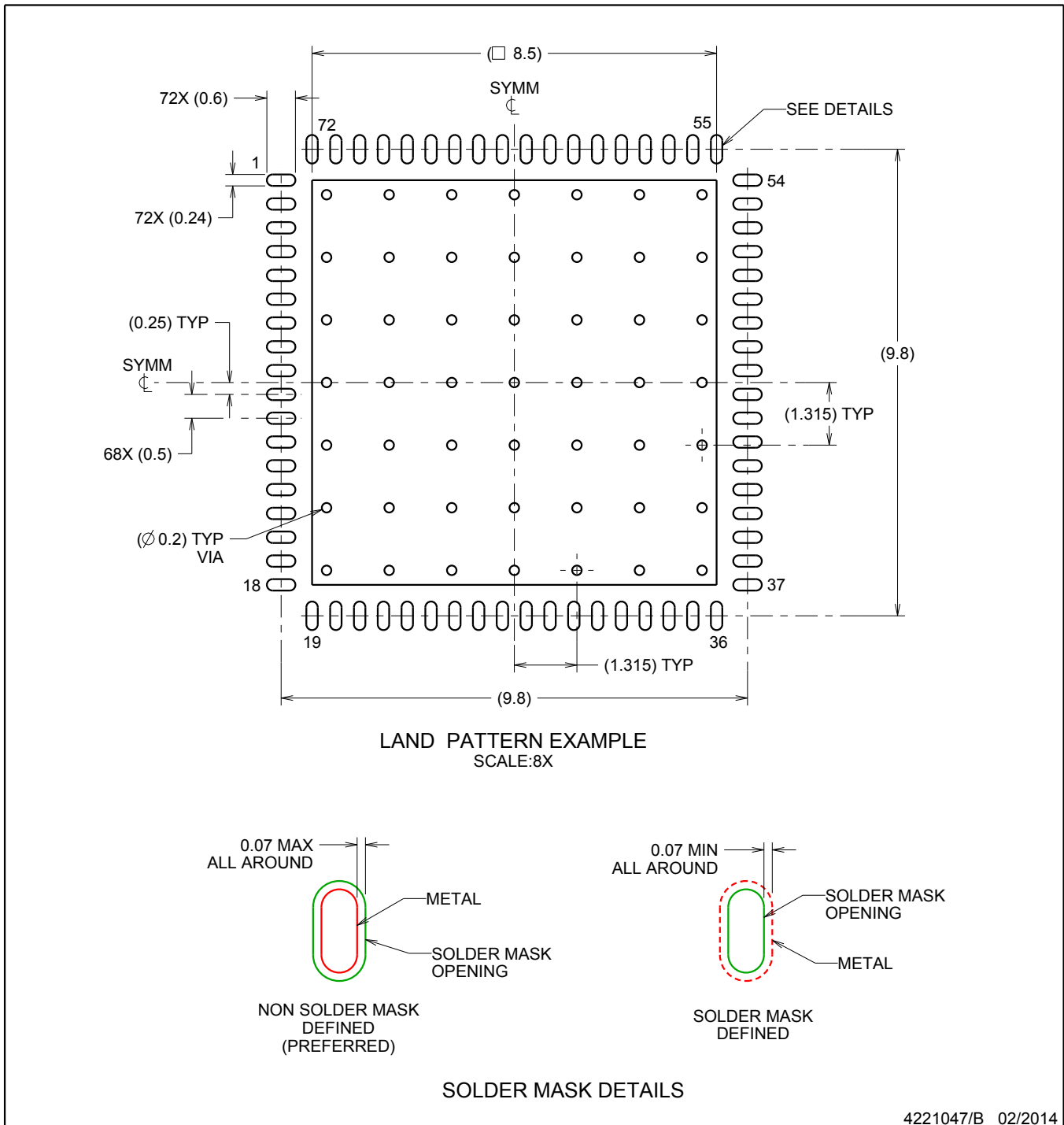


# EXAMPLE BOARD LAYOUT

RMP0072A

VQFN - 0.9 mm max height

VQFN



4221047/B 02/2014

NOTES: (continued)

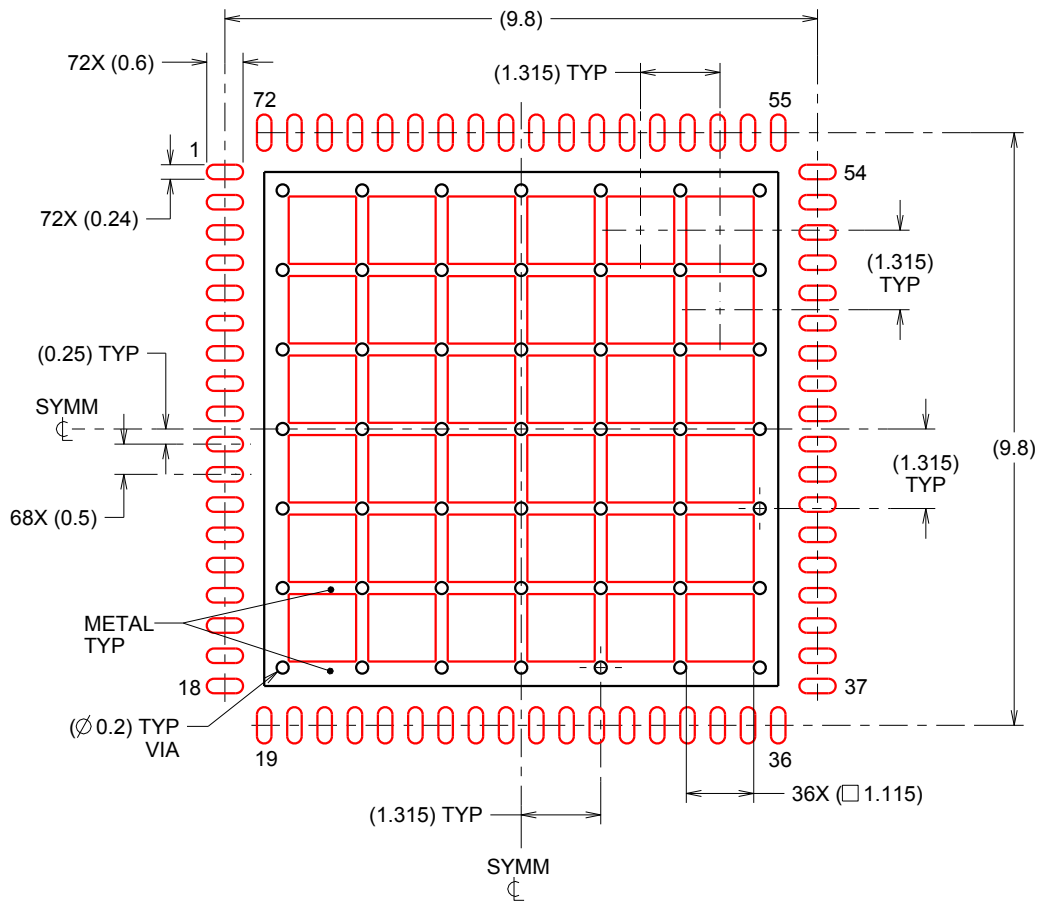
- This package is designed to be soldered to a thermal pad on the board. For more information, see QFN/SON PCB application report in literature No. SLUA271 ([www.ti.com/lit/slua271](http://www.ti.com/lit/slua271)).

# EXAMPLE STENCIL DESIGN

RMP0072A

VQFN - 0.9 mm max height

VQFN



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD  
62% PRINTED SOLDER COVERAGE BY AREA  
SCALE:8X

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NOTES: (continued)

5. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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